Data sheet

# **BMA280**Digital, triaxial acceleration sensor

**Bosch Sensortec** 





#### **BMA280: Data sheet**

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#### **BMA280**

# 14 BIT, DIGITAL, TRIAXIAL ACCELERATION SENSOR WITH INTELLIGENT ON-CHIP MOTION-TRIGGERED INTERRUPT CONTROLLER

#### **Key features**

Ultra-Small package

Digital interface

Programmable functionality

On-chip FIFO

On-chip interrupt controller

LGA package (12 pins), footprint 2mm x 2mm,

height 0.95mm

SPI (4-wire, 3-wire), I2C, 2 interrupt pins

V<sub>DDIO</sub> voltage range: 1.2V to 3.6V

Acceleration ranges ±2g/±4g/±8g/±16g Low-pass filter bandwidths 500Hz - <8Hz

up to an max. output data read out of 2kHz (unfiltered)

Integrated FIFO with a depth of 32 frames

Motion-triggered interrupt-signal generation for

- new data

- any-motion (slope) detection

- tap sensing (single tap / double tap)

- orientation recognition

- flat detection

- low-g/high-g detection

- no-motion / inactivity detection

Low current consumption, short wake-up time, advanced features for system power management

Ultra-low power

Temperature sensor

RoHS compliant, halogen-free

#### **Typical applications**

- Display profile switching
- Menu scrolling, tap / double tap sensing
- Gaming
- Pedometer / step counting
- Free-fall detection
- E-compass tilt compensation
- · Drop detection for warranty logging
- Advanced system power management for mobile applications

#### General description

The BMA280 is a triaxial, low-g acceleration sensor with digital output for consumer applications. It allows measurements of acceleration in three perpendicular axes. An evaluation circuitry (ASIC) converts the output of a micromechanical acceleration-sensing structure (MEMS) that works according to the differential capacitance principle.

Package and interfaces of the BMA280 have been defined to match a multitude of hardware requirements. Since the sensor features an ultra-small footprint and a flat package it is ingeniously suited for mobile applications.



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The BMA280 offers a variable  $V_{\text{DDIO}}$  voltage range from 1.2V to 3.6V and can be programmed to optimize functionality, performance and power consumption in customer specific applications. In addition it features an on-chip interrupt controller enabling motion-based applications without use of a microcontroller.

The BMA280 senses tilt, motion, inactivity and shock vibration in cell phones, handhelds, computer peripherals, man-machine interfaces, virtual reality features and game controllers.

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# 1. Specification

Unless stated otherwise, the given values are over lifetime, operating temperature and voltage ranges. Minimum/maximum values are  $\pm 3\sigma$ .

Table 1: Parameter specification

OPERATING CONDITIONS						
Parameter	Symbol	Condition	Min	Тур	Max	Units
	<b>g</b> FS2g			±2		g
Acceleration	<b>g</b> FS4g	Selectable via serial digital		±4		g
Range	<b>g</b> FS8g	interface		±8		g
	<b>g</b> FS16g			±16		g
Supply Voltage Internal Domains	$V_{DD}$		1.62	2.4	3.6	V
Supply Voltage I/O Domain	$V_{\text{DDIO}}$		1.2	2.4	3.6	V
Voltage Input Low Level	V <sub>IL</sub>	SPI & I <sup>2</sup> C			0.3V <sub>DDIO</sub>	-
Voltage Input High Level	$V_{IH}$	SPI & I <sup>2</sup> C	$0.7V_{\text{DDIO}}$			-
Voltage Output Low Level	V <sub>OL</sub>	I <sub>OL</sub> = 3mA, SPI & I <sup>2</sup> C			0.2V <sub>DDIO</sub>	-
Voltage Output High Level	$V_{OH}$	I <sub>OH</sub> = 3mA, SPI	0.8V <sub>DDIO</sub>			-
Total Supply Current in Normal Mode	I <sub>DD</sub>	$T_A=25$ °C, ODR <sub>max</sub> . $V_{DD}=V_{DDIO}=2.4V$		130		μΑ
Total Supply Current in Suspend Mode	$I_{DDsum}$	$T_A=25$ °C $V_{DD}=V_{DDIO}=2.4V$		2.1		μΑ
Total Supply Current in Deep Suspend Mode	$I_{ m DDdsum}$	$T_A=25$ °C $V_{DD}=V_{DDIO}=2.4V$		1		μΑ
Total Supply Current in Low-power Mode 1	I <sub>DDlp1</sub>	$T_A$ =25°C, unfiltered $V_{DD}$ = $V_{DDIO}$ = 2.4V sleep duration = 25ms		6.5		μΑ



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Total Supply Current in Low-power Mode 2	I <sub>DDlp2</sub>	$T_A$ =25°C, unfiltered $V_{DD}$ = $V_{DDIO}$ = 2.4V sleep duration = 25ms		66		μА
Total Supply Current in Standby Mode	$I_{DDsbm}$	$T_A=25^{\circ}C$ $V_{DD}=V_{DDIO}=2.4V$		62		μΑ
Wake-Up Time 1	t <sub>w,up1</sub>	from Low-power Mode 1 or Suspend Mode or Deep Suspend Mode, ODR <sub>max.</sub>		1.3	1.8	ms
Wake-Up Time 2	$t_{w,up2}$	from Low-power Mode 2 or Stand-by Mode, ODR <sub>max.</sub>		1	1.2	ms
Start-Up Time	t <sub>s,up</sub>	POR, ODR <sub>max</sub> .			3	ms
Non-volatile memory (NVM) write-cycles	n <sub>NVM</sub>				15	cycles
Operating Temperature	T <sub>A</sub>		-40		+85	°C

#### **OUTPUT SIGNAL**

Parameter	Symbol	Condition	Min	Тур	Max	Units
	$S_{2g}$	$g_{FS2g}$ , $T_A=25$ °C		4096		LSB/g
Sensitivity	S <sub>4g</sub>	$g_{FS4g}$ , $T_A=25$ °C		2048		LSB/g
Sensitivity	S <sub>8g</sub>	$g_{FS8g}$ , $T_A=25$ °C		1024		LSB/g
	S <sub>16g</sub>	$g_{FS16g}$ , $T_A=25$ °C		512		LSB/g
Sensitivity Temperature Drift	TCS	$g_{FS2g}$ , Nominal $V_{DD}$ supplies		0.015		%/K
Zero-g Offset	Off	$g_{FS2g}$ , $T_A$ =25°C, nominal $V_{DD}$ supplies, over life-time		±50		mg
Zero-g Offset Temperature Drift	TCO	$g_{FS2g}$ , Nominal $V_{DD}$ supplies		±1.0		mg/K
	bw <sub>8</sub>			8		Hz
	bw <sub>16</sub>			16		Hz
	bw <sub>31</sub>	2 <sup>nd</sup> order filter, bandwidth		31		Hz
Bandwidth	bw <sub>63</sub>			63		Hz
	bw <sub>125</sub>	programmable		125		Hz
	bw <sub>250</sub>			250		Hz
	bw <sub>500</sub>			500		Hz
Output Data Rate	$ODR_{max.}$	unfiltered		2000		Hz
Nonlinearity	NL	best fit straight line, g <sub>FS2g</sub>		±0.5		%FS

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Output Noise Density	n <sub>rms</sub>	$g_{FS2g}$ , $T_A$ =25°C Nominal $V_{DD}$ supplies Normal mode		120		µg/√Hz
Temperature Sensor Measurement Range	Ts		-40		85	°C
Temperature Sensor Slope	dTs			0.5		K/LSB
Temperature Sensor Offset	OTs			±2		К
		MECHANICAL CHARACT	EDISTICS			
Dawawastaw	Cl			T	NA	11:4
Parameter	Symbol	Condition	Min	Тур	Max	Units
Cross Axis Sensitivity	S	relative contribution between any two of the three axes		1		%
Alignment Error	E <sub>A</sub>	relative to package outline		±0.5		0



# 2. Absolute maximum ratings

Table 2: Absolute maximum ratings

Parameter	Condition	Min	Max	Units
Voltago at Supply Din	$V_{DD}$ Pin	-0.3	4.25	V
Voltage at Supply Pin	V <sub>DDIO</sub> Pin	-0.3	4.25	V
Voltage at any Logic Pin	Non-Supply Pin	-0.3	V <sub>DDIO</sub> +0.3	V
Passive Storage Temp. Range	≤ 65% rel. H.	-50	+150	°C
None-volatile memory (NVM)  Data Retention	T = 85°C, after 15 cycles	10		у
	Duration ≤ 200µs		10,000	g
Mechanical Shock	Duration ≤ 1.0ms		2,000	g
	Free fall onto hard surfaces		1.8	m
	HBM, at any Pin		2	kV
ESD	CDM		500	V
	MM		200	V

#### Note:

Stress above these limits may cause damage to the device. Exceeding the specified electrical limits may affect the device reliability or cause malfunction.



# 3. Block diagram

Figure 1 shows the basic building blocks of the BMA280:

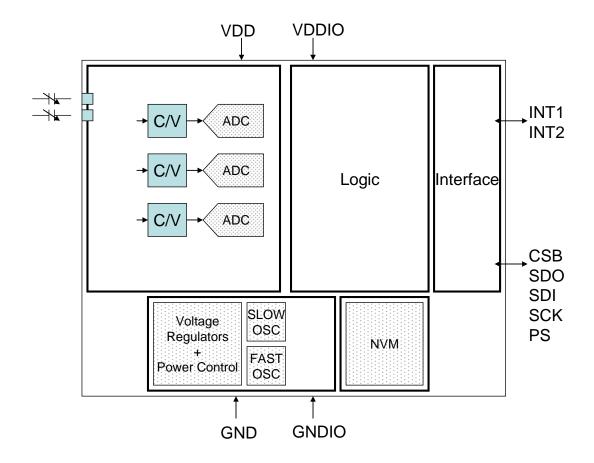


Figure 1: Block diagram of BMA280



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#### 4. Functional description

Note: Default values for registers can be found in chapter 6.

#### 4.1 Supply voltage and power management

The BMA280 has two distinct power supply pins:

- V<sub>DD</sub> is the main power supply for the internal blocks;
- V<sub>DDIO</sub> is a separate power supply pin used for supplying power for the interface

There are no limitations on the voltage levels of both pins relative to each other, as long as each of them lies within its operating range. Furthermore, the device can be completely switched off (VDD = 0V) while keeping the VDDIO supply on (VDDIO > 0V) or vice versa.

When the  $V_{DDIO}$  supply is switched off, all interface pins (CSB, SDI, SCK, PS) must be kept close to  $GND_{IO}$  potential.

The device contains a power-on reset (POR) generator. It resets the logic part and the register values after powering-on  $V_{DD}$  and  $V_{DDIO}$ . Please note, that all application specific settings which are not equal to the default settings (refer to 6.2 register map), must be re-set to its designated values after POR.

There are no constraints on the switching sequence of both supply voltages. In case the  $I^2C$  interface shall be used, a direct electrical connection between  $V_{DDIO}$  supply and the PS pin is needed in order to ensure reliable protocol selection. For SPI interface mode the PS pin must be directly connected to  $GND_{IO}$ .



#### 4.2 Power modes

The BMA280 has six different power modes. Besides normal mode, which represents the fully operational state of the device, there are five energy saving modes: deep-suspend mode, suspend mode, standby mode, low-power mode 1 and low-power mode 2.

The possible transitions between the power modes are illustrated in figure 2:

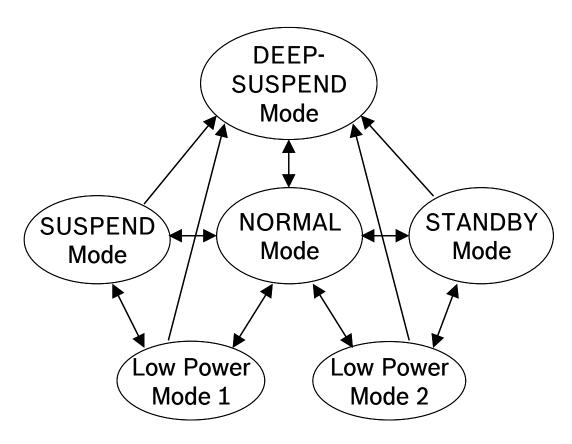


Figure 2: Power mode transition diagram

After **power-up** BMA280 is in normal mode so that all parts of the device are held powered-up and data acquisition is performed continuously.

In **deep-suspend** mode the device reaches the lowest possible power consumption. Only the interface section is kept alive. No data acquisition is performed and the content of the configuration registers is lost. Deep suspend mode is entered (left) by writing '1' ('0') to the (0x11) deep\_suspend bit while (0x11) suspend bit is set to '0'. The I<sup>2</sup>C watchdog timer remains functional. The (0x11) deep\_ suspend bit, the (0x34) spi3 bit, (0x34) i2c\_wdt\_en bit and the (0x34) i2c\_wdt\_sel bit are functional in deep-suspend mode. Equally the interrupt level and driver configuration registers (0x20) int1\_lvl, (0x20) int1\_od, (0x20) int2\_lvl, and (0x20) int2\_od are accessible. Still it is possible to enter normal mode by performing a softreset as described in chapter 4.8. Please note, that all application specific settings which are not equal to the default settings (refer to 6.2 register map), must be re-set to its designated values after leaving deep-suspend mode.



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In **suspend mode** the whole analog part is powered down. No data acquisition is performed. While in suspend mode the latest acceleration data and the content of all configuration registers are kept. Writing to and reading from registers is supported except from the (0x3E) fifo\_config\_1, (0x30) fifo\_config\_0 and (0x3F) fifo\_data register. It is possible to enter normal mode by performing a softreset as described in chapter 4.8.

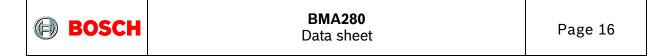
Suspend mode is entered (left) by writing '1' ('0') to the (0x11) suspend bit after bit (0x12) lowpower\_mode has been set to '0'. Although write access to registers is supported at the full interface clock speed (SCL or SCK), a waiting period must be inserted between two consecutive write cycles (please refer also to section 7.2.1).

In **standby mode** the analog part is powered down, while the digital part remains largely operational. No data acquisition is performed. Reading and writing registers is supported without any restrictions. The latest acceleration data and the content of all configuration registers are kept. Standby mode is entered (left) by writing '1' ('0') to the (0x11) suspend bit after bit (0x12) lowpower\_mode has been set to '1'. It is also possible to enter normal mode by performing a softreset as described in chapter 4.8.

In **low-power mode 1**, the device is periodically switching between a sleep phase and a wake-up phase. The wake-up phase essentially corresponds to operation in normal mode with complete power-up of the circuitry. The sleep phase essentially corresponds to operation in suspend mode. Low-power mode is entered (left) by writing '1' ('0') to the (0x11) lowpower\_en bit with bit (0x12) lowpower\_mode set to '0'. Read access to registers is possible except from the (0x3F) fifo\_data register. However, unless the register access is synchronised with the wake-up phase, the restrictions of the suspend mode apply.

**Low-power mode 2** is very similar to low-power mode 1, but register access is possible at any time without restrictions. It consumes more power than low-power mode 1. In low-power mode 2 the device is periodically switching between a sleep phase and a wake-up phase. The wake-up phase essentially corresponds to operation in normal mode with complete power-up of the circuitry. The sleep phase essentially corresponds to operation in standby mode. Low-power mode is entered (left) by writing '1' ('0') to the (0x11) lowpower\_en bit with bit (0x12) lowpower\_mode set to '1'.

The **timing behaviour** of the low-power modes 1 and 2 depends on the setting of the (0x12) sleeptimer\_mode bit. When (0x12) sleeptimer\_mode is set to '0', the event-driven time-base mode (EDT) is selected. In EDT the duration of the wake-up phase depends on the number of samples required by the enabled interrupt engines. If an interrupt is detected, the device stays in the wake-up phase as long as the interrupt condition endures (non-latched interrupt), or until the latch time expires (temporary interrupt), or until the interrupt is reset (latched interrupt). If no interrupt is detected, the device enters the sleep phase immediately after the required number of acceleration samples have been taken and an active interface access cycle has ended. The EDT mode is recommended for power-critical applications which do not use the FIFO. Also, EDT mode is compatible with legacy BST sensors. Figure 3 shows the timing diagram for low-power modes 1 and 2 when EDT is selected.



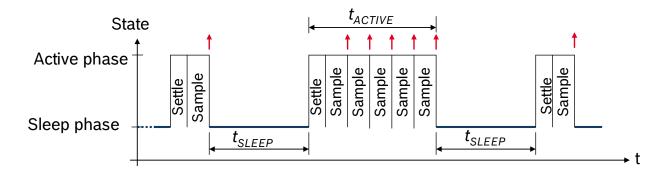


Figure 3: Timing Diagram for low-power mode 1/2, EDT

When (0x12) sleeptimer\_mode is set to '1', the equidistant-sampling mode (EST) is selected. The use of the EST mode is recommended when the FIFO is used since it ensures that equidistant samples are sampled into the FIFO regardless of whether the active phase is extended by active interrupt engines or interface activity. In EST mode the sleep time  $t_{SLEEP}$  is defined as shown in Figure 4. The FIFO sampling time  $t_{SAMPLE}$  is the sum of the sleep time  $t_{SLEEP}$  and the sensor data sampling time  $t_{SSMP}$ . Since interrupt engines can extend the active phase to exceed the sleep time  $t_{SLEEP}$ , equidistant sampling is only guaranteed if the bandwidth has been chosen such that  $1/(2 * bw) = n * t_{SLEEP}$  where n is an integer. If this condition is infringed, equidistant sampling is not possible. Once the sleep time has elapsed the device will store the next available sample in the FIFO. This set-up condition is not recommended as it may result in timing jitter.

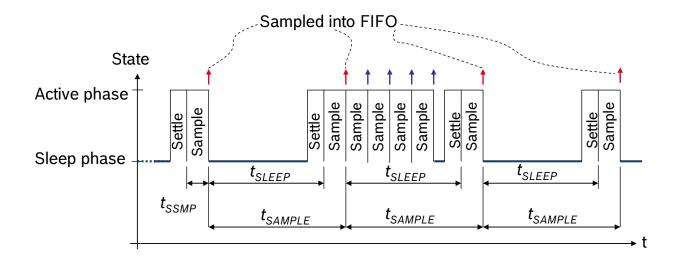


Figure 4: Timing Diagram for low-power mode 1/2, EST

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The sleep time for lower-power mode 1 and 2 is set by the (0x11) sleep\_dur bits as shown in the following table:

Table 3: Sleep phase duration settings

(0x11) sleep_dur	Sleep Phase Duration t <sub>sleep</sub>
0000b	0.5ms
0001b	0.5ms
0010b	0.5ms
0011b	0.5ms
0100b	0.5ms
0101b	0.5ms
0110b	1ms
0111b	2ms
1000b	4ms
1001b	6ms
1010b	10ms
1011b	25ms
1100b	50ms
1101b	100ms
1110b	500ms
1111b	1s

The current consumption of the BMA280 in low-power mode 1 ( $I_{DDlp2}$ ) and low-power mode 2 ( $I_{DDlp2}$ ) can be estimated with the following formulae:

$$I_{DDlpl} \approx \frac{t_{sleep} \cdot I_{DDsum} + t_{active} \cdot I_{DD}}{t_{sleep} + t_{active}} \,.$$

$$I_{DDlp2} \approx \frac{t_{sleep} \cdot I_{DDsbm} + t_{active} \cdot I_{DD}}{t_{sleep} + t_{active}}$$

When estimating the length of the wake-up phase  $t_{active}$ , the corresponding typical wake-up time,  $t_{w,up1}$  or  $t_{w,up2}$  and  $t_{ut}$  (given in Table 4) have to be considered:

If bandwidth is >=31.25 Hz:  

$$t_{active} = t_{ut} + t_{w,up1} - 0.9 \text{ ms (or } t_{active} = t_{ut} + t_{w,up2} - 0.9 \text{ ms)}$$
  
else:  
 $t_{active} = 4 t_{ut} + t_{w,up1} - 0.9 \text{ ms (or } t_{active} = 4 t_{ut} + t_{w,up2} - 0.9 \text{ ms)}$ 

During the wake-up phase all analog modules are held powered-up, while during the sleep phase most analog modules are powered down. Consequently, a wake-up time of at least  $t_{w,up1}$  ( $t_{w,up2}$ ) is needed to settle the analog modules so that reliable acceleration data are generated.



