

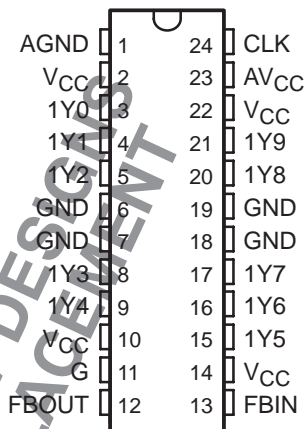
# CDC2510C

## 3.3-V PHASE-LOCK LOOP CLOCK DRIVER

SCAS621A – DECEMBER 1998 – REVISED DECEMBER 2004

- Use **CDCVF2510A** as a Replacement for this Device
- Designed to Meet PC SDRAM Registered DIMM Design Support Document Rev. 1.2
- Spread Spectrum Clock Compatible
- Operating Frequency 25 MHz to 125 MHz
- Static tPhase Error Distribution at 66 MHz to 100 MHz is  $\pm 150$  ps
- Drop-In Replacement for TI CDC2510A With Enhanced Performance
- Jitter (cyc – cyc) at 66 MHz to 100 MHz is |100 ps|
- Available in Plastic 24-Pin TSSOP
- Phase-Lock Loop Clock Distribution for Synchronous DRAM Applications
- Distributes One Clock Input to One Bank of Ten Outputs
- External Feedback (FBIN) Terminal Is Used to Synchronize the Outputs to the Clock Input
- On-Chip Series Damping Resistors
- No External RC Network Required
- Operates at 3.3 V

PW PACKAGE  
(TOP VIEW)



### description

The CDC2510C is a high-performance, low-skew, low-jitter, phase-lock loop (PLL) clock driver. It uses a PLL to precisely align, in both frequency and phase, the feedback (FBOUT) output to the clock (CLK) input signal. It is specifically designed for use with synchronous DRAMs. The CDC2510C operates at  $V_{CC} = 3.3$  V. It also provides integrated series-damping resistors that make it ideal for driving point-to-point loads.

One bank of ten outputs provides ten low-skew, low-jitter copies of CLK. Output signal duty cycles are adjusted to 50 percent, independent of the duty cycle at CLK. All outputs can be enabled or disabled via a single output enable input. When the G input is high, the outputs switch in phase and frequency with CLK; when the G input is low, the outputs are disabled to the logic-low state.

Unlike many products containing PLLs, the CDC2510C does not require external RC networks. The loop filter for the PLL is included on-chip, minimizing component count, board space, and cost.

Because it is based on PLL circuitry, the CDC2510C requires a stabilization time to achieve phase lock of the feedback signal to the reference signal. This stabilization time is required, following power up and application of a fixed-frequency, fixed-phase signal at CLK, and following any changes to the PLL reference or feedback signals. The PLL can be bypassed for test purposes by strapping  $AV_{CC}$  to ground.

The CDC2510C is characterized for operation from 0°C to 85°C.

For application information, see the *High Speed Distribution Design Techniques for CDC509/516/2509/2510/2516* (literature number SLMA003) and *Using CDC2509A/2510A PLL with Spread Spectrum Clocking (SSC)* (literature number SCAA039) application reports.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

