

### Applications

- EPON FTTH modules
- Gigabit Ethernet modules
- 1G/2G Fibre Channel modules
- Digital Video

### Features

- High speed operation; suitable for SFP/SFF applications from 155Mbps to 2.0 Gbps. Typical rise/fall times < 110 ps.
- Supports Common Anode VCSEL, Common Cathode VCSEL, or FP LASER. May be used with or without a monitor photodiode.
- Programmable temperature compensation. Modulation output and bias output can be controlled using a few discrete resistors.
- Supports DDMI (SFF-8472) diagnostics.
- DC or AC coupled modulation drive.
- Peaking circuit to optimize VCSEL response.
- Low overshoot allows high extinction ratio with low jitter.
- Supports E-PON burst mode with no extra components
- Automatic Laser Power Control, with “Slow-Start”.
- 3.3V or 5V operation

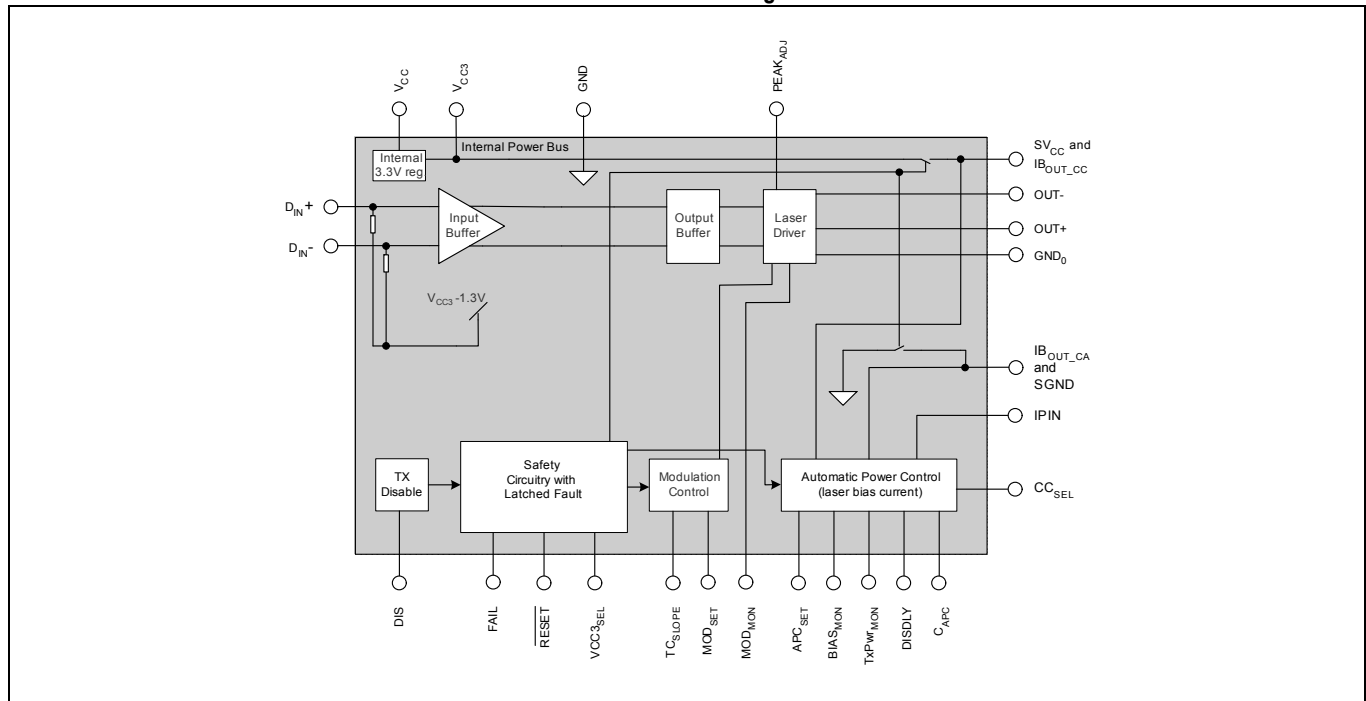
The M02094 is a highly integrated, programmable VCSEL driver intended for SFP/SFF modules to 2.0 Gbps. Using differential PECL data inputs, the M02094 supplies the bias and modulation current required to drive a VCSEL or edge-emitting laser. The modulation output can be AC or DC-coupled to a FP laser diode or AC coupled to a common anode or common cathode VCSEL.

Peaking adjustment is available to improve VCSEL fall time.

EPON burst mode operation is supported with no extra components.

Integrated safety circuitry detects faults and provides latched bias and modulation current shutdown.