#### 4.3 Sensor data

#### 4.3.1 Acceleration data

The width of acceleration data is 14 bits given in two's complement representation. The 14 bits for each axis are split into an MSB upper part (one byte containing bits 13 to 6) and an LSB lower part (one byte containing bits 5 to 0 of acceleration and a (0x02, 0x04, 0x06) new\_data flag). Reading the acceleration data registers shall always start with the LSB part. In order to ensure the integrity of the acceleration data, the content of an MSB register is locked by reading the corresponding LSB register (shadowing procedure). When shadowing is enabled, the MSB must always be read in order to remove the data lock. The shadowing procedure can be disabled (enabled) by writing '1' ('0') to the bit shadow\_dis. With shadowing disabled, the content of both MSB and LSB registers is updated by a new value immediately. Unused bits of the LSB registers may have any value and should be ignored. The (0x02, 0x04, 0x06) new\_data flag of each LSB register is set if the data registers have been updated. The flag is reset if either the corresponding MSB or LSB part is read.

Two different streams of acceleration data are available, unfiltered and filtered. The unfiltered data is sampled with 2kHz. The sampling rate of the filtered data depends on the selected filter bandwidth and is always twice the selected bandwidth (BW = ODR/2). Which kind of data is stored in the acceleration data registers depends on bit (0x13)  $data_high_bw$  is '0' ('1'), then filtered (unfiltered) data is stored in the registers. Both data streams are offset-compensated.

The bandwidth of filtered acceleration data is determined by setting the (0x10) bw bit as followed:

bw	Bandwidth	Update Time t <sub>ut</sub>
00xxx	*)	-
01000	7.81Hz	64ms
01001	15.63Hz	32ms
01010	31.25Hz	16ms
01011	62.5Hz	8ms
01100	125Hz	4ms
01101	250Hz	2ms
01110	500Hz	1ms
01111	unfiltered	0.5ms
1xxxx	*)	-

Table 4: Bandwidth configuration

<sup>\*)</sup> Note: Settings 00xxx result in a bandwidth of 7.81 Hz; settings 1xxxx result in  $ODR_{max}$  (unfiltered signal). It is recommended to actively set an application specific and an appropriate bandwidth and to use the range from '01000b' to '01111b' only in order to be compatible with future products.



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The BMA280 supports four different acceleration measurement ranges. A measurement range is selected by setting the (0x0F) range bits as follows:

Table 5: Range selection

Range	Acceleration measurement range	Resolution
0011	±2g	0.244mg/LSB
0101	±4g	0.488mg/LSB
1000	±8g	0.977mg/LSB
1100	±16g	1.953mg/LSB
others	reserved	-

#### 4.3.2 Temperature sensor

The width of temperature data is 8 bits given in two's complement representation. Temperature values are available in the (0x08) temp register.

The slope of the temperature sensor is 0.5 K/LSB, its center temperature is  $23^{\circ}\text{C}$  [(0x08) temp = 0x00].



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#### 4.4 Self-test

This feature permits to check the sensor functionality by applying electrostatic forces to the sensor core instead of external accelerations. By actually deflecting the seismic mass, the entire signal path of the sensor can be tested. Activating the self-test results in a static offset of the acceleration data; any external acceleration or gravitational force applied to the sensor during active self-test will be observed in the output as a superposition of both acceleration and self-test signal.

Before the self-test is enabled the g-range should be set to 4 g. The self-test is activated individually for each axis by writing the proper value to the (0x32) self\_test\_axis bits ('01b' for x-axis, '10b' for y-axis, '11b' for z-axis, '00b' to deactivate self-test). It is possible to control the direction of the deflection through bit (0x32) self\_test\_sign. The excitation occurs in negative (positive) direction if (0x32) self\_test\_sign = '0b' ('1b'). After the self-test is enabled, the user should wait 50ms before interpreting the acceleration data.

In order to ensure a proper interpretation of the self-test signal it is recommended to perform the self-test for both (positive and negative) directions and then to calculate the difference of the resulting acceleration values. Table 6 shows the minimum differences for each axis. The actually measured signal differences can be significantly larger.

Table 6: Self-test difference values

	x-axis signal	y-axis signal	z-axis signal
resulting minimum difference signal	800 mg	800 mg	400 mg

It is recommended to perform a reset of the device after a self-test has been performed. If the reset cannot be performed, the following sequence must be kept to prevent unwanted interrupt generation: disable interrupts, change parameters of interrupts, wait for at least 50ms, enable desired interrupts.



#### 4.5 Offset compensation

Offsets in measured signals can have several causes but they are always unwanted and disturbing in many cases. Therefore, the BMA280 offers an advanced set of four digital offset compensation methods which are closely matched to each other. These are slow, fast, and manual compensation as well as inline calibration.

The compensation is performed with unfiltered data, and is then applied to both, unfiltered and filtered data. If necessary the result of this computation is saturated to prevent any overflow errors (the smallest or biggest possible value is set, depending on the sign). However, the registers used to read and write compensation values have a width of 8 bits.

An overview of the offset compensation principle is given in figure 5:

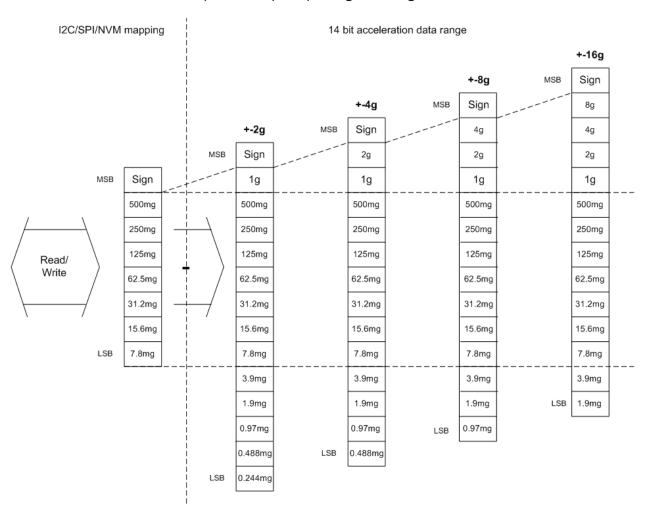


Figure 5: Principle of offset compensation



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The public offset compensation registers (0x38) offset\_x, (0x39) offset\_y, (0x3A) offset\_z are images of the corresponding registers in the NVM. With each image update (see section 4.6 Non-volatile memory for details) the contents of the NVM registers are written to the public registers. The public registers can be over-written by the user at any time. After changing the contents of the public registers by either an image update or manually, all 8bit values are extended to 14bit values for internal computation. In the opposite direction, if an internally computed value changes it is converted to an 8bit value and stored in the public register.

Depending on the selected g-range the conversion from 14bit to 8bit values can result in a loss of accuracy of one to several LSB. This is shown in figure 5.

In case an internally computed compensation value is too small or too large to fit into the corresponding register, it is saturated in order to prevent an overflow error.

By writing '1' to the (0x36) offset reset bit, all offset compensation registers are reset to zero.



#### 4.5.1 Slow compensation

Slow compensation is based on a  $1^{st}$  order high-pass filter, which continuously drives the average value of the output data stream of each axis to zero. The bandwidth of the high-pass filter is configured with bit (0x37) cut\_off according to Table 7.

Table 7: Compensation period settings

(0x37) cut_off	high-pass filter bandwidth	Example bw = 500 Hz
0b	$\frac{1Hz \times bw *}{1000 Hz}$	$\frac{1Hz \times 500  Hz}{1000  Hz} = 0.5  Hz$
1b	$\frac{10Hz \times bw *}{1000 Hz}$	$\frac{10Hz \times 500Hz}{1000Hz} = 5Hz$

<sup>\*</sup>bw: please insert selected decimal data bandwidth value [Hz] from table 4

The slow compensation can be enabled (disabled) for each axis independently by setting the bits  $(0x36) hp_xen, hp_yen, hp_zen$  to '1' ('0'), respectively.

Slow compensation should not be used in combination with low-power mode. In low-power mode the conditions (availability of necessary data) for proper function of slow compensation are not fulfilled.

#### 4.5.2 Fast compensation

Fast compensation is a one-shot process by which the compensation value is set in such a way that when added to the raw acceleration, the resulting acceleration value of each axis approaches the target value. This is best suited for "end-of-line trimming" with the customer's device positioned in a well-defined orientation. For fast compensation the g-range has to be switched to 2g.

The algorithm in detail: An average of 16 consecutive acceleration values is computed and the difference between target value and computed value is written to (0x38, 0x39, 0x34) offset\_filt\_x/y/z. The public registers (0x38, 0x39, 0x34) offset\_filt\_x/y/z are updated with the contents of the internal registers (using saturation if necessary) and can be read by the user.

Fast compensation is triggered for each axis individually by setting the (0x36) cal\_trigger bits as shown in Table 8:

Table 8: Fast compensation axis selection

(0x36) cal_trigger	Selected Axis
00b	none
01b	X
10b	у
11b	Z



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Register (0x36) cal\_trigger is a write-only register. Once triggered, the status of the fast correction process is reflected in the status bit (0x36) cal\_rdy. Bit (0x36) cal\_rdy is '0' while the correction is in progress. Otherwise it is '1'. Bit (0x36) cal\_rdy is '0' when (0x36) cal\_trigger is not '00'.

For the fast offset compensation, the compensation target can be chosen by setting the bits (0x37) offset\_target\_x, (0x37) offset\_target\_y, and (0x37) offset\_target\_z according to Table 9:

 (0x37)
 Target value

 00b
 0g

 01b
 +1g

 10b
 -1g

0g

11b

Table 9: Offset target settings

Fast compensation should not be used in combination with any of the low-power modes. In low-power mode the conditions (availability of necessary data) for proper function of fast compensation are not fulfilled.

#### 4.5.3 Manual compensation

The contents of the public compensation registers (0x38, 0x39, 0x3A) offset\_filt\_x/y/z can be set manually via the digital interface. It is recommended to write into these registers directly after a new data interrupt has occurred in order not to disturb running offset computations.

Writing to the offset compensation registers is not allowed while the fast compensation procedure is running.

#### 4.5.4 Inline calibration

For certain applications, it is often desirable to calibrate the offset once and to store the compensation values permanently. This can be achieved by using one of the aforementioned offset compensation methods to determine the proper compensation values and then storing these values permanently in the NVM. See section 4.6 Non-volatile memory for details of the storing procedure.

Each time the device is reset, the compensation values are loaded from the non-volatile memory into the image registers and used for offset compensation until they are possibly overwritten using one of the other compensation methods.



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#### 4.6 Non-volatile memory

The entire memory of the BMA280 consists of three different kinds of registers: hard-wired, volatile, and non-volatile. Part of it can be both read and written by the user. Access to non-volatile memory is only possible through (volatile) image registers.

Altogether, there are eight registers (octets) with NVM backup which are accessible by the user. The addresses of the image registers range from 0x38 to 0x3C. While the addresses up to 0x3A are used for offset compensation (see 4.4 Offset Compensation), addresses 0x3B and 0x3C are general purpose registers not linked to any sensor-specific functionality.

The content of the NVM is loaded to the image registers after a reset (either POR or softreset) or after a user request which is performed by writing '1' to the write-only bit (0x33)  $nvm_load$ . As long as the image update is in progress, bit (0x33)  $nvm_rdy$  is '0', otherwise it is '1'.

The image registers can be read and written like any other register.

Writing to the NVM is a three-step procedure:

- 1. Write the new contents to the image registers.
- 2. Write '1' to bit (0x33) nvm\_prog\_mode in order to unlock the NVM.
- 3. Write '1' to bit (0x33) nvm\_prog\_trig and keep '1' in bit (0x33) nvm\_prog\_mode in order to trigger the write process.

Writing to the NVM always renews the entire NVM contents. It is possible to check the write status by reading bit  $(0x33) \ nvm_rdy$ . While  $(0x33) \ nvm_rdy = '0'$ , the write process is still in progress; if  $(0x33) \ nvm_rdy = '1'$ , then writing is completed. As long as the write process is ongoing, no change of power mode and image registers is allowed. Also, the NVM write cycle must not be initiated while image registers are updated, in low-power mode, and in suspend mode.

Please note that the number of permitted NVM write-cycles is limited as specified in Table 1. The number of remaining write-cycles can be obtained by reading bits (0x33) nvm\_remain.

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#### 4.7 Interrupt controller

The BMA280 is equipped with eight programmable interrupt engines. Each interrupt can be independently enabled and configured. If the trigger condition of an enabled interrupt is fulfilled, the corresponding status bit is set to '1' and the selected interrupt pin is activated. The BMA280 provides two interrupt pins, INT1 and INT2; interrupts can be freely mapped to any of these pins. The state of a specific interrupt pin is derived from a logic 'or' combination of all interrupts mapped to it.

The interrupt status registers are updated when a new data word is written into the acceleration data registers. If an interrupt is disabled, all active status bits associated with it are immediately reset.

#### 4.7.1 General features

An interrupt is cleared depending on the selected interrupt mode, which is common to all interrupts. There are three different interrupt modes: non-latched, latched, and temporary. The mode is selected by the (0x21) latch\_int bits according to Table 10.

(0x21)Interrupt mode latch\_int 0000b non-latched 0001b temporary, 250ms 0010b temporary, 500ms 0011b temporary, 1s 0100b temporary, 2s 0101b temporary, 4s 0110b temporary, 8s 0111b latched 1000b non-latched 1001b temporary, 250µs 1010b temporary, 500µs 1011b temporary, 1ms 1100b temporary, 12.5ms 1101b temporary, 25ms 1110b temporary, 50ms 1111b latched

Table 10: Interrupt mode selection

An interrupt is generated if its activation condition is met. It can not be cleared as long as the activation condition is fulfilled. In the non-latched mode the interrupt status bit and the selected pin (the contribution to the 'or' condition for INT1 and/or INT2) are cleared as soon as the activation condition is no more valid. Exceptions to this behavior are the new data, orientation, and flat interrupts, which are automatically reset after a fixed time.



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In latched mode an asserted interrupt status and the selected pin are cleared by writing '1' to bit (0x21) reset\_int. If the activation condition still holds when it is cleared, the interrupt status is asserted again with the next change of the acceleration registers.

In the temporary mode an asserted interrupt and selected pin are cleared after a defined period of time. The behaviour of the different interrupt modes is shown graphically in figure 6. The timings in this mode are subject to the same tolerances as the bandwidths (see Table 1).

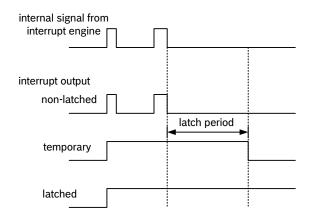


Figure 6: Interrupt modes

Several interrupt engines can use either unfiltered or filtered acceleration data as their input. For these interrupts, the source can be selected with the bits in register (0x1E). These are (0x1E)  $int\_src\_data$ , (0x1E)  $int\_src\_tap$ , (0x1E)  $int\_src\_slo\_no\_mot$ , (0x1E)  $int\_src\_slope$ , (0x1E)  $int\_src\_lope$ , (0x1E)  $int\_src\_lop$ 

It is strongly recommended to set interrupt parameters prior to enabling the interrupt. Changing parameters of an already enabled interrupt may cause unwanted interrupt generation and generation of a false interrupt history. A safe way to change parameters of an enabled interrupt is to keep the following sequence: disable the desired interrupt, change parameters, wait for at least 10ms, and then re-enable the desired interrupt.

#### 4.7.2 Mapping to physical interrupt pins (inttype to INT Pin#)

Registers (0x19) to (0x1B) are dedicated to mapping of interrupts to the interrupt pins "INT1" or "INT2". Setting (0x19) int1\_"inttype" to '1' ('0') maps (unmaps) "inttype" to pin "INT1". Correspondingly setting (0x1B) int2\_"inttype" to '1' ('0') maps (unmaps) "inttype" to pin "INT2".

Note: "inttype" to be replaced with the precise notation, given in the memory map in chapter 6.

Example: For flat interrupt (int1\_flat): Setting (0x19) int1\_flat to '1' maps int1\_flat to pin "INT1".



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#### 4.7.3 Electrical behaviour (INT pin# to open-drive or push-pull)

Both interrupt pins can be configured to show the desired electrical behaviour. The 'active' level of each interrupt pin is determined by the (0x20) int1\_lvl and (0x20) int2\_lvl bits.

If (0x20)  $int1_lvl = '1'$  ('0') / (0x20)  $int2_lvl = '1'$  ('0'), then pin "INT1" / pin "INT2" is active '1' ('0'). The characteristic of the output driver of the interrupt pins may be configured with bits (0x20)  $int1_lod$  and (0x20)  $int2_lod$ . By setting bits (0x20)  $int1_lod$  / (0x20)  $int2_lod$  to '1', the output driver shows open-drive characteristic, by setting the configuration bits to '0', the output driver shows push-pull characteristic. When open-drive characteristic is selected in the design, external pull-up or pull-down resistor should be applied according the int\_lvl configuration.

#### 4.7.4 New data interrupt

This interrupt serves for synchronous reading of acceleration data. It is generated after storing a new value of z-axis acceleration data in the data register. The interrupt is cleared automatically when the next data acquisition cycle starts. The interrupt status is '0' for at least 50µs.

The interrupt mode of the new data interrupt is fixed to non-latched.

It is enabled (disabled) by writing '1' ('0') to bit (0x17) data\_en. The interrupt status is stored in bit (0x0A) data\_int.

Due to the settling time of the filter, the first interrupt after wake-up from suspend or standby mode will take longer than the update time.



#### 4.7.5 Slope / any-motion detection

Slope / any-motion detection uses the slope between successive acceleration signals to detect changes in motion. An interrupt is generated when the slope (absolute value of acceleration difference) exceeds a preset threshold. It is cleared as soon as the slope falls below the threshold. The principle is made clear in figure 7.

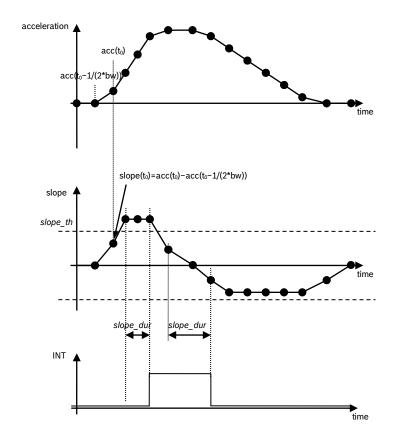


Figure 7: Principle of any-motion detection

The threshold is defined through register (0x28) slope\_th. In terms of scaling 1 LSB of (0x28) slope\_th corresponds to 3.91 mg in 2g-range (7.81 mg in 4g-range, 15.6 mg in 8g-range and 31.3 mg in 16g-range). Therefore the maximum value is 996 mg in 2g-range (1.99g in 4g-range, 3.98g in 8g-range and 7.97g in 16g-range).

The time difference between the successive acceleration signals depends on the selected bandwidth and equates to 1/(2\*bandwidth) ()t=1/(2\*bw)). In order to suppress false triggers, the interrupt is only generated (cleared) if a certain number N of consecutive slope data points is larger (smaller) than the slope threshold given by (0x28) slope\_th. This number is set by the (0x27) slope dur bits. It is N = (0x27) slope dur + 1 for (0x27).

Example: (0x27) slope dur = 00b, ..., 11b = 1decimal, ..., 4decimal.