 **TEXAS  
INSTRUMENTS**

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# CDC2510C

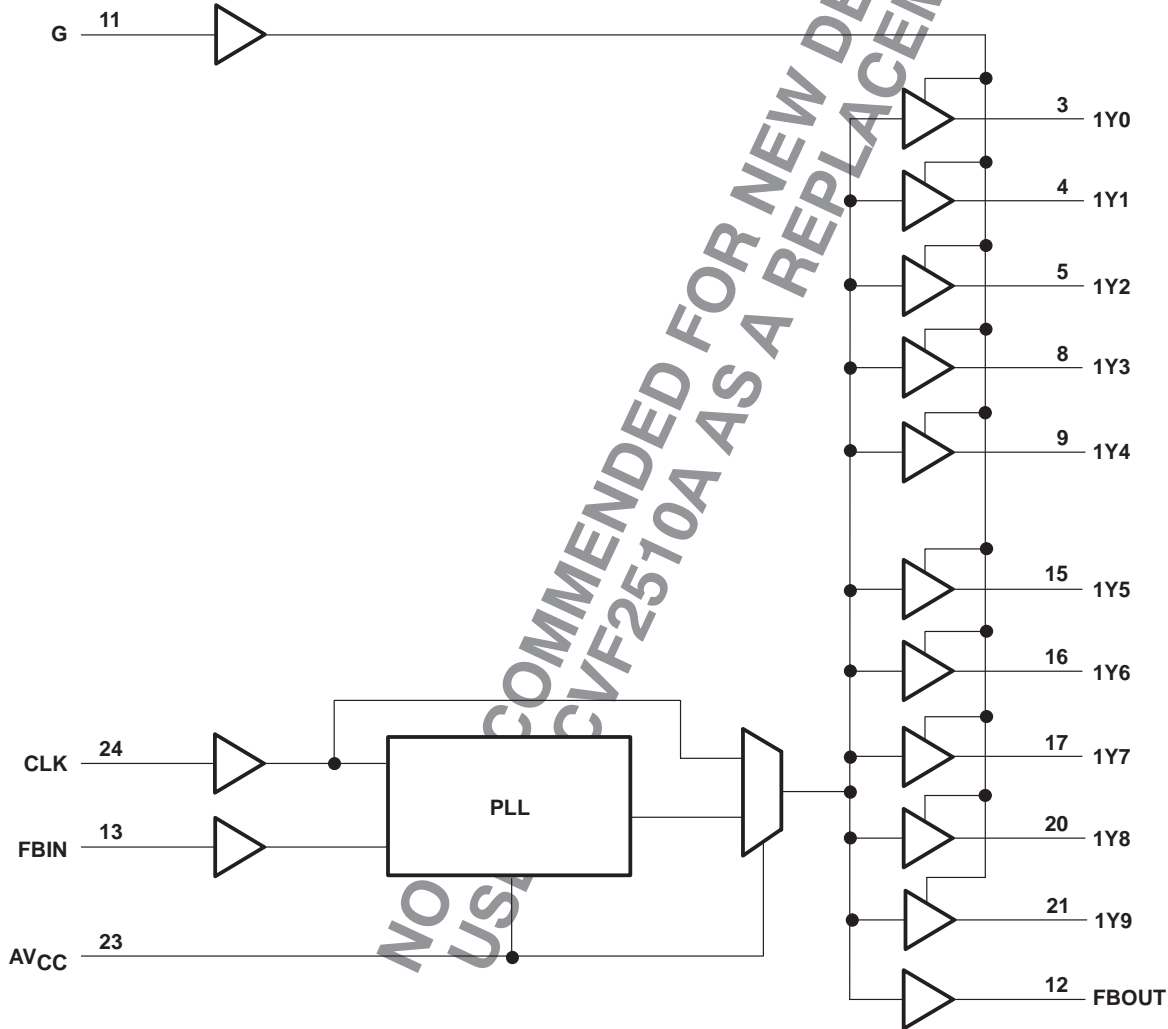
## 3.3-V PHASE-LOCK LOOP CLOCK DRIVER

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FUNCTION TABLE

INPUTS		OUTPUTS	
G	CLK	1Y (0:9)	FBOUT
X	L	L	L
L	H	L	H
H	H	H	H

functional block diagram



AVAILABLE OPTIONS

T <sub>A</sub>	PACKAGE
	SMALL OUTLINE (PW)
0°C to 85°C	CDC2510CPWR

### Terminal Functions

TERMINAL		TYPE	DESCRIPTION
NAME	NO.		
CLK	24	I	Clock input. CLK provides the clock signal to be distributed by the CDC2510C clock driver. CLK is used to provide the reference signal to the integrated PLL that generates the clock output signals. CLK must have a fixed frequency and fixed phase for the PLL to obtain phase lock. Once the circuit is powered up and a valid CLK signal is applied, a stabilization time is required for the PLL to phase lock the feedback signal to its reference signal.
FBIN	13	I	Feedback input. FBIN provides the feedback signal to the internal PLL. FBIN must be hardwired to FBOUT to complete the PLL. The integrated PLL synchronizes CLK and FBIN so that there is nominally zero phase error between CLK and FBIN.
G	11	I	Output bank enable. G is the output enable for outputs 1Y(0:9). When G is low, outputs 1Y(0:9) are disabled to a logic-low state. When G is high, all outputs 1Y(0:9) are enabled and switch at the same frequency as CLK.
FBOUT	12	O	Feedback output. FBOUT is dedicated for external feedback. It switches at the same frequency as CLK. When externally wired to FBIN, FBOUT completes the feedback loop of the PLL. FBOUT has an integrated 25-Ω series-damping resistor.
1Y(0:9)	3, 4, 5, 8, 9 15, 16, 17, 20, 21	O	Clock outputs. These outputs provide low-skew copies of CLK. Output bank 1Y(0:9) is enabled via the G input. These outputs can be disabled to a logic-low state by deasserting the G control input. Each output has an integrated 25-Ω series-damping resistor.
AV <sub>CC</sub>	23	Power	Analog power supply. AV <sub>CC</sub> provides the power reference for the analog circuitry. In addition, AV <sub>CC</sub> can be used to bypass the PLL for test purposes. When AV <sub>CC</sub> is strapped to ground, PLL is bypassed and CLK is buffered directly to the device outputs.
AGND	1	Ground	Analog ground. AGND provides the ground reference for the analog circuitry.
V <sub>CC</sub>	2, 10, 14, 22	Power	Power supply
GND	6, 7, 18, 19	Ground	Ground

NOT RECOMMENDED FOR NEW DESIGNS  
USE CDCVCF2510A AS REPLACEMENT

# CDC2510C

## 3.3-V PHASE-LOCK LOOP CLOCK DRIVER

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### absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, $AV_{CC}$ (see Note 1)	$AV_{CC} < V_{CC} + 0.7\text{ V}$
Supply voltage range, $V_{CC}$ , $AV_{CC}$	-0.5 V to 4.6 V
Input voltage range, $V_I$ (see Note 2)	-0.5 V to 6.5 V
Voltage range applied to any output in the high or low state, $V_O$ (see Notes 2 and 3)	-0.5 V to $V_{CC} + 0.5\text{ V}$
Input clamp current, $I_{IK}$ ( $V_I < 0$ )	-50 mA
Output clamp current, $I_{OK}$ ( $V_O < 0$ or $V_O > V_{CC}$ )	$\pm 50\text{ mA}$
Continuous output current, $I_O$ ( $V_O = 0$ to $V_{CC}$ )	$\pm 50\text{ mA}$
Continuous current through each $V_{CC}$ or GND	$\pm 100\text{ mA}$
Maximum power dissipation at $T_A = 55^\circ\text{C}$ (in still air) (see Note 4)	0.7 W
Storage temperature range, $T_{stg}$	$-65^\circ\text{C}$ to $150^\circ\text{C}$

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES:
- $AV_{CC}$  **must not** exceed  $V_{CC}$ .
  - The input and output negative-voltage ratings may be exceeded if the input and output clamp-current ratings are observed.
  - This value is limited to 4.6 V maximum.
  - The maximum package power dissipation is calculated using a junction temperature of  $150^\circ\text{C}$  and a board trace length of 750 mils. For more information, refer to the *Package Thermal Considerations* application note in the *ABT Advanced BiCMOS Technology Data Book*, literature number SCBD002.

### recommended operating conditions (see Note 5)

	MIN	MAX	UNIT
$V_{CC}$ , $AV_{CC}$ Supply voltage	3	3.6	V
$V_{IH}$ High-level input voltage	2		V
$V_{IL}$ Low-level input voltage		0.8	V
$V_I$ Input voltage	0	$V_{CC}$	V
$I_{OH}$ High-level output current		-12	mA
$I_{OL}$ Low-level output current		12	mA
$T_A$ Operating free-air temperature	0	85	$^\circ\text{C}$

NOTE 5: Unused inputs must be held high or low to prevent them from floating.

### timing requirements over recommended ranges of supply voltage and operating free-air temperature

	MIN	MAX	UNIT
$f_{clk}$ Clock frequency	25	125	MHz
Input clock duty cycle	40%	60%	
Stabilization time†		1	ms

† Time required for the integrated PLL circuit to obtain phase lock of its feedback signal to its reference signal. For phase lock to be obtained, a fixed-frequency, fixed-phase reference signal must be present at CLK. Until phase lock is obtained, the specifications for propagation delay, skew, and jitter parameters given in the switching characteristics table are not applicable. This parameter does not apply for input modulation under SSC application.