Functional Block Diagram



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### Ordering Information

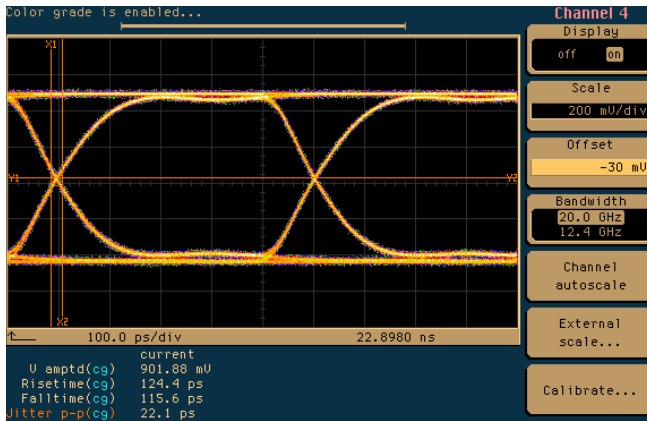
Part Number	Package	Operating Temperature
M02094-12 *	QFN24	
M02094-EVM	Combination Electrical and Optical Evaluation board	

\* The letter "G" designator after the part number indicates that the device is RoHS-compliant.

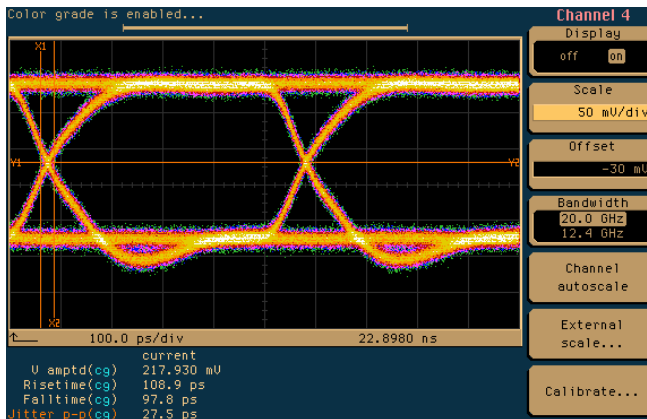
### Revision History

Revision	Level	Date	Description
V5	Release	May 2015	Updated logos and page layout. No content changes.
F (V4)	Release	July 2008	-12 part.
E (V3)	Release	February 2006	Custom part.
D (V2)	Release	September 2005	Corrected error in Pin 6 description in <a href="#">Table 2-1</a> .
C (V1)	Release	March 2005	Released. New datasheet format. Changes to Absolute Maximum Ratings, Operating Temperature, AC and DC specifications. New section on video operation.
B	Preliminary	March 2004	Preliminary release.
A	Advance	October 2003	Advance.

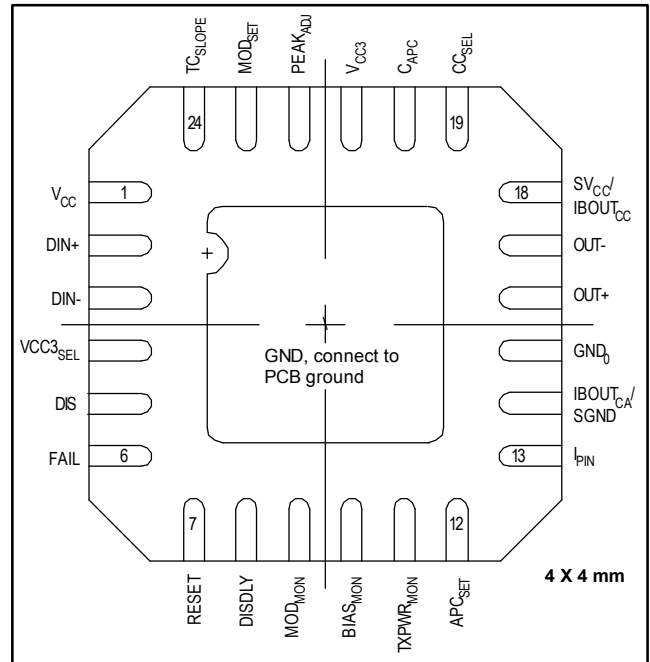
**Typical Electrical Eye Diagram, Common Anode Mode**



**Typical Electrical Eye Diagram, Common Cathode Mode with Peaking**



**Pin Configuration**



# 1.0 Product Specification

## 1.1 Absolute Maximum Ratings

These are the absolute maximum ratings at or beyond which the IC can be expected to fail or be damaged. Reliable operation at these extremes for any length of time is not implied.

**Table 1-1. Absolute Maximum Ratings**

Symbol	Parameter	Rating	Units
$V_{CC}$	Power supply voltage	-0.4 to +6.0	V
$V_{CC3}$	3.3V power supply voltage	-0.4 to +4.0	V
$T_{STG}$	Storage temperature	-65 to +150	°C
$I_{BIASOUTCA (MAX)}$	Maximum bias output current for common anode laser	75	mA
$I_{BIASOUTCC (MAX)}$	Maximum bias output current for common cathode laser	30	mA
$I_{MODCA (MAX)}$	Maximum modulation current for common anode laser	70	mA
$I_{MODCC (MAX)}$	Maximum modulation current for common cathode laser	30	mA
$D_{IN+/-}$	Data inputs	-0.4 to $V_{CC3} + 0.4$	V
DIS	Disable input	-0.4 to $V_{CC} + 0.4$	V
$BIAS_{MON}, MOD_{MON}$	Bias and modulation output current mirror compliance voltage	-0.4 to $V_{CC3} + 0.4$	V
$I_{PIN}$	Photodiode anode voltage	-0.4 to $V_{CC3} + 0.4$	V
$I_{PIN}$	Sink or Source Current	2.0	mA
FAIL	Status flag	-0.4 to $V_{CC} + 0.4$	V
$APC_{SET}, MOD_{SET}$	Set inputs	-0.4 to $V_{CC3} + 0.4$	V
$TC_{SLOPE}$	Temperature compensation slope	-0.4 to $V_{CC3} + 0.4$	V
OUT+, OUT-	Output	-0.4 to $V_{CC3} + 0.4$	V