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#### 4.7.5.1 Enabling (disabling) for each axis

Any-motion detection can be enabled (disabled) for each axis separately by writing '1' ('0') to bits (0x16)  $slope\_en\_x$ , (0x16)  $slope\_en\_y$ , (0x16)  $slope\_en\_z$ . The criteria for any-motion detection are fulfilled and the slope interrupt is generated if the slope of any of the enabled axes exceeds the threshold (0x28)  $slope\_th$  for [(0x27)  $slope\_dur +1]$  consecutive times. As soon as the slopes of all enabled axes fall or stay below this threshold for [(0x27)  $slope\_dur +1]$  consecutive times the interrupt is cleared unless interrupt signal is latched.

#### 4.7.5.2 Axis and sign information of slope / any motion interrupt

The interrupt status is stored in bit (0x09)  $slope_int$ . The any-motion interrupt supplies additional information about the detected slope. The axis which triggered the interrupt is given by that one of bits (0x0B)  $slope_first_x$ , (0x0B)  $slope_first_y$ , (0x0B)  $slope_first_z$  that contains a value of '1'. The sign of the triggering slope is held in bit (0x0B)  $slope_sign$  until the interrupt is retriggered. If (0x0B)  $slope_sign = '0'$  ('1'), the sign is positive (negative).



#### 4.7.6 Tap sensing

Tap sensing has a functional similarity with a common laptop touch-pad or clicking keys of a computer mouse. A tap event is detected if a pre-defined slope of the acceleration of at least one axis is exceeded. Two different tap events are distinguished: A 'single tap' is a single event within a certain time, followed by a certain quiet time. A 'double tap' consists of a first such event followed by a second event within a defined time frame.

Single tap interrupt is enabled (disabled) by writing '1' ('0') to bit (0x16) s\_tap\_en. Double tap interrupt is enabled (disabled) by writing '1' ('0') to bit (0x16) d\_tap\_en.

While temporary latching is used do not simultaneously enable single tap interrupt and double tap interrupt.

The status of the single tap interrupt is stored in bit (0x09) s\_tap\_int, the status of the double tap interrupt is stored in bit (0x09) d\_tap\_int.

The slope threshold for detecting a tap event is set by bits (0x2B) tap\_th. The meaning of (0x2B) tap\_th depends on the range setting. 1 LSB of (0x2B) tap\_th corresponds to a slope of 62.5mg in 2g-range, 125mg in 4g-range, 250mg in 8g-range, and 500mg in 16g-range.

In figure 8 the meaning of the different timing parameters is visualized:

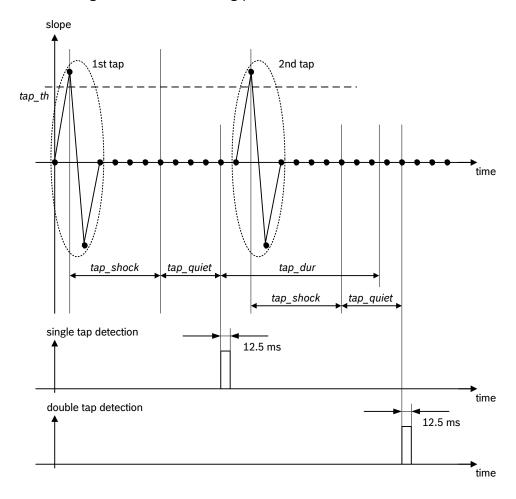


Figure 8: Timing of tap detection

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The parameters (0x2A)  $tap\_shock$  and (0x2A)  $tap\_quiet$  apply to both single tap and double tap detection, while (0x2A)  $tap\_dur$  applies to double tap detection only. Within the duration of (0x2A)  $tap\_shock$  any slope exceeding (0x2B)  $tap\_th$  after the first event is ignored. Contrary to this, within the duration of (0x2A)  $tap\_quiet$  no slope exceeding (0x2B)  $tap\_th$  must occur, otherwise the first event will be cancelled.

#### 4.7.6.1 Single tap detection

A single tap is detected and the single tap interrupt is generated after the combined durations of (0x2A)  $tap\_shock$  and (0x2A)  $tap\_quiet$ , if the corresponding slope conditions are fulfilled. The interrupt is cleared after a delay of 12.5 ms.

Do not map single-tap to any INT pin if you do not want to use it.

#### 4.7.6.2 Double tap detection

A double tap interrupt is generated if an event fulfilling the conditions for a single tap occurs within the set duration in (0x2A)  $tap\_dur$  after the completion of the first tap event. The interrupt is automatically cleared after a delay of 12.5 ms.

#### 4.7.6.3 Selecting the timing of tap detection

For each of parameters (0x2A)  $tap\_shock$  and (0x2A)  $tap\_quiet$  two values are selectable. By writing '0' ('1') to bit (0x2A)  $tap\_shock$  the duration of (0x2A)  $tap\_shock$  is set to 50 ms (75 ms). By writing '0' ('1') to bit (0x2A)  $tap\_quiet$  the duration of (0x2A)  $tap\_quiet$  is set to 30 ms (20 ms).

The length of (0x2A)  $tap\_dur$  can be selected by setting the (0x2A)  $tap\_dur$  bits according to Table 11:

(0x2A)length of tap\_dur tap\_dur 000b 50 ms 001b 100 ms 010b 150 ms 011b 200 ms 100b 250 ms 101b 375 ms 110b 500 ms 111b 700 ms

Table 11: Selection of tap dur



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#### 4.7.6.4 Axis and sign information of tap sensing

The sign of the slope of the first tap which triggered the interrupt is stored in bit (0x0B) tap\_sign ('0' means positive sign, '1' means negative sign). The value of this bit persists after clearing the interrupt.

The axis which triggered the interrupt is indicated by bits (0x0B) tap\_first\_x, (0x0B) tap\_first\_y, and (0x0B) tap\_first\_z.

The bit corresponding to the triggering axis contains a '1' while the other bits hold a '0'. These bits are cleared together with clearing the interrupt status.

#### 4.7.6.5 Tap sensing in low power mode

In low-power mode, a limited number of samples is processed after wake-up to decide whether an interrupt condition is fulfilled. The number of samples is selected by bits (0x2B) tap\_samp according to Table 12.

Table 12: Meaning of (0x2B) tap\_samp

(0x2B) tap_samp	Number of Samples
00b	2
01b	4
10b	8
11b	16



#### 4.7.7 Orientation recognition

The orientation recognition feature informs on an orientation change of the sensor with respect to the gravitational field vector 'g'. The measured acceleration vector components with respect to the gravitational field are defined as shown in figure 9.

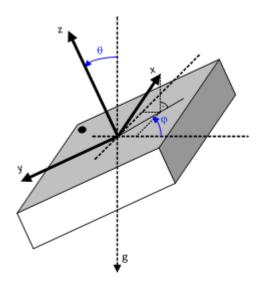


Figure 9: Definition of vector components

Therefore, the magnitudes of the acceleration vectors are calculated as follows:

$$acc_x = 1g x sin\theta x cos\phi$$
  
 $acc_y = -1g x sin\theta x sin\phi$   
 $acc_z = 1g x cos\theta$   
 $acc_y/acc_x = -tan\phi$ 

Depending on the magnitudes of the acceleration vectors the orientation of the device in the space is determined and stored in the three (0x0C) orient bits. These bits may not be reset in the sleep phase of low-power mode. There are three orientation calculation modes with different thresholds for switching between different orientations: symmetrical, high-asymmetrical, and low-asymmetrical. The mode is selected by setting the (0x2C) orient\_mode bits as given in Table 13.

Table 13: Orientation mode settings

(0x2C) orient_mode	Orientation Mode
00b	symmetrical
01b	high-asymmetrical
10b	low-asymmetrical
11b	symmetrical

For each orientation mode the (0x0C) orient bits have a different meaning as shown in Table 14 to Table 16:

Table 14: Meaning of the (0x0C) orient bits in symmetrical mode

(0x0C) orient	Name	Angle	Condition
x00	portrait upright	315° < φ < 45°	acc_y  <  acc_x  - 'hyst' and acc_x - 'hyst' ≥ 0
x01	portrait upside down	135° < φ < 225°	acc_y  <  acc_x  - 'hyst' and acc_x + 'hyst' < 0
x10	landscape left	45° < φ < 135°	$ acc_y  \ge  acc_x  + 'hyst'$ and $acc_y < 0$
x11	landscape right	225° < φ < 315°	$ acc_y  \ge  acc_x  + 'hyst'$ and $acc_y \ge 0$

Table 15: Meaning of the (0x0C) orient bits in high-asymmetrical mode

(0x0C) orient	Name	Angle	Condition
x00	portrait upright	297° < φ < 63°	acc_y  < 2 ·  acc_x  - 'hyst' and acc_x - 'hyst' ≥ 0
x01	portrait upside down	117° < φ < 243°	acc_y  < 2 ·  acc_x  - 'hyst' and acc_x + 'hyst' < 0
x10	landscape left	63° < φ < 117°	$ acc_y  \ge 2 \cdot  acc_x  + 'hyst'$ and $acc_y < 0$
x11	landscape right	243° < φ < 297°	$ acc_y  \ge 2 \cdot  acc_x  + \text{'hyst'}$ and $acc_y \ge 0$

Table 16: Meaning of the (0x0C) orient bits in low-asymmetrical mode

(0x0C) orient	Name	Angle	Condition
x00	portrait upright	333° < φ < 27°	$ acc_y  < 0.5 \cdot  acc_x  - 'hyst'$ and $acc_x - 'hyst' \ge 0$
x01	portrait upside down	153° < φ < 207°	$ acc_y  < 0.5 \cdot  acc_x  - 'hyst'$ and $acc_x + 'hyst' < 0$
x10	landscape left	27° < φ < 153°	$ acc_y  \ge 0.5 \cdot  acc_x  + 'hyst'$ and $acc_y < 0$
x11	landscape right	207° < φ < 333°	$ acc_y  \ge 0.5 \cdot  acc_x  + \text{ 'hyst'}$ and $acc_y \ge 0$

In the preceding tables, the parameter 'hyst' stands for a hysteresis, which can be selected by setting the (0x2C) orient\_hyst bits. 1 LSB of (0x2C) orient\_hyst always corresponds to 62.5 mg, in any g-range (i.e. increment is independent from g-range setting). It is important to note that by using a hysteresis  $\neq$  0 the actual switching angles become different from the angles given in the tables since there is an overlap between the different orientations.

The most significant bit of the *(0x0C)* orient bits (which is displayed as an 'x' in the above given tables) contains information about the direction of the z-axis. It is set to '0' ('1') if  $acc_z \ge 0$  ( $acc_z < 0$ ).



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Figure 10 shows the typical switching conditions between the four different orientations for the symmetrical mode i.e. without hysteresis:

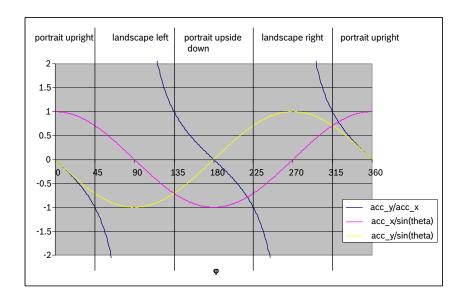


Figure 10: Typical orientation switching conditions w/o hysteresis

The orientation interrupt is enabled (disabled) by writing '1' ('0') to bit (0x16) orient\_en. The interrupt is generated if the value of (0x0C) orient has changed. It is automatically cleared after one stable period of the (0x0C) orient value. The interrupt status is stored in the (0x09) orient\_int bit. The register (0x0C) orient always reflects the current orientation of the device, irrespective of which interrupt mode has been selected. Bit (0x0C) orient<2> reflects the device orientation with respect to the z-axis. The bits (0x0C) orient<1:0> reflect the device orientation in the x-y-plane. The conventions associated with register (0x0C) orient are detailed in chapter 6.

#### 4.7.7.1 Orientation blocking

The change of the (0x0C) orient value and – as a consequence – the generation of the interrupt can be blocked according to conditions selected by setting the value of the (0x2C) orient\_blocking bits as described by Table 17.

(0x2C) **Conditions** orient\_blocking 00b no blocking theta blocking 01b or acceleration in any axis > 1.5g theta blocking 10b acceleration slope in any axis > 0.2 g acceleration in any axis > 1.5g theta blocking acceleration slope in any axis > 0.4 g 11b acceleration in any axis > 1.5g and value of orient is not stable for at least 100 ms

Table 17: Blocking conditions for orientation recognition

The theta blocking is defined by the following inequality:

$$|\tan \theta| < \frac{\sqrt{blocking\_theta}}{8}$$

The parameter *blocking\_theta* of the above given equation stands for the contents of the (0x2D) orient\_theta bits. It is possible to define a blocking angle between 0° and 44.8°. The internal blocking algorithm saturates the acceleration values before further processing. As a consequence, the blocking angles are strictly valid only for a device at rest; they can be different if the device is moved.

#### Example:

To get a maximum blocking angle of  $19^{\circ}$  the parameter *blocking\_theta* is determined in the following way:  $(8 * tan(19^{\circ}))^2 = 7.588$ , therefore, *blocking\_value* = 8 dec = 001000b has to be chosen.

In order to avoid unwanted generation of the orientation interrupt in a nearly flat position ( $z \sim 0$ , sign change due to small movements or noise), a hysteresis of 0.2 g is implemented for the z-axis, i. e. a after a sign change the interrupt is only generated after |z| > 0.2 g.

#### 4.7.7.2 Up-Down Interrupt Suppression Flag

Per default an orientation interrupt is triggered when any of the bits in register (0x0C) orient changes state. The BMA280 can be configured to trigger orientation interrupts only when the



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device position changes in the x-y-plane while orientation changes with respect to the z-axis are ignored. A change of the orientation of the z-axis, and hence a state change of bit (0x0C) orient<2> is ignored (considered) when bit (0x2D) orient\_ud\_en is set to '0' ('1').

#### 4.7.8 Flat detection

The flat detection feature gives information about the orientation of the devices' z-axis relative to the g-vector, i. e. it recognizes whether the device is in a flat position or not.

The flat angle  $\Theta$  is adjustable by (0x2E) flat\_theta from 0° to 44.8°. The flat angle can be set according to following formula:

$$\Theta = \operatorname{atan}\left(\frac{1}{8}\sqrt{\operatorname{flat\_theta}}\right)$$

A hysteresis of the flat detection can be enabled by (0x2F) flat\_hy bits. In this case the flat position is set if the angle drops below following threshold:

$$\Theta_{hyst,ll} = \operatorname{atan}\left(\frac{1}{8}\sqrt{\operatorname{flat\_theta} \cdot \left(1 - \frac{flat\_hy}{1024}\right) - \frac{flat\_hy}{16}}\right)$$

The flat position is reset if the angle exceeds the following threshold:

$$\Theta_{hyst,ul} = \operatorname{atan}\left(\frac{1}{8}\sqrt{\operatorname{flat\_theta} \cdot \left(1 + \frac{flat\_hy}{1024}\right) + \frac{flat\_hy}{16}}\right)$$

The flat interrupt is enabled (disabled) by writing '1' ('0') to bit (0x16) flat\_en. The flat value is stored in the (0x0C) flat bit if the interrupt is enabled. This value is '1' if the device is in the flat position, it is '0' otherwise. The flat interrupt is generated if the flat value has changed and the new value is stable for at least the time given by the (0x2F) flat\_hold\_time bits. A flat interrupt may be also generated if the flat interrupt is enabled. The actual status of the interrupt is stored in the (0x09) flat\_int bit. The flat orientation of the sensor can always be determined from reading the (0x0C) flat bit after interrupt generation. If unlatched interrupt mode is used, the (0x09) flat\_int value and hence the interrupt is automatically cleared after one sample period. If temporary or latched interrupt mode is used, the (0x09) flat\_int value is kept fixed until the latch time expires or the interrupt is reset.

The meaning of the (0x2F) flat\_hold\_time bits can be seen from Table 18.

Table 18: Meaning of flat\_hold\_time

(0x2F) flat_hold_time	Time
00b	0
01b	512 ms
10b	1024 ms
11b	2048 ms



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#### 4.7.9 Low-g interrupt

This interrupt is based on the comparison of acceleration data against a low-g threshold, which is most useful for free-fall detection.

The interrupt is enabled (disabled) by writing '1' ('0') to the (0x17) low\_en bit. There are two modes available, 'single' mode and 'sum' mode. In 'single' mode, the acceleration of each axis is compared with the threshold; in 'sum' mode, the sum of absolute values of all accelerations  $|acc_x| + |acc_y| + |acc_z|$  is compared with the threshold. The mode is selected by the contents of the (0x24) low\_mode bit: '0' means 'single' mode, '1' means 'sum' mode.

The low-g threshold is set through the (0x23) low\_th register. 1 LSB of (0x23) low\_th always corresponds to an acceleration of 7.81 mg (i.e. increment is independent from g-range setting).

A hysteresis can be selected by setting the (0x24) low\_hy bits. 1 LSB of (0x24) low\_hy always corresponds to an acceleration difference of 125 mg in any g-range (as well, increment is independent from g-range setting).

The low-g interrupt is generated if the absolute values of the acceleration of all axes ('and' relation, in case of single mode) or their sum (in case of sum mode) are lower than the threshold for at least the time defined by the  $(0x22) low_dur$  register. The interrupt is reset if the absolute value of the acceleration of at least one axis ('or' relation, in case of single mode) or the sum of absolute values (in case of sum mode) is higher than the threshold plus the hysteresis for at least one data acquisition. In bit  $(0x09) low_int$  the interrupt status is stored.

The relation between the content of (0x22)  $low_dur$  and the actual delay of the interrupt generation is: delay [ms] = [(0x22)  $low_dur + 1] \cdot 2$  ms. Therefore, possible delay times range from 2 ms to 512 ms.



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#### 4.7.10 High-g interrupt

This interrupt is based on the comparison of acceleration data against a high-g threshold for the detection of shock or other high-acceleration events.

The high-g interrupt is enabled (disabled) per axis by writing '1' ('0') to bits (0x17) high\_en\_x, (0x17) high\_en\_y, and (0x17) high\_en\_z, respectively. The high-g threshold is set through the (0x26) high\_th register. The meaning of an LSB of (0x26) high\_th depends on the selected grange: it corresponds to 7.81 mg in 2g-range, 15.63 mg in 4g-range, 31.25 mg in 8g-range, and 62.5 mg in 16g-range (i.e. increment depends from g-range setting).

A hysteresis can be selected by setting the (0x24) high\_hy bits. Analogously to (0x26) high\_th, the meaning of an LSB of (0x24) high\_hy is g-range dependent: It corresponds to an acceleration difference of 125 mg in 2g-range, 250 mg in 4g-range, 500 mg in 8g-range, and 1000mg in 16g-range (as well, increment depends from g-range setting).

The high-g interrupt is generated if the absolute value of the acceleration of at least one of the enabled axes ('or' relation) is higher than the threshold for at least the time defined by the (0x25) high\_dur register. The interrupt is reset if the absolute value of the acceleration of all enabled axes ('and' relation) is lower than the threshold minus the hysteresis for at least the time defined by the (0x25) high\_dur register. In bit (0x09) high\_int the interrupt status is stored. The relation between the content of (0x25) high\_dur and the actual delay of the interrupt generation is delay [ms] = [(0x22) low\_dur + 1] • 2 ms. Therefore, possible delay times range from 2 ms to 512 ms. The interrupt will be cleared immediately once acceleration is lower than threshold.

#### 4.7.10.1 Axis and sign information of high-g interrupt

The axis which triggered the interrupt is indicated by bits (0x0C) high\_first\_x, (0x0C) high\_first\_y, and (0x0C) high\_first\_z. The bit corresponding to the triggering axis contains a '1' while the other bits hold a '0'. These bits are cleared together with clearing the interrupt status. The sign of the triggering acceleration is stored in bit (0x0C) high\_sign. If (0x0C) high\_sign = '0' ('1'), the sign is positive (negative).

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#### 4.7.11 No-motion / slow motion detection

The slow-motion/no-motion interrupt engine can be configured in two modes.