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**electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)**

PARAMETER		TEST CONDITIONS	AV <sub>CC</sub> , V <sub>CC</sub>	MIN	TYP <sup>‡</sup>	MAX	UNIT
V <sub>IK</sub>	Input clamp voltage	I <sub>I</sub> = -18 mA	3 V			-1.2	V
V <sub>OH</sub>	High-level output voltage	I <sub>OH</sub> = -100 μA	MIN to MAX	V <sub>CC</sub> - 0.2			V
		I <sub>OH</sub> = -12 mA	3 V	2.1			
		I <sub>OH</sub> = -6 mA	3 V	2.4			
V <sub>OL</sub>	Low-level output voltage	I <sub>OL</sub> = 100 μA	MIN to MAX			0.2	V
		I <sub>OL</sub> = 12 mA	3 V			0.8	
		I <sub>OL</sub> = 6 mA	3 V			0.55	
I <sub>OH</sub>	High-level output current	V <sub>O</sub> = 1 V	3.135 V	-32			mA
		V <sub>O</sub> = 1.65 V	3.3 V			-36	
		V <sub>O</sub> = 3.135 V	3.465 V			-12	
I <sub>OL</sub>	Low-level output current	V <sub>O</sub> = 1.95 V	3.135 V	34			mA
		V <sub>O</sub> = 1.65 V	3.3 V			40	
		V <sub>O</sub> = 0.4 V	3.465 V			14	
I <sub>I</sub>	Input current	V <sub>I</sub> = V <sub>CC</sub> or GND	3.6 V			±5	μA
I <sub>CC</sub> <sup>§</sup>	Supply current	V <sub>I</sub> = V <sub>CC</sub> or GND, I <sub>O</sub> = 0, Outputs: low or high	3.6 V			10	μA
ΔI <sub>CC</sub>	Change in supply current	One input at V <sub>CC</sub> - 0.6 V, Other inputs at V <sub>CC</sub> or GND	3.3 V to 3.6 V			500	μA
C <sub>i</sub>	Input capacitance	V <sub>I</sub> = V <sub>CC</sub> or GND	3.3 V			4	pF
C <sub>o</sub>	Output capacitance	V <sub>O</sub> = V <sub>CC</sub> or GND	3.3 V			6	pF

<sup>‡</sup> For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

<sup>§</sup> For I<sub>CC</sub> of AV<sub>CC</sub> and I<sub>CC</sub> vs Frequency (see Figures 11 and 12).

**switching characteristics over recommended ranges of supply voltage and operating free-air temperature, C<sub>L</sub> = 30 pF (see Note 6 and Figures 1 and 2)<sup>‡</sup>**

PARAMETER	FROM (INPUT)/CONDITION	TO (OUTPUT)	V <sub>CC</sub> , AV <sub>CC</sub> = 3.3 V ± 0.165 V			UNIT
			MIN	TYP	MAX	
Phase error time – static (normalized) (See Figures 3 – 8)	CLKIN <sup>†</sup> = 66 MHz to 100 MHz	FBIN <sup>†</sup>	-150		150	ps
t <sub>sk(o)</sub> Output skew time <sup>§</sup>	Any Y or FBOUT	Any Y or FBOUT			200	ps
Phase error time – jitter (see Note 7)	Clkin = 66 MHz to 100 MHz	Any Y or FBOUT	-50		50	ps
Jitter <sub>(cycle-cycle)</sub> (See Figures 9 and 10)	Clkin = 66 MHz to 100 MHz	Any Y or FBOUT			100	ps
Duty cycle	F(clkin > 60 MHz)	Any Y or FBOUT	45%		55%	
t <sub>r</sub> Rise time (See Notes 8 and 9)	V <sub>O</sub> = 1.2 V to 1.8 V, IBIS simulation	Any Y or FBOUT	2.5		1	V/ns
t <sub>f</sub> Fall time (See Notes 8 and 9)	V <sub>O</sub> = 1.2 V to 1.8 V, IBIS simulation	Any Y or FBOUT	2.5		1	V/ns

<sup>†</sup> These parameters are not production tested.

<sup>§</sup> The t<sub>sk(o)</sub> specification is only valid for equal loading of all outputs.

NOTES: 6. The specifications for parameters in this table are applicable only after any appropriate stabilization time has elapsed.

7. Calculated per PC DRAM SPEC (t<sub>phase error, static – jitter<sub>(cycle-to-cycle)</sub></sub>).

8. This is equivalent to 0.8 ns/2.5 ns and 0.8 ns/2.7 ns into standard 500 Ω/ 30 pF load for output swing of 0.4 V to 2 V.

9. 64 MB DIMM configuration according to PC SDRAM Registered DIMM Design Support Document, Figure 20 and Table 13.

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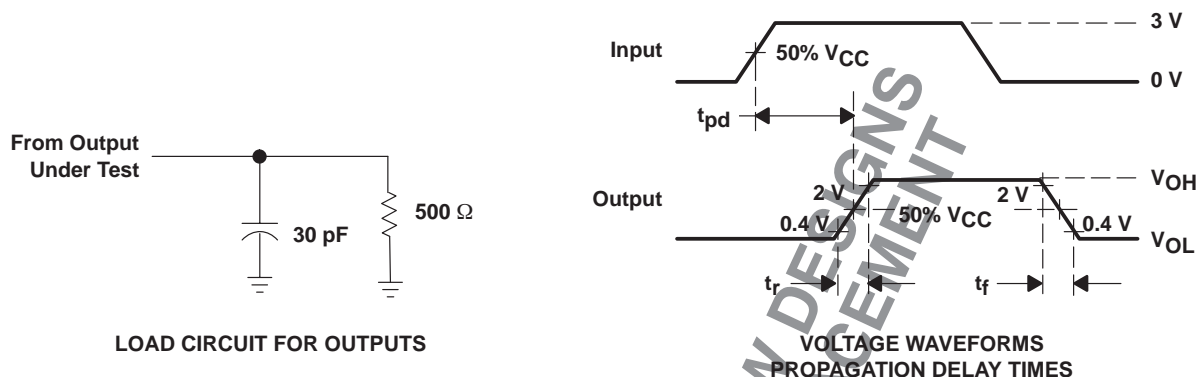


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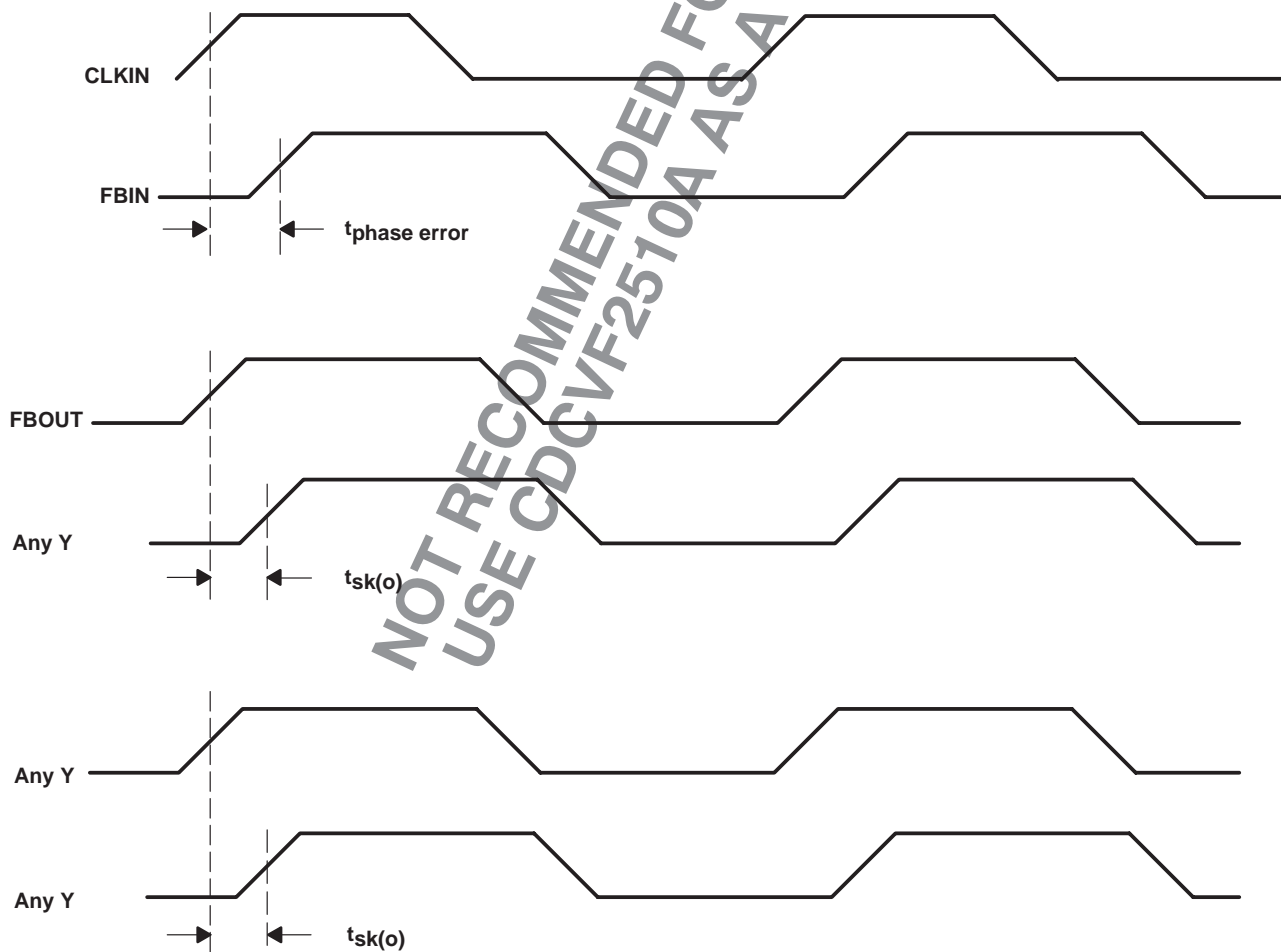
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### PARAMETER MEASUREMENT INFORMATION



- NOTES:
- A.  $C_L$  includes probe and jig capacitance.
  - B. All input pulses are supplied by generators having the following characteristics:  $PRR \leq 100$  MHz,  $Z_O = 50 \Omega$ ,  $t_r \leq 1.2$  ns,  $t_f \leq 1.2$  ns.
  - C. The outputs are measured one at a time with one transition per measurement.