In slow-motion mode an interrupt is triggered when the measured slope of at least one enabled axis exceeds the programmable slope threshold for a programmable number of samples. Hence the engine behaves similar to the any-motion interrupt, but with a different set of parameters. In order to suppress false triggers, the interrupt is only generated (cleared) if a certain number N of consecutive slope data points is larger (smaller) than the slope threshold given by (0x27)  $slo_no_mot_dur<1:0>$  . The number is N = (0x27)  $slo_no_mot_dur<1:0>$  + 1.

In no-motion mode an interrupt is generated if the slope on all selected axes remains smaller than a programmable threshold for a programmable delay time. Figure 11 shows the timing diagram for the no-motion interrupt. The scaling of the threshold value is identical to that of the slow-motion interrupt. However, in no-motion mode register (0x27) slo\_no\_mot\_dur defines the delay time before the no-motion interrupt is triggered. Table 19 lists the delay times adjustable with register (0x27) slo\_no\_mot\_dur. The timer tick period is 1 second. Hence using short delay times can result in considerable timing uncertainty.

If bit (0x18) slo\_no\_mot\_sel is set to '1' ('0') the no-motion/slow-motion interrupt engine is configured in the no-motion (slow-motion) mode. Common to both modes, the engine monitors the slopes of the axes that have been enabled with bits (0x18) slo\_no\_mot\_en\_x, (0x18) slo\_no\_mot\_en\_y, and (0x18) slo\_no\_mot\_en\_z for the x-axis, y-axis and z-axis, respectively. The measured slope values are continuously compared against the threshold value defined in register (0x29) slo\_no\_mot\_th. The scaling is such that 1 LSB of (0x29) slo\_no\_mot\_th corresponds to 3.91 mg in 2g-range (7.81 mg in 4g-range, 15.6 mg in 8g-range and 31.3 mg in 16g-range). Therefore the maximum value is 996 mg in 2g-range (1.99g in 4g-range, 3.98g in 8g-range and 7.97g in 16g-range). The time difference between the successive acceleration samples depends on the selected bandwidth and equates to 1/(2 \* bw).

Table 19: No-motion time-out periods

(0x27) slo_no_mot_dur	Delay time	(0x27) slo_no_mot_dur	Delay time	(0x27) slo_no_mot_dur	Delay Time
0	1 s	16	40 s	32	88 s
1	2 s	17	48 s	33	96 s
2	3 s	18	56 s	34	104 s
	•••	19	64 s.		
14	15 s	20	72 s	62	328 s
15	16 s	21	80 s	63	336 s

Note: slo\_no\_mot\_dur values 22 to 31 are not specified

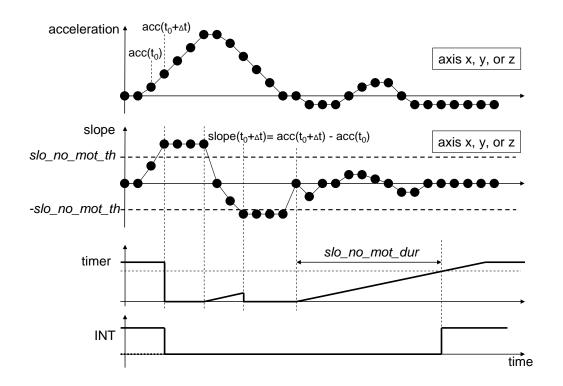


Figure 11: Timing of No-motion interrupt



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## 4.8 Softreset

A softreset causes all user configuration settings to be overwritten with their default value and the sensor to enter normal mode.

A softreset is initiated by means of writing value 0xB6 to register (0x14) softreset. Subsequently a waiting time of  $t_{w,up1}$  (max.) is required prior to accessing any configuration registers.



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# 5. FIFO Operation

## **5.1 FIFO Operating Modes**

The BMA280 features an integrated FIFO memory capable of storing up to 32 frames. Conceptually each frame consists of three 16 bit words corresponding to the x, y and z- axis, which are sampled at the same point in time. At the core of the FIFO is a buffer memory, which can be configured to operate in the following modes:

- FIFO Mode: In FIFO mode the acceleration data of the selected axes are stored in the
  buffer memory. If enabled, a watermark interrupt is triggered when the buffer has filled
  up to a configurable level. The buffer will be continuously filled until the fill level reaches
  32 frames. When it is full the data collection is stopped, and all additional samples are
  ignored. Once the buffer is full, a FIFO-full interrupt is generated if it has been enabled.
- STREAM Mode: In STREAM mode the acceleration data of the selected axes are stored in the buffer until it is full. The buffer has a depth of 31 frames. When the buffer is full the data collection continues and oldest entry is discarded. If enabled, a watermark interrupt is triggered when the buffer is filled to a configurable level. Once the buffer is full, a FIFO-full interrupt is generated if it has been enabled.
- BYPASS Mode: In bypass mode, only the current sensor data can be read out from the FIFO address. Essentially, the FIFO behaves like the STREAM mode with a depth of 1. Compared to reading the data from the normal data registers, the advantage to the user is that the packages X, Y, Z are from the same timestamp, while the data registers are updated sequentially and hence mixing of data from different axes can occur.

The primary FIFO operating mode is selected with register (0x3E) fifo\_mode according to '00b' for BYPASS mode, '01b' for FIFO mode, and '10b' for STREAM mode. Writing to register (0x3E) clears the buffer content and resets the FIFO-full and watermark interrupts. When reading register (0x3E) fifo\_mode always contains the current operating mode.

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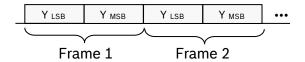
#### 5.2 FIFO Data Readout

The FIFO stores the data that are also available at the acceleration read-out registers (0x02) to (0x07). Thus, all configuration settings apply to the FIFO data as well as the acceleration data readout registers. The FIFO read out is possible through register (0x3F). The readout can be performed using burst mode since the read address counter is no longer incremented, when it has reached address (0x3F). This implies that the trapping also occurs when the burst read access starts below address (0x3F). A single burst can read out one or more frames at a time. Register (0x3E) fifo\_data\_select controls the acceleration data of which axes are stored in the FIFO. Possible settings for register (0x3E) fifo\_data\_select are '00b' for x, y- and z-axis, '01b' for x-axis only, '10b' for y-axis, '11b' for z-axis only. The depth of the FIFO is independent of whether all or a single axis have been selected. Writing to register (0x3E) clears the buffer content and resets the FIFO-full and watermark interrupts.

If all axes are enabled, the format of the data read-out from register (0x3F) is as follows:



If only one axis is enabled, the format of the data read-out from register (0x3F) is as follows (example shown: y-axis only, other axes are equivalent).



If a frame is not completely read due to an incomplete read operation, the remaining part of the frame is discarded. In this case the FIFO aligns to the next frame during the next read operation. In order for the discarding mechanism to operate correctly, there must be a delay of at least 1.5 us between the last data bit of the partially read frame and the first address bit of the next FIFO read access. Otherwise frames must not be read out partially.

If the FIFO is read beyond the FIFO fill level zeroes (0) will be read. If the FIFO is read beyond the FIFO fill level the read or burst read access time must not exceed the sampling time  $t_{SAMPLE}$ . Otherwise frames may be lost.

# 5.3 FIFO Frame Counter and Overrun Flag

Register (0x0E) fifo\_frame\_counter reflects the current fill level of the buffer. If additional frames are written to the buffer although the FIFO is full, the (0x0E) fifo\_overrun bit is set to '1'. The FIFO buffer is cleared, the FIFO fill level indicated in register (0x0E) fifo\_frame\_counter and the (0x0E) fifo\_overrun bit are both set to '0' each time one a write access to one of the FIFO configuration registers (0x3E) or (0x30) occurs. The (0x0E) fifo\_overrun bit is not reset when the FIFO fill level (0x0E) fifo\_frame\_counter has decremented to '0' due to reading from register (0x3F).



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#### 5.4 FIFO Interrupts

The FIFO controller can generate two different interrupt events, a FIFO-full and a watermark event. The FIFO-full and watermark interrupts are functional in all FIFO operating modes. The watermark interrupt is asserted when the fill level in the buffer has reached the frame count defined by register (0x30) fifo\_water\_mark\_trigger\_retain. In order to enable (disable) the watermark interrupt, the (0x17) int\_fwm\_en bit must be set to '1' ('0'). To map the watermark interrupt signal to INT1 pin (INT2 pin), (0x1A) int1\_fwm ((0x1A) int2\_fwm) bit must be set to '1'. The status of the watermark interrupt may be read back through the (0x0A) fifo\_wm\_int bit. Writing to register (0x30) fifo\_water\_mark\_trigger\_retain clears the FIFO buffer.

The FIFO-full interrupt is triggered when the buffer has been completely filled. In FIFO mode this occurs 32, in STREAM mode 31 samples, and in BYPASS mode 1 sample after the buffer has been cleared. In order to enable the FIFO-full interrupt, bit (0x17) int\_ffull\_en as well as one or both of bits (0x1A) int1\_fful or (0x1A) int2\_fful must also be set to '1'. The status of the FIFO-full interrupt may be read back through bit (0x0A) fifo full int.



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# 6. Register description

#### 6.1 General remarks

The entire communication with the device is performed by reading from and writing to registers. Registers have a width of 8 bits; they are mapped to a common space of 64 addresses from (0x00) up to (0x3F). Within the used range there are several registers which are either completely or partially marked as 'reserved'. Any reserved bit is ignored when it is written and no specific value is guaranteed when read. It is recommended not to use registers at all which are completely marked as 'reserved'. Furthermore it is recommended to mask out (logical and with zero) reserved bits of registers which are partially marked as reserved.

Registers with addresses from (0x00) up to (0x0E) are read-only. Any attempt to write to these registers is ignored. There are bits within some registers that trigger internal sequences. These bits are configured for write-only access, e. g. (0x21) reset\_int or the entire (0x14) softreset register, and read as value 0.



# 6.2 Register map

Register Address	bit7	bit6	bit5	bit4	bit3	bit2	bit1	bit0	Access	Default
0x3F				fifo_data_outpu	t_register<7:0>				ro	0x00
0x3E	fito_mo	de<1:0>					fifo_data_s	select<1:0>	w/r	0x00
0x3D 0x3C				CD4	<7:0>				w/r w/r	0xFF 0x00
0x3C 0x3B					<7:0> <7:0>				w/r w/r	0x00
0x3A					z<7:0>				w/r	0x00
0x39					y<7:0>				w/r	0x00
0x38					x<7:0>				w/r	0x00
0x37		offset_targ	et_z<1:0>		get_y<1:0>	offset_targ	et_x<1:0>	cut_off	w/r	0x00
0x36	offset_reset	cal_trigg	ger<1:0>	cal_rdy		hp_z_en	hp_y_en	hp_x_en	w/r	0x10
0x35									w/r	0x00
0x34						i2c_wdt_en	i2c_wdt_sel	spi3	w/r	0x00
0x33	nvm_remain<3:0> nvm_load nvm_rdy nvm_prog_trig nvm_prog_mode						w/r	0xF0		
0x32						self_test_sign	self_test_	axis<1:0>	w/r	0x00
0x31 0x30					fifo_water_mark_leve	l trigger retain-5:0>			w/r w/r	0xFF 0x00
0x2F			flat_hold_t	time-1:0>	IIIO_watei_mark_leve	I_IIIggeI_IetailI<5.0>	flat_hy<2:0>		w/r	0x00
0x2E			iiat_rioid_		flat the	ta<5:0>	.at_ny <2.02		w/r	0x08
0x2D		orient ud en			orient_th				w/r	0x48
0x2C			orient_hyst<2:0>		orient_bloc		orient m	ode<1:0>	w/r	0x18
0x2B	tap_sar	mp<1:0>				tap_th<4:0>			w/r	0x0A
0x2A	tap_quiet	tap_shock					tap_dur<2:0>		w/r	0x04
0x29					ot_th<7:0>				w/r	0x14
0x28					th<7:0>				w/r	0x14
0x27			slo_no_mo	t_dur<5:0>			slope_c	dur<1:0>	w/r	0x00
0x26 0x25					h<7:0>				w/r	0xC0 0x0F
0x25 0x24	high h	N =1:05		nign_a	ur<7:0>	low mode	low b	v1:0s	w/r w/r	0x0F
0x24 0x23						w/r	0x30			
0x22					ur<7:0>				w/r	0x09
0x21	reset int			1011_01	21.07	latch in	nt<3:0>		w/r	0x00
0x20					int2_od	int2_lvl	int1_od	int1_lvl	w/r	0x05
0x1F									w/r	0xFF
0x1E			int_src_data	int_src_tap	int_src_slo_no_mot	int_src_slope	int_src_high	int_src_low	w/r	0x00
0x1D									w/r	0xFF
0x1C					1				w/r	0xFF
0x1B	int2_flat	int2_orient	int2_s_tap	int2_d_tap	int2_slo_no_mot	int2_slope	int2_high	int2_low	w/r	0x00
0x1A 0x19	int2_data int1 flat	int2_fwm int1 orient	int2_ffull int1_s_tap	int1 d tap	int1 slo no mot	int1_ffull int1_slope	int1_fwm int1_high	int1_data int1_low	w/r w/r	0x00 0x00
0x19	IIILI_IIAL	IIIL1_OHerit	IIILI_S_tap	IIII I_U_IAP	slo_no_mot_sel	slo_no_mot_en_z	slo_no_mot_en_y	slo_no_mot_en_x	w/r	0x00
0x17		int_fwm_en	int_ffull_en	data_en	low_en	high_en_z	high_en_y	high_en_x	w/r	0x00
0x16	flat_en	orient_en	s_tap_en	d_tap_en		slope_en_z	slope_en_y	slope_en_x	w/r	0x00
0x15									w/r	0xFF
0x14				soft	reset				wo	0x00
0x13	data_high_bw	shadow_dis							w/r	0x00
0x12		lowpower_mode	sleeptimer_mode						w/r	0x00
0x11	suspend	lowpower_en	deep_suspend		sleep_d				w/r	0x00
0x10 0x0F						bw<4:0>	<3:0>		w/r w/r	0x0F 0x03
0x0F 0x0E	fifo overrun			fi	fo frame counter<6:0		CU.U2		ro ro	0x03
0x0D	ino_overruit			"	io_name_counter<0.0				w/r	0xFF
0x0C	flat		orient<2:0>		high sign	high first z	high_first_y	high first x	ro	0x00
0x0B	tap_sign	tap_first_z	tap_first_y	tap_first_x	slope_sign	slope_first_z	slope_first_y	slope_first_x	ro	0x00
0x0A	data_int	fifo_wm_int	fifo_full_int						ro	0x00
0x09	flat_int	orient_int	s_tap_int	d_tap_int	slo_no_mot_int	slope_int	high_int	low_int	ro	0x00
0x08					<7:0>				ro	0x00
0x07					sb<13:6>				ro	0x00
0x06			acc_z_l		-1- 40.0			new_data_z	ro	0x00
0x05					isb<13:6>			now date	ro	0x00
0x04 0x03			acc_y_l		ısb<13:6>			new_data_y	ro ro	0x00 0x00
0x03 0x02			acc v I		13.02			new data v	ro	0x00
	acc_x_lsb<5:0> new_data_x					ro				
0x01										

common w/r registers: Application specific settings which are not equal to the default settings, must be re-set to its designated values after POR, soft-reset and wake up from deep suspend.

user w/r registers: Initial default content = 0x00. Freely programmable by the user.

Remains unchanged after POR, soft-reset and wake up from deep suspend.

Figure 12: Register map



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### Register 0x00 (BGW\_CHIPID)

The register contains the chip identification code.

Name	0x00	BGW_CHIPID		
Bit	7	6	5	4
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	chip_id<7:4>			

Bit	3	2	1	0
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	chip id<3:0>			

chip id<7:0>: Fixed value b'1111'1011

#### Register 0x02 (ACCD\_X\_LSB)

The register contains the least-significant bits of the X-channel acceleration readout value. When reading out X-channel acceleration values, data consistency is guaranteed if the ACCD\_X\_LSB is read out before the ACCD\_X\_MSB and shadow\_dis='0'. In this case, after the ACCD\_X\_LSB has been read, the value in the ACCD\_X\_MSB register is locked until the ACCD\_X\_MSB has been read. This condition is inherently fulfilled if a burst-mode read access is performed. Acceleration data may be read from register ACCD\_X\_LSB at any time except during power-up and in DEEP\_SUSPEND mode.

Name	0x02	ACCD_X_LSB		
Bit	7	6	5	4
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	acc_x_lsb<5:2>			

Bit	3	2	1	0
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	acc x lsb<1:0>		undefined	new data x

acc\_x\_lsb<5:2>: Upper 4 bits of least significant 6 bits of acceleration read-back value; (two's-

complement format)

acc\_x\_lsb<1:0>: Lower 2 bits of least significant 6 bits of acceleration read-back value; (two's-

complement format)

undefined: random data; to be ignored.

new\_data\_x: ,0': acceleration value has not been updated since it has been read out last

,1': acceleration value has been updated since it has been read out last



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# Register 0x03 (ACCD\_X\_MSB)

The register contains the most-significant bits of the X-channel acceleration readout value. When reading out X-channel acceleration values, data consistency is guaranteed if the ACCD\_X\_LSB is read out before the ACCD\_X\_MSB and shadow\_dis='0'. In this case, after the ACCD\_X\_LSB has been read, the value in the ACCD\_X\_MSB register is locked until the ACCD\_X\_MSB has been read. This condition is inherently fulfilled if a burst-mode read access is performed. Acceleration data may be read from register ACCD\_X\_MSB at any time except during power-up and in DEEP\_SUSPEND mode.

Name	0x02	ACCD_X_MSB		
Bit	7	6	5	4
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	acc_x_msb<13:10	>		
				<u></u>
Bit	3	2	1	0
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	acc_x_msb<9:6>			

acc\_x\_msb<13:6>: Most significant 8 bits of acceleration read-back value (two's-complement format)



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### Register 0x04 (ACCD\_Y\_LSB)

The register contains the least-significant bits of the Y-channel acceleration readout value. When reading out Y-channel acceleration values, data consistency is guaranteed if the ACCD\_Y\_LSB is read out before the ACCD\_Y\_MSB and shadow\_dis='0'. In this case, after the ACCD\_Y\_LSB has been read, the value in the ACCD\_Y\_MSB register is locked until the ACCD\_Y\_MSB has been read. This condition is inherently fulfilled if a burst-mode read access is performed. Acceleration data may be read from register ACCD\_Y\_LSB at any time except during power-up and in DEEP\_SUSPEND mode.

0x04	ACCD_Y_LSB		
7	6	5	4
R	R	R	R
n/a	n/a	n/a	n/a
acc_y_lsb<5:2>			
	7 R n/a	7 6 R R n/a n/a	7 6 5 R R R n/a n/a n/a

Bit	3	2	1	0
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	acc_y_lsb<1:0>		undefined	new_data_y

acc\_y\_lsb<5:2>: Upper 4 bits of least significant 6 bits of acceleration read-back value; (two's-

complement format)

acc\_y\_lsb<1:0>: Lower 2 bits of least significant 6 bits of acceleration read-back value; (two's-

complement format)

undefined: random data; to be ignored

new\_data\_y: ,0': acceleration value has not been updated since it has been read out last

,1': acceleration value has been updated since it has been read out last



Value Content

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## Register 0x05 (ACCD\_Y\_MSB)

acc y msb<9:6>

The register contains the most-significant bits of the Y-channel acceleration readout value. When reading out Y-channel acceleration values, data consistency is guaranteed if the ACCD\_Y\_LSB is read out before the ACCD\_Y\_MSB and shadow\_dis='0'. In this case, after the ACCD\_Y\_LSB has been read, the value in the ACCD\_Y\_MSB register is locked until the ACCD\_Y\_MSB has been read. This condition is inherently fulfilled if a burst-mode read access is performed. Acceleration data may be read from register ACCD\_Y\_MSB at any time except during power-up and in DEEP\_SUSPEND mode.