Figure 1. Load Circuit and Voltage Waveforms



TYPICAL CHARACTERISTICS

CDC2510C  
PHASE ADJUSTMENT SLOPE AND PHASE ERROR  
vs  
LOAD CAPACITANCE

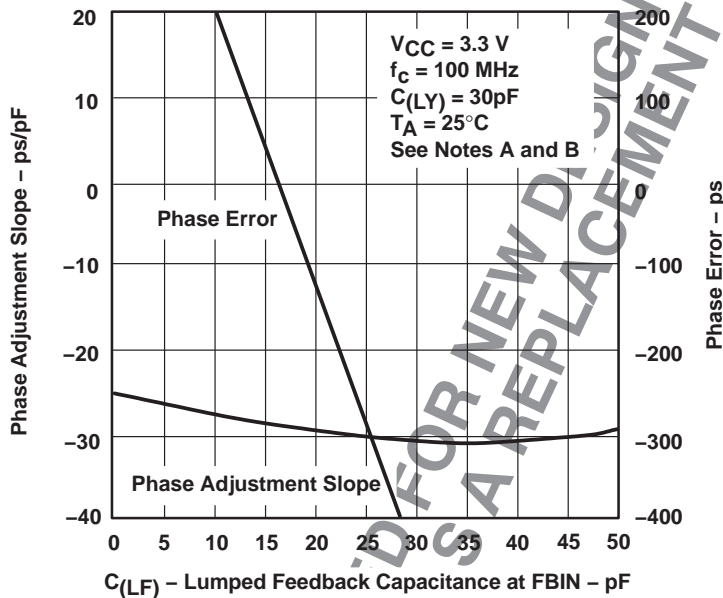


Figure 3

CDC2510A  
PHASE ADJUSTMENT SLOPE AND PHASE ERROR  
vs  
LOAD CAPACITANCE

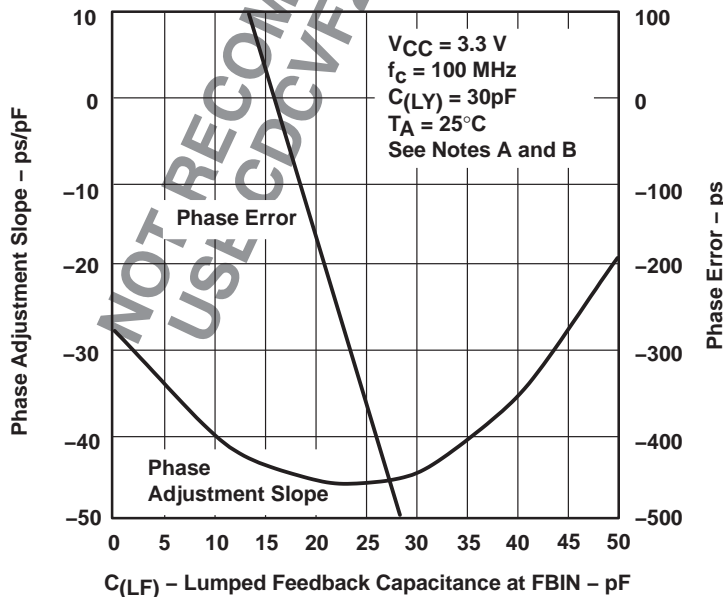


Figure 4

NOTES: A. Trace feedback length FBOUT to FBIN = 5 mm,  $Z_0 = 50\ \Omega$  Phase error measured from CLK to Y  
 B. CLF = Lumped feedback capacitance at FBIN

# CDC2510C

## 3.3-V PHASE-LOCK LOOP CLOCK DRIVER

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### TYPICAL CHARACTERISTICS

PHASE ERROR  
vs  
CLOCK FREQUENCY

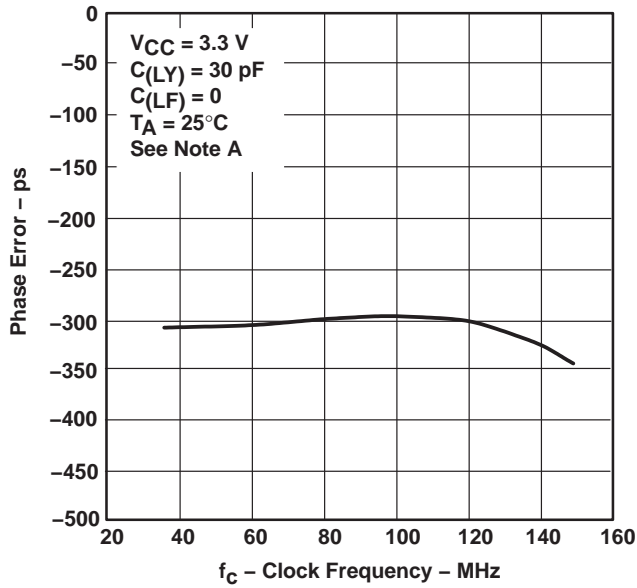


Figure 5

PHASE ERROR  
vs  
SUPPLY VOLTAGE

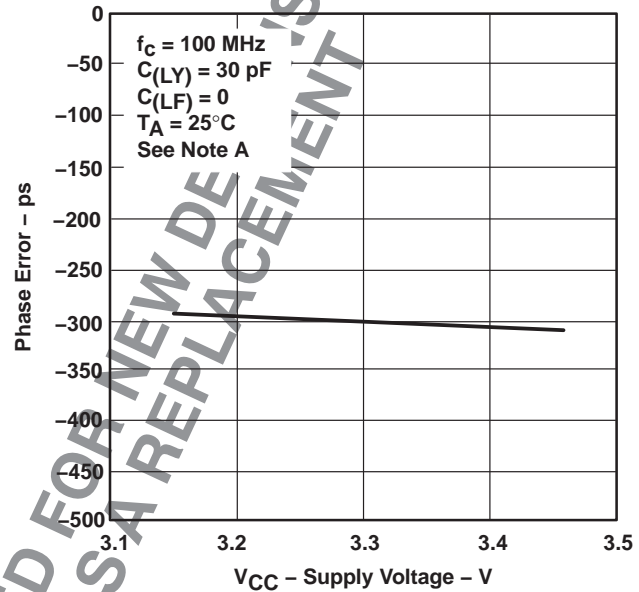


Figure 6

CDC2510C  
STATIC PHASE ERROR  
vs  
CLOCK FREQUENCY

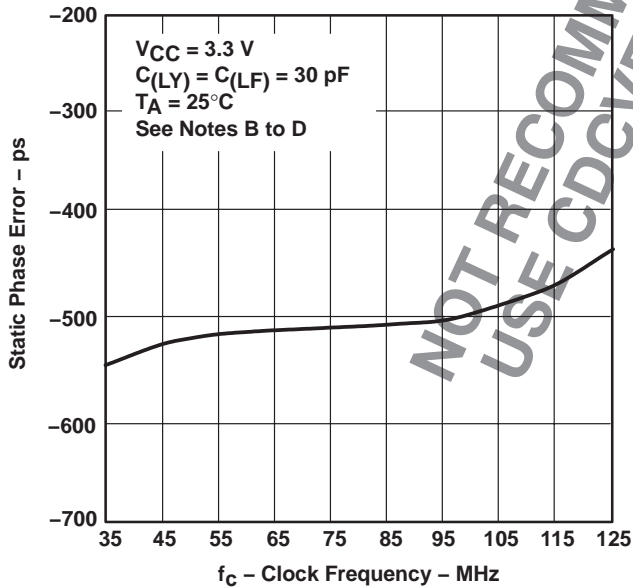


Figure 7

CDC2510A  
STATIC PHASE ERROR  
vs  
CLOCK FREQUENCY

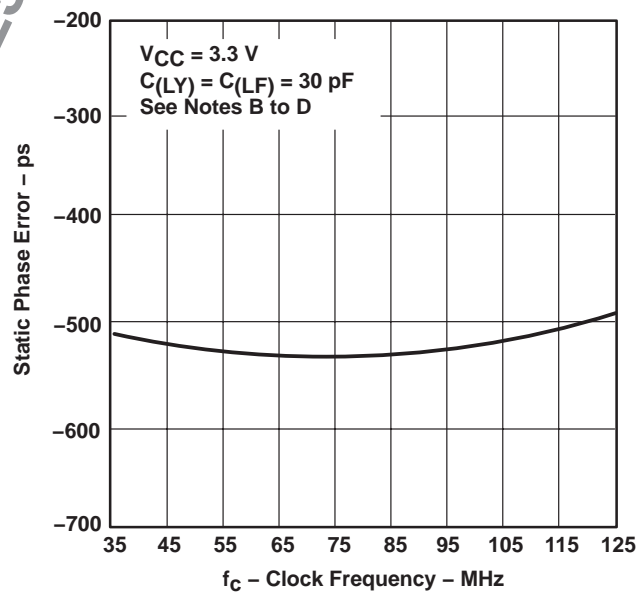
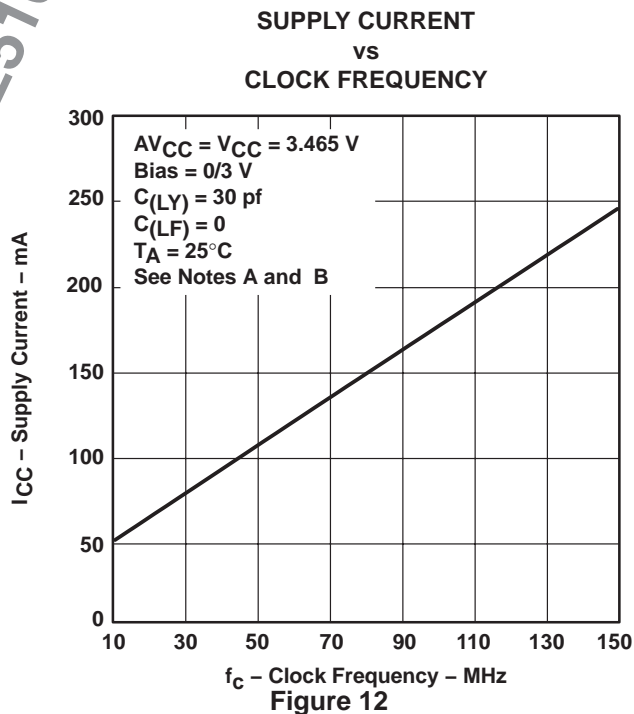
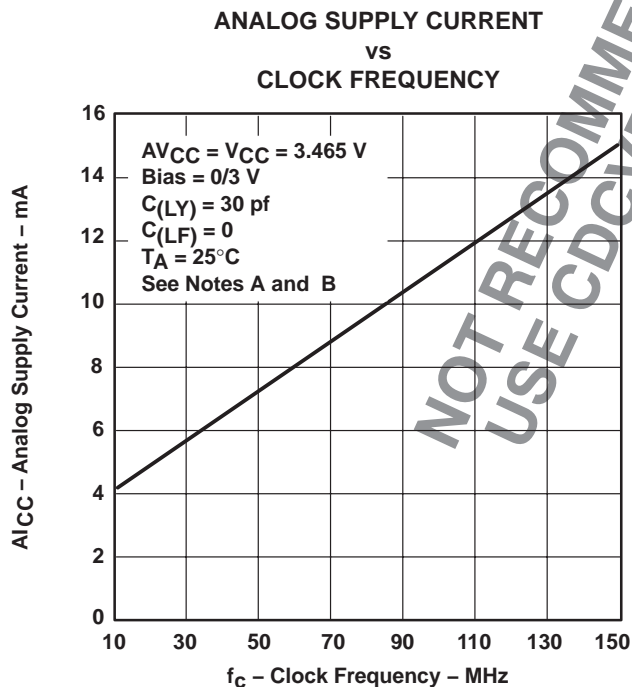
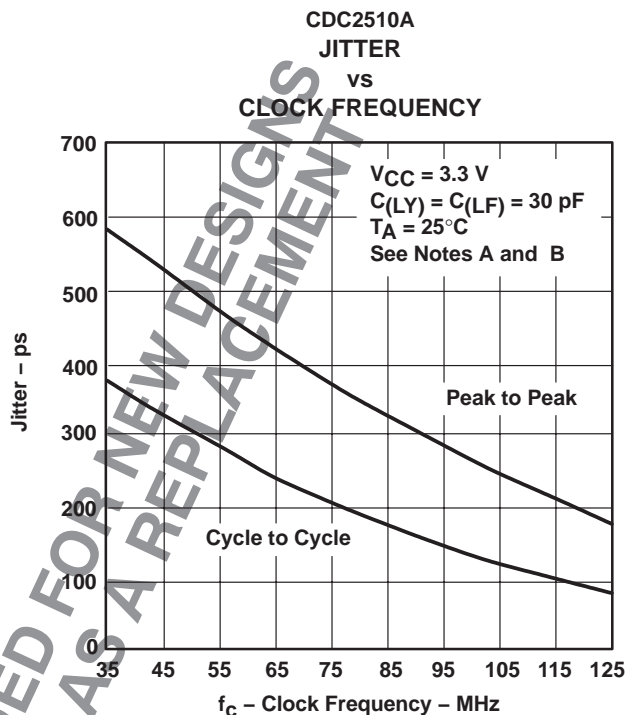
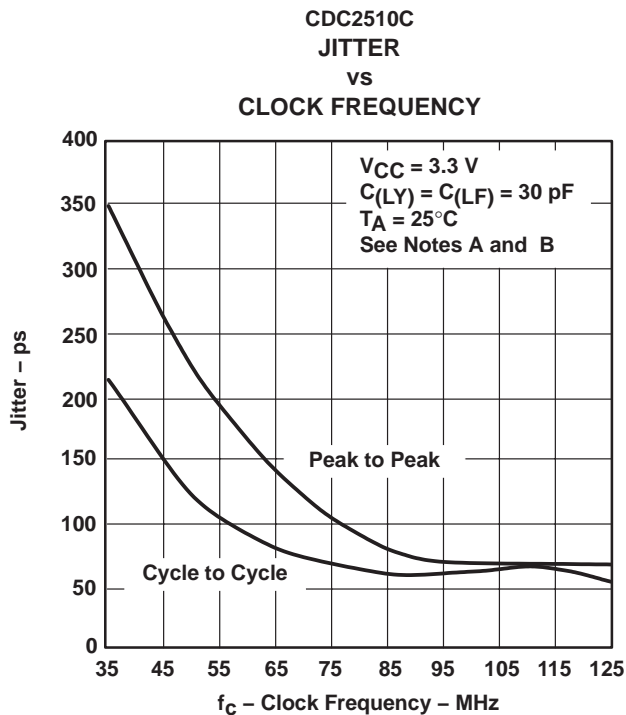