Name	0x05	ACCD_Y_MSB		
Bit	7	6	5	4
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	acc_y_msb<13:10>	•		
Bit	3	2	1	0
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a

acc\_y\_msb<13:6>: Most significant 8 bits of acceleration read-back value (two's-complement format)



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# Register 0x06 (ACCD\_Z\_LSB)

The register contains the least-significant bits of the Z-channel acceleration readout value. When reading out Z-channel acceleration values, data consistency is guaranteed if the ACCD\_Z\_LSB is read out before the ACCD\_Z\_MSB and shadow\_dis='0'. In this case, after the ACCD\_Z\_LSB has been read, the value in the ACCD\_Z\_MSB register is locked until the ACCD\_Z\_MSB has been read. This condition is inherently fulfilled if a burst-mode read access is performed. Acceleration data may be read from register ACCD\_Z\_LSB at any time except during power-up and in DEEP\_SUSPEND mode.

Name	0x06	ACCD_Z_LSB		
Bit	7	6	5	4
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	acc_z_lsb<5:2>			

Bit	3	2	1	0
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	acc_z_lsb<1:0>		undefined	new_data_z

Acc\_z\_lsb<5:2>: Upper 4 bits of least significant 6 bits of acceleration readback value; (two's-

complement format)

Acc z lsb<1:0>: Loewer 2 bits of least significant 6 bits of acceleration read-back value;

(two's-complement format)

undefined: random data; to be ignored

new\_data\_z: ,0': acceleration value has not been updated since it has been read out last

,1': acceleration value has been updated since it has been read out last



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# Register 0x07 (ACCD\_Z\_MSB)

The register contains the most-significant bits of the Z-channel acceleration readout value. When reading out Z-channel acceleration values, data consistency is guaranteed if the ACCD\_Z\_LSB is read out before the ACCD\_Z\_MSB and shadow\_dis='0'. In this case, after the ACCD\_Z\_LSB has been read, the value in the ACCD\_Z\_MSB register is locked until the ACCD\_Z\_MSB has been read. This condition is inherently fulfilled if a burst-mode read access is performed. Acceleration data may be read from register ACCD\_Z\_MSB at any time except during power-up and in DEEP\_SUSPEND mode.

Name	0x07	ACCD_Z_MSB		
Bit	7	6	5	4
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	acc_z_msb<13:10>	>		
Bit	3	2	1	0
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	acc_z_msb<9:6>			

acc\_z\_msb<13:6>: Most significant 8 bits of acceleration read-back value (two's-complement format)



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# Register 0x08 (ACCD\_TEMP)

The register contains the current chip temperature represented in two's complement format. A readout value of temp<7:0>=0x00 corresponds to a temperature of 23°C.

Name	0x08	ACCD_TEMP			
Bit	7	6	5	4	
Read/Write	R	R	R	R	
Reset Value	n/a	n/a	n/a	n/a	
Content	temp<7:4>				
Bit	3	2	1	0	
DoodAMrito	D	-   D		D	

R	R	R
n/a	n/a	n/a
	R n/a	R R n/a

temp<7:0>: Temperature value (two s-complement format)



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## Register 0x09 (INT\_STATUS\_0)

The register contains interrupt status flags. Each flag is associated with a specific interrupt function. It is set when the associated interrupt triggers. The setting of latch\_int<3:0> controls if the interrupt signal and hence the respective interrupt flag will be permanently latched, temporarily latched or not latched. The interrupt function associated with a specific status flag must be enabled.

Name	0x09	INT_STATUS_0		
Bit	7	6	5	4
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	flat_int	orient_int	s_tap_int	d_tap_int
Bit	3	2	1	0
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	slo_no_mot_int	slope_int	high_int	low_int

flat\_int: flat interrupt status: '0'→inactive, '1' →active

orient\_int: orientation interrupt status: '0' $\rightarrow$ inactive, '1'  $\rightarrow$ active s\_tap\_int: single tap interrupt status: '0' $\rightarrow$ inactive, '1'  $\rightarrow$ active d\_tap\_int double tap interrupt status: '0' $\rightarrow$ inactive, '1'  $\rightarrow$ active slo\_not\_mot\_int: slow/no-motion interrupt status: '0' $\rightarrow$ inactive, '1'  $\rightarrow$ active

slope\_int: slope interrupt status: '0'→inactive, '1' →active high\_int: high-g interrupt status: '0'→inactive, '1' →active low\_int: low-g interrupt status: '0'→inactive, '1' →active



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# Register 0x0A (INT\_STATUS\_1)

The register contains interrupt status flags. Each flag is associated with a specific interrupt function. It is set when the associated interrupt engine triggers. The setting of latch\_int<3:0> controls if the interrupt signal and hence the respective interrupt flag will be permanently latched, temporarily latched or not latched. The interrupt function associated with a specific status flag must be enabled.

Name	0x0A	INT_STATUS_1		
Bit	7	6	5	4
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	data_int	fifo_wm_int	fifo_full_int	reserved
Bit	3	2	1	0
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	Reserved			

data\_int: data ready interrupt status: '0'→inactive, '1' →active fifo\_wm\_int: FIFO watermark interrupt status: '0'→inactive, '1' →active

fifo\_full\_int: FIFO full interrupt status: '0'→inactive, '1' →active

reserved; reserved, write to '0'



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# Register 0x0B (INT\_STATUS\_2)

The register contains interrupt status flags. Each flag is associated with a specific interrupt engine. It is set when the associated interrupt engine triggers. The setting of latch\_int<3:0> controls if the interrupt signal and hence the respective interrupt flag will be permanently latched, temporarily latched or not latched. The interrupt function associated with a specific status flag must be enabled.

Name	0x0B	INT_STATUS_2		
Bit	7	6	5	4
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	tap_sign	tap_first_z	tap_first_y	tap_first_x
	T		1	
Bit	3	2	1	0
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	slope_sign	slope_first_z	slope_first_y	slope_first_x

tap_sign:	sign of single/double tap triggering signal was '0'→positive, or '1' →negative
tap_first_z:	single/double tap interrupt: '1' → triggered by, or '0'→not triggered by z-axis
tap_first_y:	single/double tap interrupt: '1' → triggered by, or '0'→not triggered by y-axis
tap_first_x:	single/double tap interrupt: '1' → triggered by, or '0'→not triggered by x-axis
slope_sign:	slope sign of slope tap triggering signal was '0'→positive, or '1' →negative
slope_first_z:	slope interrupt: '1' → triggered by, or '0'→not triggered by z-axis
slope_first_y:	slope interrupt: '1' → triggered by, or '0'→not triggered by y-axis
slope_first_x:	slope interrupt: '1' $\rightarrow$ triggered by, or '0' $\rightarrow$ not triggered by x-axis



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# Register 0x0C (INT\_STATUS\_3)

The register contains interrupt status flags. Each flag is associated with a specific interrupt engine. It is set when the associated interrupt engine triggers. With the exception of orient<3:0> the setting of latch\_int<3:0> controls if the interrupt signal and hence the respective interrupt flag will be permanently latched, temporarily latched or not latched. The interrupt function associated with a specific status flag must be enabled.

Name	0x0C	INT_STATUS_3		
Bit	7	6	5	4
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	flat	orient<2:0>		
Bit	3	2	1	0
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	high_sign	high_first_z	high_first_y	high_first_x

flat: device is in '1'  $\rightarrow$  flat, or '0'  $\rightarrow$  non flat position;

only valid if (0x16) flat en = '1'

orient<2>: Orientation value of z-axis:  $0' \rightarrow$  upward looking, or  $1' \rightarrow$  downward

looking. The flag always reflect the current orientation status, independent of

the setting of latch int<3:0>. The flag is not updated as long as an

orientation blocking condition is active.

orient<1:0>: orientation value of x-y-plane:

'00'→portrait upright; '01'→portrait upside down; '10'→landscape left; '11'→landscape right;

The flags always reflect the current orientation status, independent of the setting of latch\_int<3:0>. The flag is not updated as long as an orientation

blocking condition is active.

high\_sign: sign of acceleration signal that triggered high-g interrupt was '0'→positive, '1'

→negative

high\_first\_z: high-g interrupt: '1'  $\rightarrow$  triggered by, or '0' $\rightarrow$ not triggered by z-axis high\_first\_y: high-g interrupt: '1'  $\rightarrow$  triggered by, or '0' $\rightarrow$ not triggered by y-axis high-g interrupt: '1'  $\rightarrow$  triggered by, or '0' $\rightarrow$ not triggered by x-axis



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# Register 0x0E (FIFO\_STATUS)

The register contains FIFO status flags.

Name	0x0E	FIFO_STATU	FIFO_STATUS		
Bit	7	6	5	4	
Read/Write	R	R	R	R	
Reset Value	n/a	n/a	n/a	n/a	
Content	fifo_overrun	fifo_frame_co	fifo_frame_counter<6:4>		
Bit	3	2	1	0	
Read/Write	R	R	R	R	
Reset Value	n/a	n/a	n/a	n/a	
Content	fifo frame counter<3:0>				

fifo\_overrun: FIFO overrun condition has '1' → occurred, or '0'→not occurred; flag can be cleared by writing to the FIFO configuration register FIFO\_CONFIG\_1 only

fifo\_frame\_counter<6:4>: Current fill level of FIFO buffer. An empty FIFO corresponds to 0x00. The frame counter can be cleared by reading out all frames from the

FIFO buffer or writing to the FIFO configuration register FIFO\_CONFIG\_1.



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# Register 0x0F (PMU\_RANGE)

The register allows the selection of the accelerometer g-range.

Name	0x0F	PMU_RANGE		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	Reserved			

 Bit
 3
 2
 1
 0

 Read/Write
 R/W
 R/W
 R/W
 R/W

 Reset
 0
 0
 1
 1

 Value
 range<3:0>
 1
 1

range<3:0>: Selection of accelerometer g-range:

'0011b'  $\rightarrow$  ±2g range; '0101b'  $\rightarrow$  ±4g range; '1000b'  $\rightarrow$  ±8g range; '1100b'  $\rightarrow$  ±16g range; all other settings  $\rightarrow$  reserved (do not use)

reserved: write '0'

# Register 0x10 (PMU\_BW)

The register allows the selection of the acceleration data filter bandwidth.

Name	0x10	PMU_BW		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	Reserved			bw<4>

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	1	1	1	1
Value				
Content	bw<3:0>			

bw<4:0>: Selection of data filter bandwidth:

 $'00xxxb' \rightarrow 7.81 \text{ Hz}, \quad '01000b' \rightarrow 7.81 \text{ Hz}, \quad '01001b' \rightarrow 15.63 \text{ Hz}, \\ '01010b' \rightarrow 31.25 \text{ Hz}, \quad '01011b' \rightarrow 62.5 \text{ Hz}, \quad '01100b' \rightarrow 125 \text{ Hz}, \\ '01101b' \rightarrow 250 \text{ Hz}, \quad '01110b' \rightarrow 500 \text{ Hz}, \quad '01111b' \rightarrow \text{ODR}_{max.},$ 

 $'1xxxxb' \rightarrow ODR_{max.}$ 

reserved: write '0'



Value Content

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reserved

## Register 0x11 (PMU\_LPW)

Selection of the main power modes and the low power sleep period.

Name	0x11	PMU_LPW		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	suspend	lowpower_en	deep_suspend	sleep_dur<3>
Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0

suspend, low\_power\_en, deep\_suspend:

Main power mode configuration setting {suspend; lowpower\_en;

deep\_suspend}:

sleep dur<2:0>

 $\{0; 0; 0\} \rightarrow NORMAL mode;$ 

 $\{0; 0; 1\} \rightarrow DEEP\_SUSPEND mode;$ 

 $\{0; 1; 0\} \rightarrow LOW\_POWER mode;$ 

 $\{1; 0; 0\} \rightarrow SUSPEND mode;$ 

{all other} → illegal

Please note that only certain power mode transitions are permitted.

sleep\_dur<3:0>: Configures the sleep phase duration in LOW\_POWER mode:

 '0000b' to '0101b'
  $\rightarrow$  0.5 ms,
 '0110b'  $\rightarrow$  1 ms,

 '0111b'
  $\rightarrow$  2 ms,
 '1000b'  $\rightarrow$  4 ms,

 '1001b'
  $\rightarrow$  6 ms,
 '1010b'  $\rightarrow$  10 ms,

 '1011b'
  $\rightarrow$  25 ms,
 '1100b'  $\rightarrow$  50 ms,

 '1101b'
  $\rightarrow$  100 ms,
 '1110b'  $\rightarrow$  500 ms,

′1111b′ → 1 s

Please note, that all application specific settings which are not equal to the default settings (refer to 6.2 register map), must be re-set to its designated values after DEEP\_SUSPEND.

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# Register 0x12 (PMU\_LOW\_POWER)

Configuration settings for low power mode.

Name	0x12	PMU_LOW_POWER		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	reserved	lowpower_mode	sleeptimer_mode	reserved
Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	Reserved	_		

lowpower\_mode: select '0'  $\rightarrow$  LPM1, or '1'  $\rightarrow$  LPM2 configuration for SUSPEND and

LOW\_POWER mode. In the LPM1 configuration the power consumption in LOW\_POWER mode and SUSPEND mode is significantly reduced when compared to LPM2 configuration, but the FIFO is not accessible and writing to registers must be slowed down. In the LPM2 configuration the power consumption in LOW\_POWER mode is reduced compared to NORMAL mode, but the FIFO is fully accessible and registers can be written to at full

speed.

sleeptimer\_mode: when in LOW\_POWER mode '0' → use event-driven time-base mode

(compatible with BMA250), or '1' → use equidistant sampling time-base

mode. Equidistant sampling of data into the FIFO is maintained in

equidistant time-base mode only.

reserved: write '0'



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## Register 0x13 (ACCD\_HBW)

Acceleration data acquisition and data output format.

Name	0x13	ACCD_HBW		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0 (1 in 8-bit	0	0
Value		mode)		
Content	data_high_bw	shadow_dis	reserved	_

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	Reserved			

data\_high\_bw: select whether '1' → unfiltered, or '0' → filtered data may be read from the

acceleration data registers.

shadow\_dis:  $(1) \rightarrow \text{disable}$ , or  $(0) \rightarrow \text{the shadowing mechanism for the acceleration data}$ 

output registers. When shadowing is enabled, the content of the acceleration data component in the MSB register is locked, when the component in the LSB is read, thereby ensuring the integrity of the acceleration data during

read-out. The lock is removed when the MSB is read.

reserved: write '0'



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# Register 0x14 (BGW\_SOFTRESET)

Controls user triggered reset of the sensor.

Softreset

Name	0x14	BGW_SOFTRESET		
Bit	7	6	5	4
Read/Write	W	W	W	W
Reset	0	0	0	0
Value				
Content	Softreset			
Bit	3	2	1	0
Read/Write	W	W	W	W
Reset	0	0	0	0

softreset:

Value Content

Value

Content

0xB6 → triggers a reset. Other values are ignored. Following a delay, all user configuration settings are overwritten with their default state or the setting stored in the NVM, wherever applicable. This register is functional in all operation modes. Please note that all application specific settings which are not equal to the default settings (refer to 6.2 register map), must be reconfigured to their designated values.

slope\_en\_y

#### Register 0x16 (INT\_EN\_0)

Controls which interrupt engines in group 0 are enabled.

Name	0x16	INT_EN_0		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	flat_en	orient_en	s_tap_en	d_tap_en
Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0

flat en: flat interrupt: '0'→disabled, or '1' →enabled

reserved

orient\_en: orientation interrupt: '0'→disabled, or '1' →enabled s\_tap\_en: single tap interrupt: '0'→disabled, or '1' →enabled d\_tap\_en double tap interrupt: '0'→disabled, or '1' →enabled

reserved: write '0'

slope\_en\_z: slope interrupt, z-axis component: '0'→disabled, or '1' →enabled slope\_en\_y: slope interrupt, y-axis component: '0'→disabled, or '1' →enabled slope\_en\_x: slope interrupt, x-axis component: '0'→disabled, or '1' →enabled

slope\_en\_z

slope\_en\_x



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# Register 0x17 (INT\_EN\_1)

Controls which interrupt engines in group 1 are enabled.

Name	0x17	INT_EN_1		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	reserved	int fwm en	int ffull en	data en

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	low en	high_en_z	high_en_y	high_en_x

reserved: write '0'

int\_fwm\_en: FIFO watermark interrupt: '0'→disabled, or '1' →enabled

int\_ffull\_en: FIFO full interrupt: '0'→disabled, or '1' →enabled data\_en data ready interrupt: '0'→disabled, or '1' →enabled low\_en: low-g interrupt: '0'→disabled, or '1' →enabled

high\_en\_z: high-g interrupt, z-axis component: '0'→disabled, or '1' →enabled high\_en\_y: high-g interrupt, y-axis component: '0'→disabled, or '1' →enabled high\_en\_x: high-g interrupt, x-axis component: '0'→disabled, or '1' →enabled



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# Register 0x18 (INT\_EN\_2)

Controls which interrupt engines in group 2 are enabled.

Name	0x18	INT_EN_2		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	Reserved			

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	slo_no_mot_sel	slo_no_mot_en_z	slo_no_mot_en_y	slo_no_mot_en_x

reserved: write '0'

slo\_no\_mot\_sel: select '0'→slow-motion, '1' →no-motion interrupt function slo\_no\_mot\_en\_z: slow/n-motion interrupt, z-axis component: '0'→disabled, or '1' →enabled slo\_no\_mot\_en\_y: slow/n-motion interrupt, y-axis component: '0'→disabled, or '1' →enabled slo\_no\_mot\_en\_x: slow/n-motion interrupt, x-axis component: '0'→disabled, or '1' →enabled



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## Register 0x19 (INT\_MAP\_0)

Controls which interrupt signals are mapped to the INT1 pin.

Name	0x19	INT_MAP_0		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset Value	0	0	0	0
Content	int1_flat	int1_orient	int1_s_tap	int1_d_tap
	1		T-	

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	int1_slo_no_mot	int1_slope	int1_high	int1_low

int1\_flat: map flat interrupt to INT1 pin: '0'→disabled, or '1' →enabled

int1\_orient: map orientation interrupt to INT1 pin: '0' $\rightarrow$ disabled, or '1'  $\rightarrow$ enabled int1\_s\_tap: map single tap interrupt to INT1 pin: '0' $\rightarrow$ disabled, or '1'  $\rightarrow$ enabled int1\_d\_tap: map double tap interrupt to INT1 pin: '0' $\rightarrow$ disabled, or '1'  $\rightarrow$ enabled int1\_slo\_no\_mot: map slow/no-motion interrupt to INT1 pin: '0' $\rightarrow$ disabled, or '1'  $\rightarrow$ enabled

int1\_slope: map slope interrupt to INT1 pin: '0'→disabled, or '1' →enabled

int1\_high: map high-g to INT1 pin: '0'→disabled, or '1' →enabled int1\_low: map low-g to INT1 pin: '0'→disabled, or '1' →enabled



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# Register 0x1A (INT\_MAP\_1)

Controls which interrupt signals are mapped to the INT1 and INT2 pins.

Name	0x1A	INT_MAP_1		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	int2_data	int2_fwm	int2_ffull	reserved
Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	reserved	int1_ffull	int1_fwm	int1_data

int2\_data: map data ready interrupt to INT2 pin: '0'→disabled, or '1' →enabled

int2\_fwm: map FIFO watermark interrupt to INT2 pin: '0'→disabled, or '1' →enabled

int2\_ffull: map FIFO full interrupt to INT2 pin: '0' → disabled, or '1' → enabled

reserved: write '0'

int1\_ffull: map FIFO full interrupt to INT1 pin: '0'→disabled, or '1' →enabled

int1\_fwm: map FIFO watermark interrupt to INT1 pin: '0'→disabled, or '1' →enabled

int1\_data: map data ready interrupt to INT1 pin: '0'→disabled, or '1' →enabled



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# Register 0x1B (INT\_MAP\_2)

Controls which interrupt signals are mapped to the INT2 pin.