Figure 8

- NOTES: A. Trace feedback length FBOUT to FBIN = 5 mm,  $Z_O = 50\ \Omega$   
 B. Phase error measured from CLK to FBIN  
 C. CLY = Lumped capacitive load at Y  
 D. CLF = Lumped feedback capacitance at FBIN



TYPICAL CHARACTERISTICS



NOTES: A.  $C(LY)$  = Lumped capacitive load at Y  
B.  $C(LF)$  = Lumped feedback capacitance at FBIN

# CDC2510C

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### TYPICAL CHARACTERISTICS

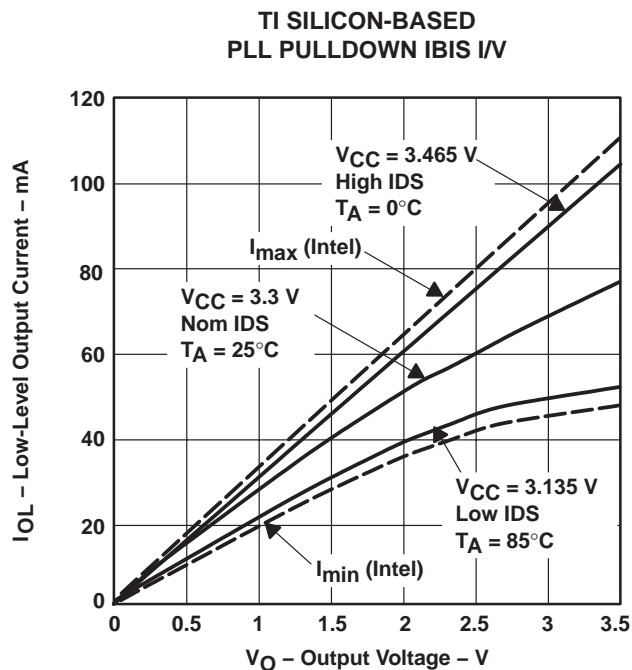


Figure 13

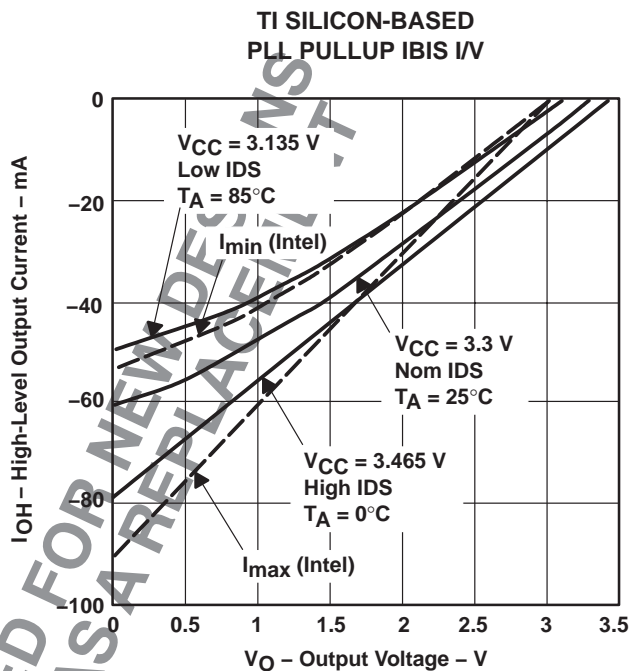


Figure 14

NOT RECOMMENDED FOR NEW DESIGNS  
USE CDCVF2510A AS A REPLACEMENT

## PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)	Op Temp (°C)	Top-Side Markings (4)	Samples
CDC2510CPW	NRND	TSSOP	PW	24	60	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM		CK2510C	
CDC2510CPWG4	NRND	TSSOP	PW	24	60	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM		CK2510C	
CDC2510CPWR	NRND	TSSOP	PW	24	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM		CK2510C	
CDC2510CPWRG4	NRND	TSSOP	PW	24	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM		CK2510C	

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) Only one of markings shown within the brackets will appear on the physical device.

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**TAPE AND REEL INFORMATION**
**REEL DIMENSIONS**

**TAPE DIMENSIONS**


A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

**TAPE AND REEL INFORMATION**

\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
CDC2510CPWR	TSSOP	PW	24	2000	330.0	16.4	6.95	8.3	1.6	8.0	16.0	Q1

**TAPE AND REEL BOX DIMENSIONS**

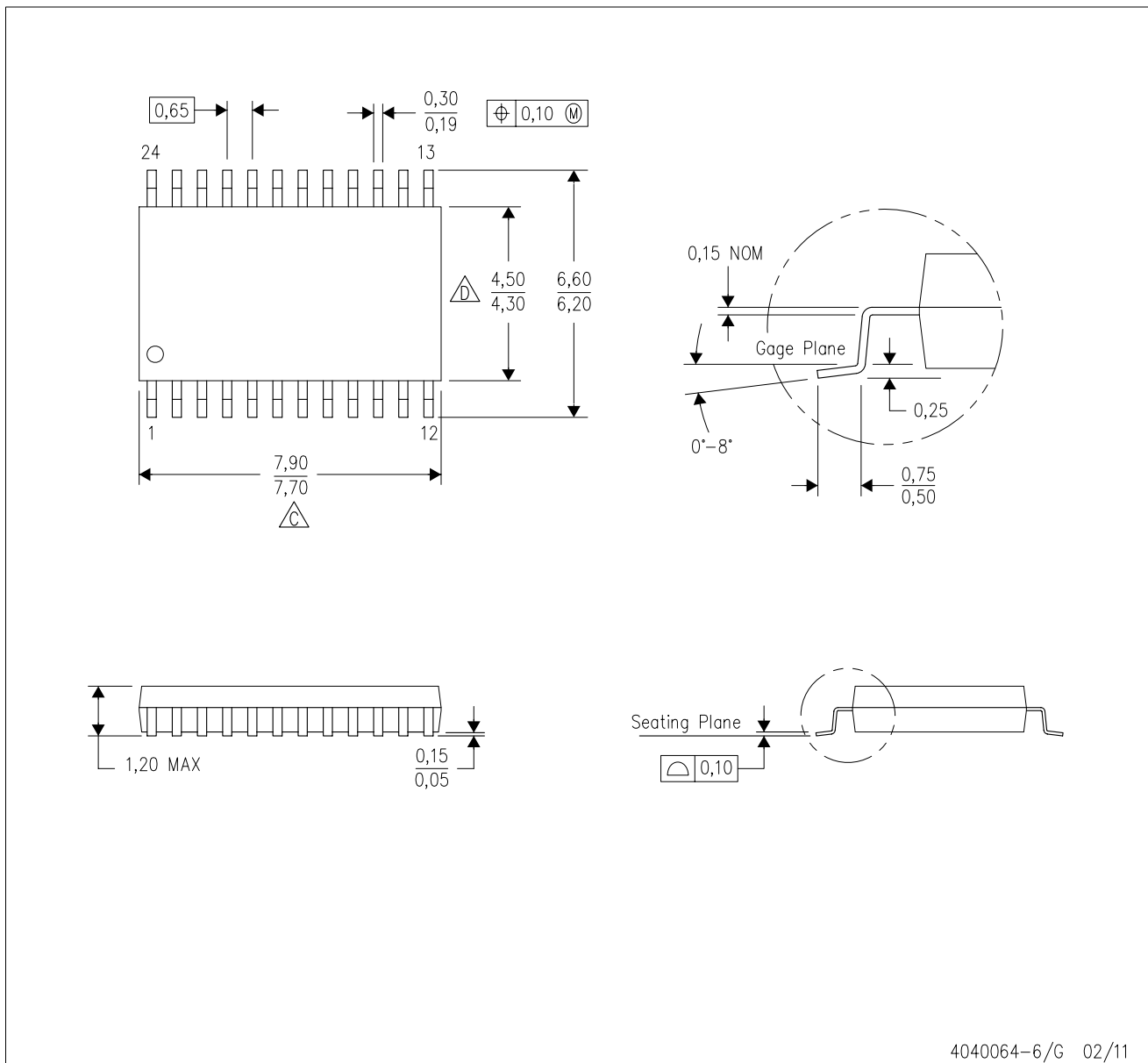


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
CDC2510CPWR	TSSOP	PW	24	2000	367.0	367.0	38.0

PW (R-PDSO-G24)

PLASTIC SMALL OUTLINE

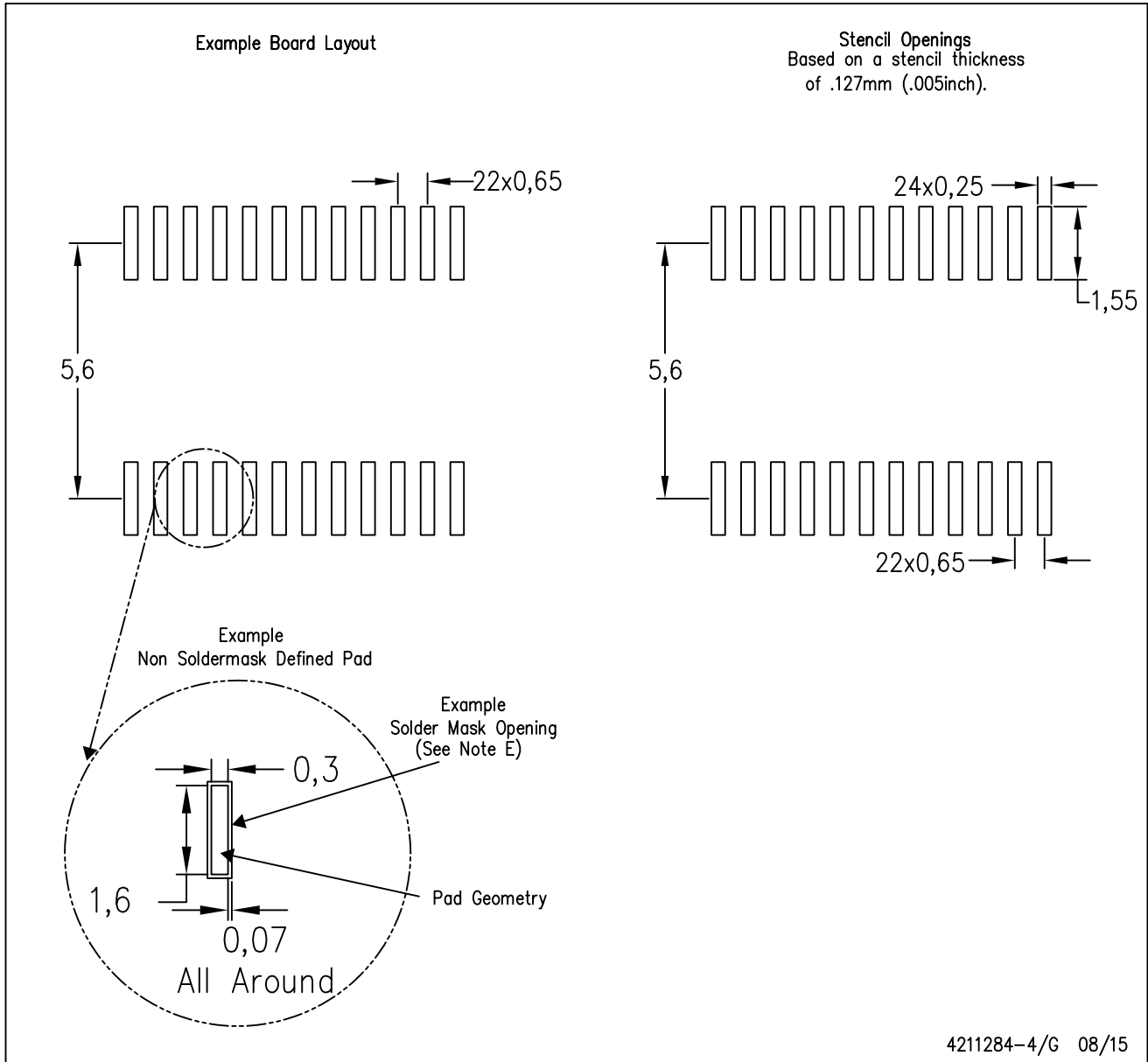


4040064-6/G 02/11

- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
  - B. This drawing is subject to change without notice.
  - C. Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 each side.
  - D. Body width does not include interlead flash. Interlead flash shall not exceed 0,25 each side.
  - E. Falls within JEDEC MO-153

PW (R-PDSO-G24)

PLASTIC SMALL OUTLINE



4211284-4/G 08/15

- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Publication IPC-7351 is recommended for alternate design.
  - D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
  - E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



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