Name	0x1B	INT_MAP_2		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	int2_flat	int2_orient	int2_s_tap	int2_d_tap
Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset Value	0	0	0	0
Content	int2_slo_no_mot	int2_slope	int2_high	int2_low

int2 flat: map flat interrupt to INT2 pin: '0' → disabled, or '1' → enabled

int2\_orient: map orientation interrupt to INT2 pin: '0'→disabled, or '1' →enabled int2\_s\_tap: map single tap interrupt to INT2 pin: '0'→disabled, or '1' →enabled int2\_d\_tap: map double tap interrupt to INT2 pin: '0'→disabled, or '1' →enabled int2\_slo\_no\_mot: map slow/no-motion interrupt to INT2 pin: '0'→disabled, or '1' →enabled

int2\_slope: map slope interrupt to INT2 pin: '0'→disabled, or '1' →enabled

int2\_high: map high-g to INT2 pin: '0'→disabled, or '1' →enabled int2\_low: map low-g to INT2 pin: '0'→disabled, or '1' →enabled



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# Register 0x1E (INT\_SRC)

Contains the data source definition for interrupts with selectable data source.

Name	0x1E	INT_SRC		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	Reserved		int_src_data	int_src_tap
Bit	3	2	1	0
Read/Write	R/M	R/W	R/M	R/W

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	int_src_slo_no_m	int_src_slope	int_src_high	int_src_low
	ot			

reserved: write '0'

int\_src\_data: select '0'→filtered, or '1' →unfiltered data for new data interrupt

int\_src\_tap: select '0'→filtered, or '1' →unfiltered data for single-/double tap interrupt int\_src\_slo\_no\_mot: select '0'→filtered, or '1' →unfiltered data for slow/no-motion interrupt

int\_src\_slope: select '0'→filtered, or '1' →unfiltered data for slope interrupt int\_src\_high: select '0'→filtered, or '1' →unfiltered data for high-g interrupt int\_src\_low: select '0'→filtered, or '1' →unfiltered data for low-g interrupt



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# Register 0x20 (INT\_OUT\_CTRL)

Contains the behavioural configuration (electrical behaviour) of the interrupt pins.

Name	0x20	INT_OUT_CTRL		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	Reserved			

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	1	0	1
Value				
Content	int2_od	int2_lvl	int1_od	int1_lvl

reserved: write '0'

int2\_od: select '0'→push-pull, or '1' →open drain behavior for INT2 pin int2\_lvl: select '0'→active low, or '1'→active high level for INT2 pin int1\_od: select '0'→push-pull, or '1' →open drain behavior for INT1 pin int1\_lvl: select '0'→active low, or '1'→active high level for INT1 pin



# Register 0x21 (INT\_RST\_LATCH)

Contains the interrupt reset bit and the interrupt mode selection.

Name	0x21	INT_RST_LATCH		
Bit	7	6	5	4
Read/Write	W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	reset_int	Reserved		

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	latch int<3:0>			

reset int: write '1'  $\rightarrow$  clear any latched interrupts, or '0'  $\rightarrow$  keep latched interrupts

active

reserved: write '0'

latch\_int<3:0>: '0000b'  $\rightarrow$  non-latched, '0001b'  $\rightarrow$  temporary, 250 ms,

'0010b' → temporary, 500 ms, '0011b' → temporary, 1 s, '0100b' → temporary, 2 s, '0101b' → temporary, 4 s,

 $'0110b' \rightarrow temporary, 8 s, '0111b' \rightarrow latched,$ 

'1000b'  $\rightarrow$  non-latched, '1001b'  $\rightarrow$  temporary, 250  $\mu$ s, '1010b'  $\rightarrow$  temporary, 500  $\mu$ s, '1011b'  $\rightarrow$  temporary, 1 ms, '1100b'  $\rightarrow$  temporary, 12.5 ms, '1101b'  $\rightarrow$  temporary, 25 ms,

 $'1110b' \rightarrow temporary, 50 ms, '1111b' \rightarrow latched$ 

#### Register 0x22 (INT 0)

Contains the delay time definition for the low-g interrupt.

Name	0x22	INT_0		
Bit	7	6	5	4
Read/Write	W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	low_dur<7:4>			

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	1	0	0	1
Value				
Content	low_dur<3:0>			

low\_dur<7:0>: low-g interrupt trigger delay according to [low\_dur<7:0> + 1] • 2 ms in a range from 2 ms to 512 ms; the default corresponds to a delay of 20 ms.

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# Register 0x23 (INT\_1)

Contains the threshold definition for the low-g interrupt.

Name	0x23	INT_1		
Bit	7	6	5	4
Read/Write	W	R/W	R/W	R/W
Reset	0	0	1	1
Value				
Content	low_th<7:4>			

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	low_th<3:0>			

low th<7:0>:

low-g interrupt trigger threshold according to  $low\_th < 7:0>$  • 7.81 mg in a range from 0 g to 1.992 g; the default value corresponds to an acceleration of 375 mg

#### Register 0x24 (INT\_2)

Contains the low-g interrupt mode selection, the low-g interrupt hysteresis setting, and the high-g interrupt hysteresis setting.

Name	0x24	INT_2		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	1	0	0	0
Value				
Content	high_hy<1:0>	<u> </u>	reserved	<u>.</u>
			<u>.</u>	
D:4	2	2	1	Δ.

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	1
Value				
Content	reserved	low_mode	low_hy<1:0>	

high\_hy<1:0>: hysteresis of high-g interrupt according to high\_hy<1:0> · 125 mg (2-g

range), high\_hy<1:0> · 250 mg (4-g range), high\_hy<1:0> · 500 mg (8-g

range), or high\_hy<1:0> · 1000 mg (16-g range)

low mode: select low-g interrupt '0' single-axis mode, or '1' axis-summing mode

low\_hy<1:0>: hysteresis of low-g interrupt according to low\_hy<1:0> · 125 mg independent

of the selected accelerometer g-range

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# Register 0x25 (INT\_3)

Contains the delay time definition for the high-g interrupt.

Name	0x25	INT_3		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	high_dur<7:4>			

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	1	1	1	1
Value				
Content	high_dur<3:0>			

high\_dur<7:0>: high-g interrupt trigger delay according to [high\_dur<7:0> + 1] • 2 ms in a range from 2 ms to 512 ms; the default corresponds to a delay of 32 ms.

# Register 0x26 (INT\_4)

Contains the threshold definition for the high-g interrupt.

Name	0x26	INT_4		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	1	1	0	0
Value				
Content	high_th<7:4>			

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	high_th<3:0>			

high\_th<7:0>: threshold of high-g interrupt according to high\_th<7:0>  $\cdot$  7.81 mg (2-g range), high\_th<7:0>  $\cdot$  15.63 mg (4-g range), high\_th<7:0>  $\cdot$  31.25 mg (8-g range), or high\_th<7:0>  $\cdot$  62.5 mg (16-g range)



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### Register 0x27 (INT\_5)

Contains the definition of the number of samples to be evaluated for the slope interrupt (anymotion detection) and the slow/no-motion interrupt trigger delay.

Name	0x27	INT_5		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	slo_no_mot_dur<5:2>			

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	slo_no_mot_dur<1:	:0>	slope_dur<1:0>	

slo\_no\_mot\_dur<5:0>:

:0>: Function depends on whether the slow-motion or no-motion interrupt function has been selected. If the slow-motion interrupt function has been enabled (slo\_no\_mot\_sel = '0') then [slo\_no\_mot\_dur<1:0>+1] consecutive slope data points must be above the slow/no-motion threshold (slo\_no\_mot\_th) for the slow-/no-motion interrupt to trigger. If the no-motion interrupt function has been enabled (slo\_no\_mot\_sel = '1') then slo\_no\_motion\_dur<5:0> defines the time for which no slope data points must exceed the slow/no-motion threshold (slo\_no\_mot\_th) for the slow/no-motion interrupt to trigger. The delay time in seconds may be calculated according with the following equation:

```
slo_no_mot_dur<5:4>='b00' \rightarrow [slo_no_mot_dur<3:0> + 1] slo_no_mot_dur<5:4>='b01' \rightarrow [slo_no_mot_dur<3:0> · 4 + 20] slo no mot dur<5>='1' \rightarrow [slo no mot dur<4:0> · 8 + 88]
```

slope dur<1:0>:

slope interrupt triggers if [slope\_dur<1:0>+1] consecutive slope data points are above the slope interrupt threshold slope\_th<7:0>



# Register 0x28 (INT\_6)

Contains the threshold definition for the any-motion interrupt.

Name	0x28	INT_6		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	1
Value				
Content	slope_th<7:4>			

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	1	0	0
Value				
Content	slope_th<3:0>			

slope th<7:0>: Threshold of the any-motion interrupt. It is range-dependent and defined as a

sample-to-sample difference according to

slope\_th<7:0> · 3.91 mg (2-g range) /

slope\_th<7:0>  $\cdot$  7.81 mg (4-g range) / slope\_th<7:0> · 15.63 mg (8-g range) /

slope\_th<7:0> · 31.25 mg (16-g range)

# Register 0x29 (INT\_7)

Contains the threshold definition for the slow/no-motion interrupt.

Name	0x29	INT_7		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	1
Value				
Content	slo_no_mot_th<7:4>			

Bit	3	2	1	0	
Read/Write	R/W	R/W	R/W	R/W	
Reset	0	1	0	0	
Value					
Content	slo no mot th<3:0>				

slo no mot th<7:0>: Threshold of slow/no-motion interrupt. It is range-dependent and defined

as a sample-to-sample difference according to slo\_no\_mot\_th<7:0> · 3..91 mg (2-g range),

 $slo_no_mot_th<7:0> \cdot 7.81 mg (4-g range),$ 

slo\_no\_mot\_th<7:0> · 15.63 mg (8-g range), slo\_no\_mot\_th<7:0> · 31.25 mg (16-g range)



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# Register 0x2A (INT\_8)

Contains the timing definitions for the single tap and double tap interrupts.

Name	0x2A	INT_8		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	tap_quiet	tap_shock	reserved	reserved
Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	1	0	0

tap\_quiet: selects a tap quiet duration of '0'  $\rightarrow$  30 ms, '1'  $\rightarrow$  20 ms tap\_shock: selects a tap shock duration of '0'  $\rightarrow$  50 ms, '1'  $\rightarrow$  75 ms

reserved: write '0'

reserved

Value

Content

tap\_dur<2:0>: selects the length of the time window for the second shock event for double

tap\_dur<2:0>

tap detection according to '000b'  $\rightarrow$  50 ms, '001b'  $\rightarrow$  100 ms, '010b'  $\rightarrow$  150 ms, '011b'  $\rightarrow$  200 ms, '100b'  $\rightarrow$  250 ms, '101b'  $\rightarrow$  375 ms, '110b'  $\rightarrow$  500

ms, '111b'  $\rightarrow$  700 ms.



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# Register 0x2B (INT\_9)

Contains the definition of the number of samples processed by the single / double-tap interrupt engine after wake-up in low-power mode. It also defines the threshold definition for the single and double tap interrupts.

Name	0x2B	INT_9		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	tap_samp<1:0>		reserved	tap_th<4>

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	1	0	1	0
Value				
Content	tap_th<3:0>			

tap\_samp<1:0>: selects the number of samples that are processed after wake-up in the low-

power mode according to '00b'  $\rightarrow$  2 samples, '01b'  $\rightarrow$  4 samples, '10b'  $\rightarrow$  8

samples, and '11b' → 16 samples

reserved: write '0'

tap\_th<4:0>: threshold of the single/double-tap interrupt corresponding to an acceleration

difference of tap\_th<4:0>  $\cdot$  62.5mg (2g-range), tap\_th<4:0>  $\cdot$  125mg (4g-range), tap\_th<4:0>  $\cdot$  250mg (8g-range), and tap\_th<4:0>  $\cdot$  500mg (16g-

range).



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# Register 0x2C (INT\_A)

Contains the definition of hysteresis, blocking, and mode for the orientation interrupt

Name	0x2C	INT_A		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	1
Value				
Content	reserved	orient_hyst<2:0>		

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	1	0	0	0
Value				
Content	orient_blocking<1:0>		orient_mode<1:0>	

reserved: write '0'

orient\_hyst<2:0>: sets the hysteresis of the orientation interrupt; 1 LSB corresponds to 62.5 mg

irrespective of the selected g-range

orient blocking<1:0>: selects the blocking mode that is used for the generation of the

orientation interrupt. The following blocking modes are available:

'00b' → no blocking,

'01b'  $\rightarrow$  theta blocking or acceleration in any axis > 1.5g,

'10b' → ,theta blocking or acceleration slope in any axis > 0.2 g or

acceleration in any axis > 1.5g

'11b' → theta blocking or acceleration slope in any axis > 0.4 g or acceleration in any axis > 1.5g and value of orient is not stable for at least 100ms

orient\_mode<1:0>: sets the thresholds for switching between the different orientations. The settings: '00b'  $\rightarrow$  symmetrical, '01b'  $\rightarrow$  high-asymmetrical, '10b'  $\rightarrow$  low-asymmetrical, '11b'  $\rightarrow$  symmetrical.



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# Register 0x2D (INT\_B)

Contains the definition of the axis orientation, up/down masking, and the theta blocking angle for the orientation interrupt.

Name	0x2D	INT_B		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	n/a	1	0	0
Value				
Content	reserved	orient_ud_en	orient_theta<5:4>	
	T			T
Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	1	0	0	0
Value				
Content	orient_theta<3:0>			

orient\_ud\_en: change of up/down-bit '1'  $\rightarrow$  generates an orientation interrupt, '0'  $\rightarrow$  is

ignored and will not generate an orientation interrupt

orient\_theta<5:0>: defines a blocking angle between 0° and 44.8°

# Register 0x2E (INT\_C)

Contains the definition of the flat threshold angle for the flat interrupt.

Name	0x2E	INT_C			
Bit	7	6	5	4	
Read/Write	R/W	R/W	R/W	R/W	
Reset	n/a	n/a	0	0	
Value					
Content	Reserved		flat_theta<5:4>		
Bit	3	2	1	0	
Read/Write	R/W	R/W	R/W	R/W	
Reset	1	0	0	0	
Value					
Content	flat_theta<3:0>				

reserved: write '0'

flat\_theta<5:0>: defines threshold for detection of flat position in range from 0° to 44.8°.



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# Register 0x2F (INT\_D)

Contains the definition of the flat interrupt hold time and flat interrupt hysteresis.

Name	0x2F	INT_D		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	1
Value				
Content	Reserved		flat_hold_time<1:0>	>

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	1
Value				
Content	reserved	flat_hy<2:0>		

reserved: write '0'

flat\_hold\_time<1:0>: delay time for which the flat value must remain stable for the flat interrupt

to be generated: '00b'  $\rightarrow$  0 ms, '01b'  $\rightarrow$  512 ms, '10b'  $\rightarrow$  1024 ms,

'11b' → 2048 ms

flat\_hy<2:0>: defines flat interrupt hysteresis; flat value must change by more than twice

the value of flat interrupt hysteresis to detect a state change. For details see

chapter 4.7.8.

'000b' → hysteresis of the flat detection disabled



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# Register 0x30 (FIFO\_CONFIG\_0)

Contains the FIFO watermark level.

Name	0x30	FIFO_CONFIG_0			
Bit	7	6	5	4	
Read/Write	R/W	R/W	R/W	R/W	
Reset Value	n/a	n/a	0	0	
Content	Reserved		fifo_water_mark_le 5:4>	ark_level_trigger_retain<	
Bit	3	2	1	0	
Read/Write	R/W	R/W	R/W	R/W	
Reset	0 0		0	0	
Value					
Content	fifo_water_mark_le	vel_trigger_retain<3:	:0>		

reserved: write '0'

fifo\_water\_mark\_level\_trigger\_retain<5:0>:

fifo\_water\_mark\_level\_trigger\_retain<5:0> defines the FIFO watermark level. An interrupt will be generated, when the number of entries in the FIFO is equal to fifo\_water\_mark\_level\_trigger\_retain<5:0>;



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# Register 0x32 (PMU\_SELF\_TEST)

Contains the settings for the sensor self-test configuration and trigger.

Name	0x32	PMU_SELF	PMU_SELF_TEST		
Bit	7	6	5	4	
Read/Write	R/W	R/W	R/W	R/W	
Reset	0	0	0	0	
Value					
Content	reserved				

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	reserved 0	self test sign	self test-axis<1:0>	

reserved: write '0x0'

self\_test\_sign: select sign of self-test excitation as '1'  $\rightarrow$  positive, or '0'  $\rightarrow$  negative

self\_test\_axis: select axis to be self-tested: '00b' → self-test disabled, '01b' → x-axis, '10b'

 $\rightarrow$  y-axis, or '11b'  $\rightarrow$  z-axis; when a self-test is performed, only the

acceleration data readout value of the selected axis is valid; after the selftest has been enabled a delay of a least 50 ms is necessary for the read-out

value to settle



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# Register 0x33 (TRIM\_NVM\_CTRL)

Contains the control settings for the few-time programmable non-volatile memory (NVM).

Name	0x33	TRIM_NVM_CTRL		
Bit	7	6	5	4
Read/Write	R	R	R	R
Reset	n/a	n/a	n/a	n/a
Value				
Content	nvm_remain<3:0>	_	_	_

Bit	3	2	1	0
Read/Write	R/W	R	W	R/W
Reset	0	n/a	0	0
Value				
Content	nvm_load	nvm_rdy	nvm_prog_trig	nvm_prog_mode

nvm\_remain<3:0>:number of remaining write cycles permitted for NVM; the number is

decremented each time a write to the NVM is triggered

nvm\_load:  $1' \rightarrow \text{trigger}$ , or '0'  $\rightarrow \text{do not trigger}$  an update of all configuration registers

from NVM; the nvm\_rdy flag must be '1' prior to triggering the update

nvm\_rdy: status of NVM controller: '0' → NVM write / NVM update operation is in

progress, '1'  $\rightarrow$  NVM is ready to accept a new write or update trigger

nvm\_prog\_trig: '1'  $\rightarrow$  trigger, or '0'  $\rightarrow$  do not trigger an NVM write operation; the trigger is

only accepted if the NVM was unlocked before and nvm\_remain<3:0> is greater than '0'; flag nvm rdy must be '1' prior to triggering the write cycle

nvm\_prog\_mode: '1' → unlock, or '0' → lock NVM write operation



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# Register 0x34 (BGW\_SPI3\_WDT)

Contains settings for the digital interfaces.

Name	0x34	BGW_SPI3_WDT		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	Reserved			

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	reserved	i2c_wdt_en	i2c_wdt_sel	spi3

reserved: write '0'

i2c\_wdt\_en: if I2C interface mode is selected then '1'  $\rightarrow$  enable, or '0'  $\rightarrow$  disables the

watchdog at the SDI pin (= SDA for I2C)

i2c\_wdt\_sel: select an I2C watchdog timer period of '0'  $\rightarrow$  1 ms, or '1'  $\rightarrow$  50 ms

spi3: select '0'  $\rightarrow$  4-wire SPI, or '1'  $\rightarrow$  3-wire SPI mode



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# Register 0x36 (OFC\_CTRL)

Contains control signals and configuration settings for the fast and the slow offset compensation.

Name	0x36	OFC_CTRL		
Bit	7	6	5	4
Read/Write	W	W	W	R
Reset	0	0	0	0
Value				
Content	offset_reset	cal_trigger<1:0	>	cal_rdy
D:4		2	14	
Bit	3	2		U
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	reserved	hp_z_en	hp_y_en	hp_x_en

offset\_reset: '1'  $\rightarrow$  set all offset compensation registers (0x38 to 0x3A) to zero, or '0'  $\rightarrow$ 

keep their values

offset\_trigger<1:0>: trigger fast compensation for '01b'  $\rightarrow$  x-axis, '10b'  $\rightarrow$  y-axis, or '11b'  $\rightarrow$ 

z-axis; '00b' → do not trigger offset compensation; offset compensation

must not be triggered when cal rdy is '0'

cal\_rdy: indicates the state of the fast compensation:  $0' \rightarrow 0$  offset compensation is in

progress, or '1'  $\rightarrow$  offset compensation is ready to be retriggered

reserved: write '0'

hp\_z\_en: '1'  $\rightarrow$  enable, or '0'  $\rightarrow$  disable slow offset compensation for the z-axis hp\_y\_en: '1'  $\rightarrow$  enable, or '0'  $\rightarrow$  disable slow offset compensation for the y-axis hp\_x\_en: '1'  $\rightarrow$  enable, or '0'  $\rightarrow$  disable slow offset compensation for the x-axis

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# Register 0x37 (OFC\_SETTING)

Contains configuration settings for the fast and the slow offset compensation.

Name	0x37	OFC_SETTING		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	reserved	offset_target_z<1:0>		offset_target_y<1
				>
	1	1	1	1
Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	offset_target_y<0	offset_target_x<1:0>		cut_off
	>			

reserved: write '0'

offset\_target\_z<1:0>: offset compensation target value for z-axis is '00b'  $\rightarrow$  0 g, '01b'  $\rightarrow$  +1 g, '10b'  $\rightarrow$  -1 g, or '11b'  $\rightarrow$  0 g

offset\_target\_y<1:0>: offset compensation target value for y-axis is '00b'  $\rightarrow$  0 g, '01b'  $\rightarrow$  +1 g, '10b'  $\rightarrow$  -1 g, or '11b'  $\rightarrow$  0 g

offset\_target\_x<1:0>: offset compensation target value for x-axis is '00b'  $\rightarrow$  0 g, '01b'  $\rightarrow$  +1 g, '10b'  $\rightarrow$  -1 g, or '11b'  $\rightarrow$  0 g

cut\_off:

(0x37) cut_off	high-pass filter bandwidth	Example <i>bw</i> = 500 Hz
0b	$\frac{1Hz \times bw *}{1000 Hz}$	$\frac{1Hz \times 500  Hz}{1000  Hz} = 0.5  Hz$
1b	$\frac{10Hz \times bw *}{1000 Hz}$	$\frac{10Hz \times 500  Hz}{1000  Hz} = 5  Hz$

<sup>\*</sup>bw: please insert selected decimal data bandwidth value [Hz] from table 4



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# Register 0x38 (OFC\_OFFSET\_X)

Contains the offset compensation value for x-axis acceleration readout data.

Name	0x38	OFC_OFFSET_X		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	offset_x<7:4>			

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	offset_x<3:0>			

offset x<7:0>:

offset value, which is subtracted from the internal filtered and unfiltered x-axis acceleration data; the offset value is represented with two's complement notation, with a mapping of  $+127 \rightarrow +0.992$ g,  $0 \rightarrow 0$  g, and  $-128 \rightarrow -1$  g; the scaling is independent of the selected g-range; the content of the offset\_x<7:0> may be written to the NVM; it is automatically restored from the NVM after each power-on or softreset; offset\_x<7:0> may be written directly by the user; it is generated automatically after triggering the fast offset compensation procedure for the x-axis



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# Register 0x39 (OFC\_OFFSET\_Y)

Contains the offset compensation value for y-axis acceleration readout data.

Name	0x39	OFC_OFFSET_Y		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	offset_y<7:4>			

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	offset_y<3:0>			

offset y < 7:0 > :

offset value, which is subtracted from the internal filtered and unfiltered y-axis acceleration data; the offset value is represented with two's complement notation, with a mapping of  $+127 \rightarrow +0.992$ g,  $0 \rightarrow 0$  g, and  $-128 \rightarrow -1$  g; the scaling is independent of the selected g-range; the content of the offset\_y<7:0> may be written to the NVM; it is automatically restored from the NVM after each power-on or softreset; offset\_y<7:0> may be written directly by the user; it is generated automatically after triggering the fast offset compensation procedure for the y-axis



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# Register 0x3A (OFC\_OFFSET\_Z)

Contains the offset compensation value for z-axis acceleration readout data.

Name	0x3A	OFC_OFFSET_Z		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	offset_z<7:4>			

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	offset_z<3:0>			

offset\_ z<7:0>:

offset value, which is subtracted from the internal filtered and unfiltered z-axis acceleration data; the offset value is represented with two's complement notation, with a mapping of +127  $\rightarrow$  +0.992g, 0  $\rightarrow$  0 g, and -128  $\rightarrow$  -1 g; the scaling is independent of the selected g-range; the content of the offset\_z<7:0> may be written to the NVM; it is automatically restored from the NVM after each power-on or softreset; offset\_z<7:0> may be written directly by the user; it is generated automatically after triggering the fast offset compensation procedure for the z-axis

# Register 0x3B (TRIM\_GP0)

Contains general purpose data register with NVM back-up.

Name	0x3B	TRIM_GP0		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	GP0<7:4>			

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	GP0<3:0>			

GP0<7:0>:

general purpose NVM image register not linked to any sensor-specific functionality; register may be written to NVM and is restored after each power-up or softreset



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# Register 0x3C (TRIM\_GP1)

Contains general purpose data register with NVM back-up.

Name	0x3C	TRIM_GP1		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	GP1<7:4>			

Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	GP1<3:0>			

GP1<7:0>:

general purpose NVM image register not linked to any sensor-specific functionality; register may be written to NVM and is restored after each power-up or softreset



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# Register 0x3E (FIFO\_CONFIG\_1)

Contains FIFO configuration settings. The FIFO buffer memory is cleared and the fifo-full flag is cleared when writing to FIFO\_CONFIG\_1 register.

Name	0x3E	FIFO_CONFIG_1		
Bit	7	6	5	4
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	fifo_mode<1:0>		Reserved	
Bit	3	2	1	0
Read/Write	R/W	R/W	R/W	R/W
Reset	0	0	0	0
Value				
Content	Reserved		fifo_data_select<1	:0>

fifo mode<1:0>: selects the FIFO operating mode:

'00b' → BYPASS (buffer depth of 1 frame; old data is discarded),

'01b' → FIFO (data collection stops when buffer is filled with 32 frames), '10b' → STREAM (sampling continues when buffer is full; old is discarded),

'11b' → reserved, do not use

fifo\_data\_select<1:0>: selects whether '00b'  $\rightarrow$  X+Y+Z, '01b'  $\rightarrow$  X only, '10b'  $\rightarrow$  Y only,

 $'11b' \rightarrow Z$  only acceleration data are stored in the FIFO



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# Register 0x3F (FIFO\_DATA)

FIFO data readout register. The format of the LSB and MSB components corresponds to that of the acceleration data readout registers. The new data flag is preserved. Read burst access may be used since the address counter will not increment when the read burst is started at the address of FIFO\_DATA. The entire frame is discarded when a fame is only partially read out.

	1						
D IAA 14 D D	4						
Read/Write R R	R						
Reset n/a n/a n/a	n/a						
Value							
Content fifo_data_output_register<7:4>	fifo_data_output_register<7:4>						

Bit	3	2	1	0			
Read/Write	R	R	R	R			
Reset	n/a	n/a	n/a	n/a			
Value							
Content	fifo_data_output_register<3:0>						

fifo\_data\_output\_register<7:0>: FIFO data readout; data format depends on the setting of register fifo\_data\_select<1:0>:

if X+Y+Z data are selected, the data of frame n is reading out in the order of X-lsb(n), X-msb(n), Y-lsb(n), Y-msb(n), Z-lsb(n), Z-msb(n);

if X-only is selected, the data of frame n and n+1 are reading out in the order of X-lsb(n), X-msb(n), X-lsb(n+1), X-msb(n+1); the Y-only and Z-only modes behave analogously



# 7. Digital interfaces

The BMA280 supports two serial digital interface protocols for communication as a slave with a host device (when operating in general mode): SPI and I<sup>2</sup>C. The active interface is selected by the state of the Pin#11 (PS) 'protocol select' pin: '0' ('1') selects SPI (I<sup>2</sup>C). For details please refer to section 8).

By default, SPI operates in the standard 4-wire configuration. It can be re-configured by software to work in 3-wire mode instead of standard 4-wire mode.

Both interfaces share the same pins. The mapping for each interface is given in the following table:

Table 20: Mapping of the interface pins

Pin#	Name	use w/ SPI	use w/ I <sup>2</sup> C	Description
1	SDO	SDO	address	SPI: Data Output (4-wire mode) I <sup>2</sup> C: Used to set LSB of I <sup>2</sup> C address
2	SDx	SDI	SDA	SPI: Data Input (4-wire mode) Data Input / Output (3-wire mode) I²C: Serial Data
10	CSB	CSB	unused	Chip Select (enable)
12	SCx	SCK	SCL	SPI: Serial Clock I <sup>2</sup> C: Serial Clock

The following table shows the electrical specifications of the interface pins:

Table 21: Electrical specification of the interface pins

Parameter	Symbol	Condition	Min	Тур	Max	Units
Pull-up Resistance, CSB pin	$R_{up}$	Internal Pull-up Resistance to VDDIO	75	100	125	kΩ
Input Capacitance	C <sub>in</sub>			5	10	pF
I <sup>2</sup> C Bus Load Capacitance (max. drive capability)	$C_{12C\_Load}$				400	pF

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# 7.1 Serial peripheral interface (SPI)

The timing specification for SPI of the BMA280 is given in the following table:

Table 22: SPI timing

Parameter	Symbol	Condition	Min	Max	Units
Clock Frequency	f <sub>SPI</sub>	Max. Load on SDI or SDO = 25pF, $V_{DDIO} \ge 1.62V$		10	MHz
		$V_{DDIO} < 1.62V$		7.5	MHz
SCK Low Pulse	$t_{SCKL}$		20		ns
SCK High Pulse	t <sub>SCKH</sub>		20		ns
SDI Setup Time	$t_{SDI\_setup}$		20		ns
SDI Hold Time	t <sub>SDI_hold</sub>		20		ns
		Load = 25pF, $V_{DDIO} \ge 1.62V$		30	ns
SDO Output Delay	t <sub>SDO_OD</sub>	Load = 25pF, V <sub>DDIO</sub> < 1.62V		50	ns
		Load = 250pF, $V_{DDIO} > 2.4V$		40	ns
CSB Setup Time	t <sub>CSB_setup</sub>		20		ns
CSB Hold Time	t <sub>CSB_hold</sub>		40		ns
Idle time between write accesses, normal mode, standby mode, low-power mode 2	t <sub>IDLE_wacc_nm</sub>		2		μs
Idle time between write accesses, suspend mode, low-power mode 1	t <sub>IDLE_wacc_sum</sub>		450		μs



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The following figure shows the definition of the SPI timings given in the following figure:

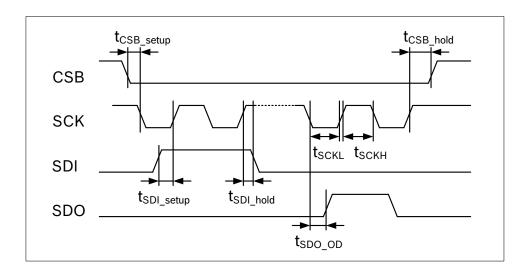


Figure 13: SPI timing diagram

The SPI interface of the BMA280 is compatible with two modes, '00' and '11'. The automatic selection between [CPOL = '0' and CPHA = '0'] and [CPOL = '1' and CPHA = '1'] is controlled based on the value of SCK after a falling edge of CSB.

Two configurations of the SPI interface are supported by the BMA280: 4-wire and 3-wire. The same protocol is used by both configurations. The device operates in 4-wire configuration by default. It can be switched to 3-wire configuration by writing '1' to (0x34) spi3. Pin SDI is used as the common data pin in 3-wire configuration.

For single byte read as well as write operations, 16-bit protocols are used. The BMA280 also supports multiple-byte read operations.

In SPI 4-wire configuration CSB (chip select low active), SCK (serial clock), SDI (serial data input), and SDO (serial data output) pins are used. The communication starts when the CSB is pulled low by the SPI master and stops when CSB is pulled high. SCK is also controlled by SPI master. SDI and SDO are driven at the falling edge of SCK and should be captured at the rising edge of SCK.



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The basic write operation waveform for 4-wire configuration is depicted in figure 14. During the entire write cycle SDO remains in high- impedance state.

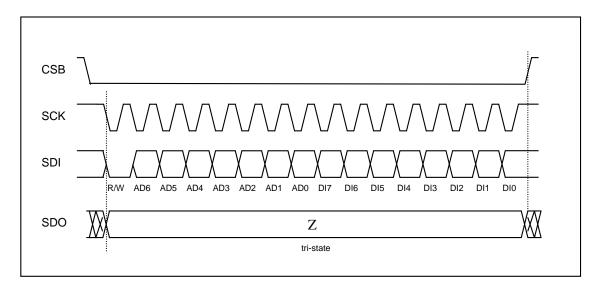


Figure 14: 4-wire basic SPI write sequence (mode '11')

The basic read operation waveform for 4-wire configuration is depicted in figure 15:

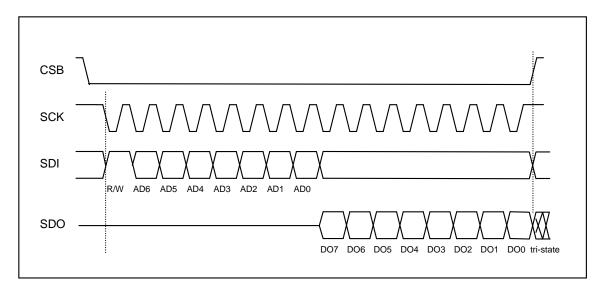


Figure 15: 4-wire basic SPI read sequence (mode '11')

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The data bits are used as follows:

Bit0: Read/Write bit. When 0, the data SDI is written into the chip. When 1, the data SDO from the chip is read.

Bit1-7: Address AD(6:0).

Bit8-15: when in write mode, these are the data SDI, which will be written into the address. When in read mode, these are the data SDO, which are read from the address.

Multiple read operations are possible by keeping CSB low and continuing the data transfer. Only the first register address has to be written. Addresses are automatically incremented after each read access as long as CSB stays active low.

The principle of multiple read is shown in figure 16:

			- 0	ontr	ol byt	е						Data	byte							Data	byte							Data	byte				
Start	RW		Re	gister	adre	ss (02	2h)			Da	ata re	gister	- adre	ess 0	2h			Da	ata re	gister	- adr	ess 0	3h			Da	ata re	gister	- adre	ess 0	4h		Stop
CSB																																	CSB
=	1	0	0	0	0	0	1	0	Х	Х	Χ	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Χ	Х	Х	Х	Х	Х	Х	Х	Х	Х	=
0																																	1

Figure 16: SPI multiple read

In SPI 3-wire configuration CSB (chip select low active), SCK (serial clock), and SDI (serial data input and output) pins are used. The communication starts when the CSB is pulled low by the SPI master and stops when CSB is pulled high. SCK is also controlled by SPI master. SDI is driven (when used as input of the device) at the falling edge of SCK and should be captured (when used as the output of the device) at the rising edge of SCK.

The protocol as such is the same in 3-wire configuration as it is in 4-wire configuration. The basic operation waveform (read or write access) for 3-wire configuration is depicted in figure 17:

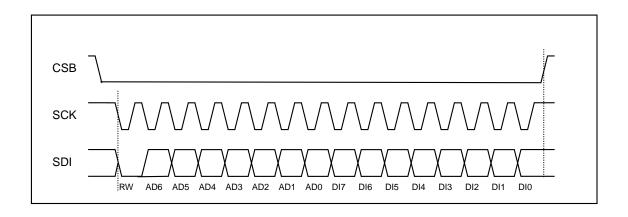


Figure 17: 3-wire basic SPI read or write sequence (mode '11')



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# 7.2 Inter-Integrated Circuit (I<sup>2</sup>C)

The I<sup>2</sup>C bus uses SCL (= SCx pin, serial clock) and SDA (= SDx pin, serial data input and output) signal lines. Both lines are connected to  $V_{\text{DDIO}}$  externally via pull-up resistors so that they are pulled high when the bus is free.

The I<sup>2</sup>C interface of the BMA280 is compatible with the I<sup>2</sup>C Specification UM10204 Rev. 03 (19 June 2007), available at http://www.nxp.com. The BMA280 supports I<sup>2</sup>C standard mode and fast mode, only 7-bit address mode is supported.

The default I<sup>2</sup>C address of the device is 0011000b (0x18). It is used if the SDO pin is pulled to 'GND'. The alternative address 0011001b (0x19) is selected by pulling the SDO pin to ' $V_{DDIO}$ '.

The timing specification for I<sup>2</sup>C of the BMA280 is given in Table 23:

Table 23: I<sup>2</sup>C timings

Parameter	Symbol	Condition	Min	Max	Units
Clock Frequency	$f_{SCL}$			400	kHz
SCL Low Period	t <sub>LOW</sub>		1.3		
SCL High Period	t <sub>HIGH</sub>		0.6		
SDA Setup Time	tsudat		0.1		
SDA Hold Time	t <sub>HDDAT</sub>		0.0		
Setup Time for a repeated Start Condition	<b>t</b> susta		0.6		μS
Hold Time for a Start Condition	t <sub>HDSTA</sub>		0.6		μ3
Setup Time for a Stop Condition	<b>t</b> susto		0.6		
Time before a new Transmission can start	t <sub>BUF</sub>		1.3		
Idle time between write accesses, normal mode, standby mode, low-power mode 2	t <sub>IDLE_wacc_n</sub>		2		μs
Idle time between write accesses, suspend mode, lowpower mode 1	t <sub>IDLE_wacc_s</sub>		450		μs



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Figure 18 shows the definition of the I<sup>2</sup>C timings given in Table 23:

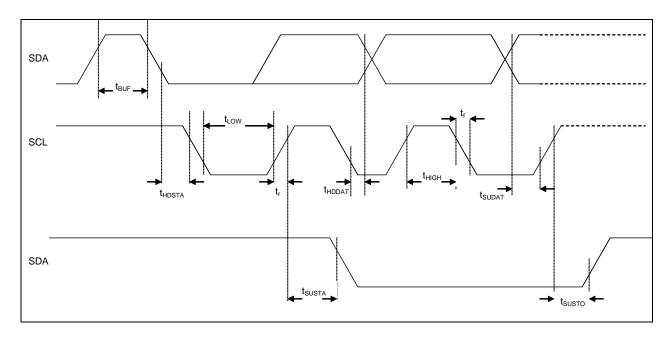


Figure 18: I2C timing diagram

The I2C protocol works as follows:

**START**: Data transmission on the bus begins with a high to low transition on the SDA line while SCL is held high (start condition (S) indicated by I<sup>2</sup>C bus master). Once the START signal is transferred by the master, the bus is considered busy.

**STOP**: Each data transfer should be terminated by a Stop signal (P) generated by master. The STOP condition is a low to HIGH transition on SDA line while SCL is held high.

**ACK**: Each byte of data transferred must be acknowledged. It is indicated by an acknowledge bit sent by the receiver. The transmitter must release the SDA line (no pull down) during the acknowledge pulse while the receiver must then pull the SDA line low so that it remains stable low during the high period of the acknowledge clock cycle.

In the following diagrams these abbreviations are used:

S	Start
Р	Stop

ACKS Acknowledge by slave
ACKM Acknowledge by master
NACKM Not acknowledge by master

RW Read / Write

A START immediately followed by a STOP (without SCK toggling from logic "1" to logic "0") is not supported. If such a combination occurs, the STOP is not recognized by the device.



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#### I<sup>2</sup>C write access:

I<sup>2</sup>C write access can be used to write a data byte in one sequence.

The sequence begins with start condition generated by the master, followed by 7 bits slave address and a write bit (RW = 0). The slave sends an acknowledge bit (ACK = 0) and releases the bus. Then the master sends the one byte register address. The slave again acknowledges the transmission and waits for the 8 bits of data which shall be written to the specified register address. After the slave acknowledges the data byte, the master generates a stop signal and terminates the writing protocol.

Example of an I2C write access:

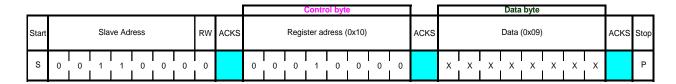


Figure 19: I2C write

#### I<sup>2</sup>C read access:

I<sup>2</sup>C read access also can be used to read one or multiple data bytes in one sequence.

A read sequence consists of a one-byte I<sup>2</sup>C write phase followed by the I<sup>2</sup>C read phase. The two parts of the transmission must be separated by a repeated start condition (Sr). The I<sup>2</sup>C write phase addresses the slave and sends the register address to be read. After slave acknowledges the transmission, the master generates again a start condition and sends the slave address together with a read bit (RW = 1). Then the master releases the bus and waits for the data bytes to be read out from slave. After each data byte the master has to generate an acknowledge bit (ACK = 0) to enable further data transfer. A NACKM (ACK = 1) from the master stops the data being transferred from the slave. The slave releases the bus so that the master can generate a STOP condition and terminate the transmission.

The register address is automatically incremented and, therefore, more than one byte can be sequentially read out. Once a new data read transmission starts, the start address will be set to the register address specified in the latest I<sup>2</sup>C write command. By default the start address is set at 0x00. In this way repetitive multi-bytes reads from the same starting address are possible.

In order to prevent the  $I^2C$  slave of the device to lock-up the  $I^2C$  bus, a watchdog timer (WDT) is implemented. The WDT observes internal  $I^2C$  signals and resets the  $I^2C$  interface if the bus is locked-up by the BMA280. The activity and the timer period of the WDT can be configured through the bits (0x34)  $i2c_wdt_en$  and (0x34)  $i2c_wdt_sel$ .

Writing '1' ('0') to (0x34)  $i2c\_wdt\_en$  activates (de-activates) the WDT. Writing '0' ('1') to (0x34)  $i2c\_wdt\_se$  selects a timer period of 1 ms (50 ms).



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Example of an I2C read access:

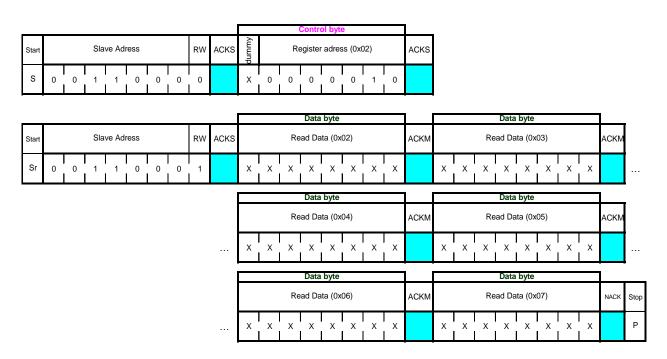


Figure 20: I2C multiple read

#### 7.2.1 SPI and I<sup>2</sup>C Access Restrictions

In order to allow for the correct internal synchronisation of data written to the BMA280, certain access restrictions apply for consecutive write accesses or a write/read sequence through the SPI as well as I2C interface. The required waiting period depends on whether the device is operating in normal mode (or standby mode, or low-power mode 2) or suspend mode (or low-power mode 1).

As illustrated in figure 21, an interface idle time of at least 2  $\mu$ s is required following a write operation when the device operates in normal mode (or standby mode, or low-power mode 2). In suspend mode (or low-power mode 1) an interface idle time of least 450  $\mu$ s is required.

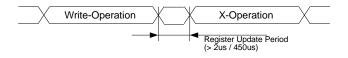
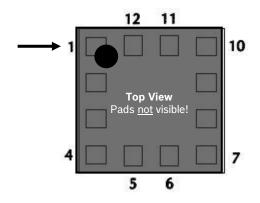


Figure 21: Post-Write Access Timing Constraints



# 8. Pin-out and connection diagram

# 8.1 Pin-out



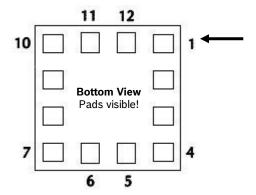


Figure 22: Pin-out top view

Figure 23: Pin-out bottom view

Table 24: Pin description

D:#	Name	1/0 T	D		Connect	to		
Pin#	Name	I/O Type	Description	in SPI 4W	In SPI 3W	in I <sup>2</sup> C		
1	SDO	Digital out	Serial data output in SPI Address select in I <sup>2</sup> C mode see chapter 7.2	SDO	DNC (float)	GND for default addr.		
2	SDx	Digital I/O	SDA serial data I/O in I <sup>2</sup> C SDI serial data input in SPI 4W SDA serial data I/O in SPI 3W	SDI	SDA	SDA		
3	VDDIO	Supply	Digital I/O supply voltage (1.2V 3.6V)	$V_{DDIO}$	$V_{DDIO}$	$V_{\text{DDIO}}$		
4	NC			GND	GND	GND		
5	INT1	Digital out	Interrupt output 1 *	INT1	INT1	INT1		
6	INT2	Digital out	Interrupt output 2 *	INT2	INT2	INT2		
7	VDD	Supply	Power supply for analog & digital domain (1.62V 3.6V)	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DD</sub>		
8	GNDIO	Ground	Ground for I/O	GND	GND	GND		
9	GND	Ground	Ground for digital & analog	GND	GND	GND		
10	CSB	Digital in	Chip select for SPI mode	CSB	CSB	DNC (float)		
11	PS	Digital in	Protocol select (GND = SPI, $V_{DDIO} = I^2C$ )	GND	GND	V <sub>DDIO</sub>		
12	SCx	Digital in	SCK for SPI serial clock SCL for I <sup>2</sup> C serial clock	SCK	SCK	SCL		

 $<sup>\</sup>mbox{{\fontfamily{\fontfamil}{\fontfamily{\fontfamil}{\fontfamil}{\fontfamil}{\fontfamil}{\fontfamil}{\fontfamil}{\fontfamil}{\fontfamil}{\fontfamil}{\fontfamil}{\fontfamil}{\fontfami$ 



# 8.2 Connection diagram 4-wire SPI

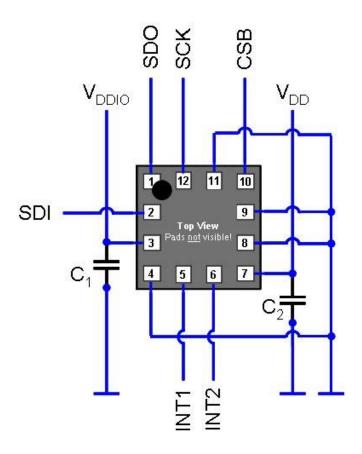


Figure 24: 4-wire SPI connection

Note: the recommended value for C1, C2 is 100 nF.

# 8.3 Connection diagram 3-wire SPI

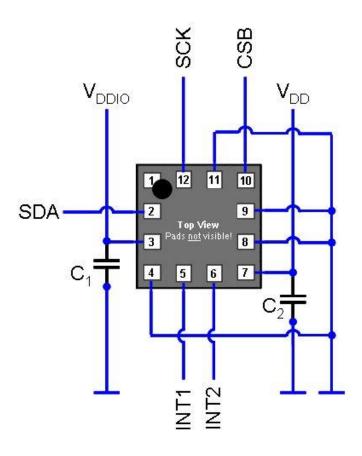


Figure 25: 3-wire SPI connection

Note: the recommended value for C1, C2 is 100 nF.



# 8.4 Connection diagram I<sup>2</sup>C

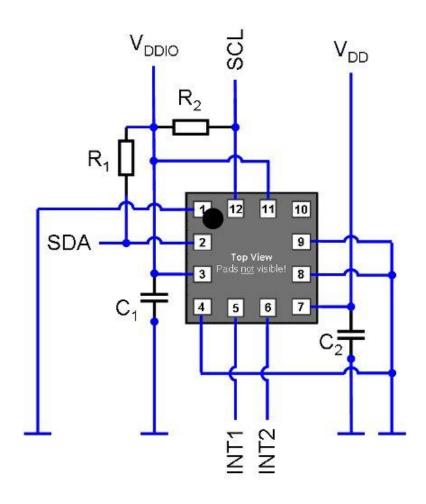


Figure 26: I<sup>2</sup>C connection

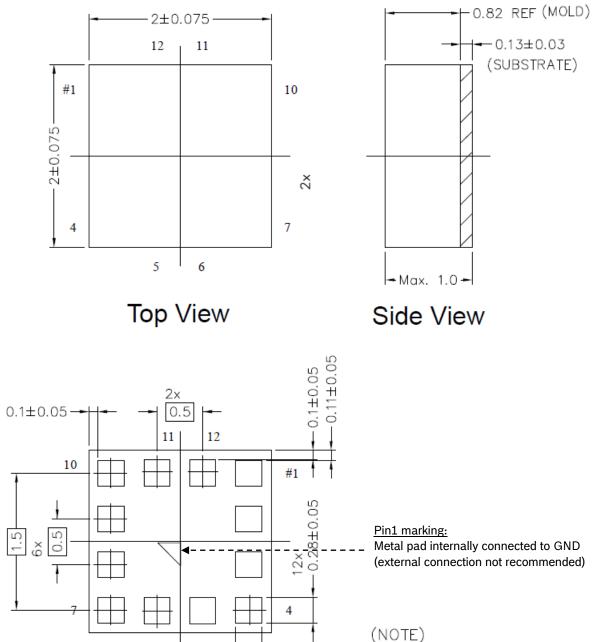
Note: the recommended value for  $C_1$ ,  $C_2$  is 100 nF.



# 9. Package

#### 9.1 Outline dimensions

The sensor housing is a standard LGA package. Its dimensions are the following.



**Bottom View** 

6

5

Figure 27: Package outline dimensions

12x

-0.29±0.05

UNLESS OTHERWISE SPECIFIED

TOLERANCE: DECIMAL±0.05

Unit: mm



# 9.2 Sensing axes orientation

If the sensor is accelerated in the indicated directions, the corresponding channel will deliver a positive acceleration signal (dynamic acceleration). If the sensor is at rest and the force of gravity is acting along the indicated directions, the output of the corresponding channel will be negative (static acceleration).

Example: If the sensor is at rest or at uniform motion in a gravity field according to the figure given below, the output signals are:

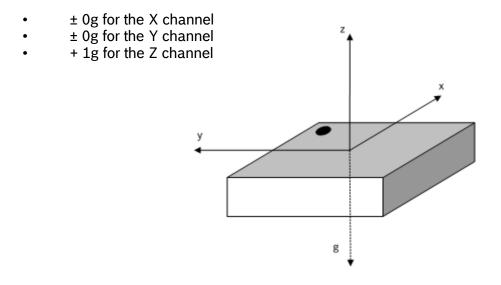


Figure 28: Orientation of sensing axis

The following table lists all corresponding output signals on X, Y, and Z while the sensor is at rest or at uniform motion in a gravity field under assumption of a  $\pm 2g$  range setting and a top down gravity vector as shown above.

Sensor Orientation upright ıhgiriqu (gravity vector⊥) Output Signal X 0g / 0LSB 1g/4096LSB 0g / 0LSB -1g / -4096LSB 0g / 0LSB 0g / 0LSB Output Signal Y 1g/-4096LSB 0g / 0LSB 1g / 4096LSB 0g / 0LSB 0g / 0LSB 0g / 0LSB Output Signal Z 0g / 0LSB 0g / 0LSB 0g / 0LSB 0g / 0LSB 1g/4096LSB -1g/-4096LSB

Table 25: Output signals depending on sensor orientation



# 9.3 Landing Pattern Recommendation

For the design of the landing patterns, we recommend the following dimensioning:

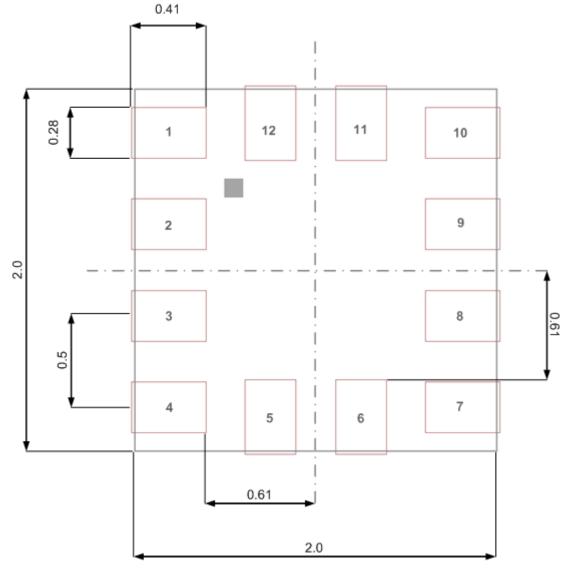


Figure 29: Landing patterns; dimensions are in mm

Same tolerances as given for the outline dimensions (Chapter 9.1, Figure 27) should be assumed.

A wiring no-go area in the top layer of the PCB below the sensor is strongly recommended (e.g. no vias, wires or other metal structures).



# 9.4 Marking

# 9.4.1 Mass production samples

Table 26: Marking of mass production samples

Labeling	Name	Symbol	Remar				
	Lot counter	CCC	3 alphanumeric digits, variable to generate mass production trace-code				
CCC	Product number	Т	1 alphanumeric digit, fixed to identify product type, T = "D"				
│	Sub-con ID	L	1 alphanumeric digit, variable to identify sub-con				
	Pin 1 identifier	•					

# 9.4.2 Engineering samples

Table 27: Marking of engineering samples

Labeling	Name	Symbol	Remark
	Eng. sample ID	N	1 alphanumeric digit, fixed to identify engineering sample, N = " * " or "e" or "E"
XXN	Sample ID	XX	2 alphanumeric digits, variable to generate trace-code.
● CC	Counter ID	СС	2 alphanumeric digits, variable to generate trace-code
	Pin 1 identifier	•	



# 9.5 Soldering guidelines

The moisture sensitivity level of the BMA280 sensors corresponds to JEDEC Level 1, see also

- IPC/JEDEC J-STD-020C "Joint Industry Standard: Moisture/Reflow Sensitivity Classification for non-hermetic Solid State Surface Mount Devices"
- IPC/JEDEC J-STD-033A "Joint Industry Standard: Handling, Packing, Shipping and Use of Moisture/Reflow Sensitive Surface Mount Devices"

The sensor fulfils the lead-free soldering requirements of the above-mentioned IPC/JEDEC standard, i.e. reflow soldering with a peak temperature up to 260°C.

Profile Feature	Pb-Free Assembly
Average Ramp-Up Rate (Ts <sub>max</sub> to Tp)	3° C/second max.
Preheat  - Temperature Min (Ts <sub>min</sub> )  - Temperature Max (Ts <sub>max</sub> )  - Time (ts <sub>min</sub> to ts <sub>max</sub> )	150 °C 200 °C 60-180 seconds
Time maintained above:  - Temperature (T <sub>L</sub> )  - Time (t <sub>L</sub> )	217 °C 60-150 seconds
Peak/Classification Temperature (Tp)	260 °C
Time within 5 °C of actual Peak Temperature (tp)	20-40 seconds
Ramp-Down Rate	6 *C/second max.
Time 25 °C to Peak Temperature	8 minutes max.

Note 1: All temperatures refer to topside of the package, measured on the package body surface.

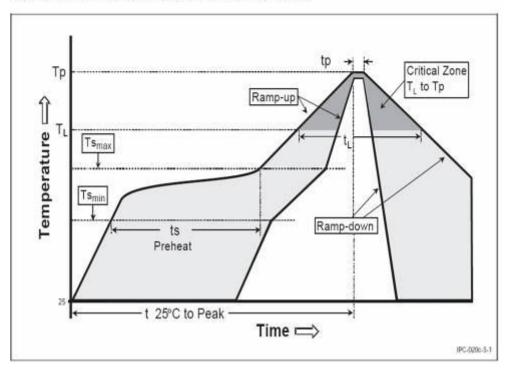


Figure 30: Soldering profile



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# 9.6 Handling instructions

Micromechanical sensors are designed to sense acceleration with high accuracy even at low amplitudes and contain highly sensitive structures inside the sensor element. The MEMS sensor can tolerate mechanical shocks up to several thousand g's. However, these limits might be exceeded in conditions with extreme shock loads such as e.g. hammer blow on or next to the sensor, dropping of the sensor onto hard surfaces etc.

We recommend to avoid g-forces beyond the specified limits during transport, handling and mounting of the sensors in a defined and qualified installation process.

This device has built-in protections against high electrostatic discharges or electric fields (e.g. 2kV HBM); however, anti-static precautions should be taken as for any other CMOS component. Unless otherwise specified, proper operation can only occur when all terminal voltages are kept within the supply voltage range. Unused inputs must always be tied to a defined logic voltage level.



# 9.7 Tape and reel specification

The BMA280 is shipped in a standard cardboard box. The box dimension for 1 reel is:  $L \times W \times H = 35 \text{cm} \times 35 \text{cm} \times 6 \text{cm}$ . BMA280 quantity: 10,000pcs per reel, please handle with care.

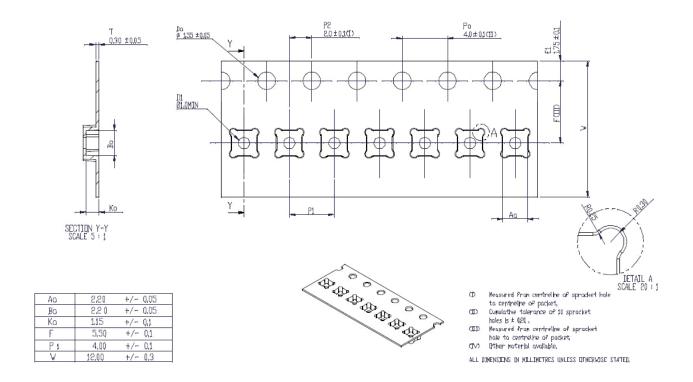


Figure 31: Tape and reel dimensions in mm



# 9.7.1 Orientation within the reel

# → Processing direction → TIL DDD

Figure 32: Orientation of the BMA280 devices relative to the tape



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# 9.8 Environmental safety

The BMA280 sensor meets the requirements of the EC restriction of hazardous substances (RoHS) directive, see also:

Directive 2002/95/EC of the European Parliament and of the Council of 8 September 2011 on the restriction of the use of certain hazardous substances in electrical and electronic equipment.

#### 9.8.1 Halogen content

The BMA280 is halogen-free. For more details on the analysis results please contact your Bosch Sensortec representative.

#### 9.8.2 Internal package structure

Within the scope of Bosch Sensortec's ambition to improve its products and secure the mass product supply, Bosch Sensortec qualifies additional sources (e.g. 2<sup>nd</sup> source) for the LGA package of the BMA280.

While Bosch Sensortec took care that all of the technical packages parameters are described above are 100% identical for all sources, there can be differences in the chemical content and the internal structural between the different package sources.

However, as secured by the extensive product qualification process of Bosch Sensortec, this has no impact to the usage or to the quality of the BMA280 product.

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# 10. Legal disclaimer

#### 10.1 Engineering samples

Engineering Samples are marked with an asterisk (\*) or (e). Samples may vary from the valid technical specifications of the product series contained in this data sheet. They are therefore not intended or fit for resale to third parties or for use in end products. Their sole purpose is internal client testing. The testing of an engineering sample may in no way replace the testing of a product series. Bosch Sensortec assumes no liability for the use of engineering samples. The Purchaser shall indemnify Bosch Sensortec from all claims arising from the use of engineering samples.

#### 10.2 Product use

Bosch Sensortec products are developed for the consumer goods industry. They may only be used within the parameters of this product data sheet. They are not fit for use in life-sustaining or safety-critical systems. Safety-critical systems are those for which a malfunction is expected to lead to bodily harm, death or severe property damage. In addition, they shall not be used directly or indirectly for military purposes (including but not limited to nuclear, chemical or biological proliferation of weapons or development of missile technology), nuclear power, deep sea or space applications (including but not limited to satellite technology).

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The purchaser accepts the responsibility to monitor the market for the purchased products, particularly with regard to product safety, and to inform Bosch Sensortec without delay of all safety-critical incidents.

#### 10.3 Application examples and hints

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# 11. Document history and modification

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