## 1.2 Recommended Operating Conditions

**Table 1-2. Recommended Operating Conditions**

Parameter	Rating	Units
Power supply ( $V_{CC-GND}$ )	$3.3 \pm 7.5\%$ or $5.0 + 10\%, -5\%$	V
Operating ambient	-40 to +95	°C

## 1.3 DC Characteristics

( $V_{CC} = +3.05V$  to  $+3.55V$  or  $4.75V$  to  $5.5V$ ,  $T_A = -40$  °C to  $+95$  °C, unless otherwise noted)

Typical values for common anode are at  $V_{CC} = 3.3$  V,  $I_{BOUTC_A} = 20$  mA,  $I_{MOD} = 20$  mA,  $T_A = 25$  °C, unless otherwise noted.

Typical values for common cathode are at  $V_{CC} = 3.3$  V,  $I_{BOUTC_C} = 5$  mA,  $I_{MOD} = 5$  mA,  $T_A = 25$  °C, unless otherwise noted.

**Table 1-3. DC Characteristics**

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units
$I_{CC}$	Supply current excluding $I_{MOD}$ and $I_{BIAS}$	PEAK <sub>ADJ</sub> high (no peaking adjust)	-			mA
		Common anode operation <sup>(1)</sup>	-	28	48	
		Common cathode operation <sup>(1)</sup>	-	35	55	
		Additional current when PEAK <sub>ADJ</sub> is used.	-	2	15	
		Additional current when operating from a 5V supply <sup>(2)</sup>				
$I_{BOUTC_A}$	Bias current adjust range, common anode mode	$V(I_{BIASOUT}) > 0.7V$	1		50	mA
$I_{BOUTC_C}$	Bias current adjust range, common cathode mode	$V(I_{BIASOUT}) < 2.5V$	0.5		15	mA
$I_{OFF}$	Bias current with optical output disabled	DIS = high $V(I_{BOUTC_A}) = V_{CC3}$ for common anode operation. $V(I_{BOUTC_C}) = 0V$ for common cathode operation.	-	-	150	µA
	Ratio of $I_{BIAS}$ current to $I_{BIASMON}$ current	CC <sub>SEL</sub> low, common anode mode CC <sub>SEL</sub> high, common cathode mode	-	45.7 13.5	-	A/A

**Table 1-3. DC Characteristics**

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units
$V_{MD}$	Monitor diode reverse bias voltage		1.5	–	–	V
$I_{MD}$	Monitor diode current adjustment range	Adjusted with $R_{APCSET}$	10	–	1500	$\mu A$
	Ratio of $TxPwr_{MON}$ current to monitor photodiode current			1		A/A
$C_{MD\_MAX}$	Maximum monitor photodiode capacitance for APC loop stability. Includes all associated parasitic capacitances.				100	pF
	TTL/CMOS input high voltage (DIS)		2.0	–	5.5	V
	TTL/CMOS input low voltage (DIS)		–	–	0.8	V
	CMOS input high voltage ( $V_{CC3SEL}$ )			2.4		V
	CMOS input low voltage ( $V_{CC3SEL}$ )			1.2		V
	Logic output high voltage (FAIL)	With external 10k $\Omega$ pull-up to $V_{CC}$ .	$V_{CC} - 0.6$	–	–	V
	Logic output low voltage (FAIL)	For 6.8k to 10k ohm resistor when pulled up to 5V For 4.7k to 10k ohm resistor when pulled up to 3.3V	–	–	0.4	V
$R_{IN}$	Differential input impedance	Data inputs	–	6800	–	$\Omega$
$V_{SELF}$	Self-biased common-mode input voltage		–	$V_{CC3} - 1.3$	–	V
$V_{INCM}$	Common-mode input compliance voltage	Data inputs	$V_{CC3} - 1.45$	–	$V_{CC3} - [V_{IN(DIFF)}]/4$	V
$V_{IN(DIFF)}$	Differential input voltage	$= 2 \times (D_{IN+HIGH} - D_{IN+LOW})$	200	–	2400	mVpp
$V_{CC3THL}^{(3)}$	3.3V supply detection, lower threshold		2.5	2.8	2.95	V
$V_{CC3THH}^{(3)}$	3.3V supply detection, upper threshold		3.65	3.8	4.1	V
$V_{CC5THL}$	5V supply detection, lower threshold		3.9	4.25	4.65	V
$V_{CC5THH}$	5V supply detection, upper threshold		5.45	5.8	6.2	V
$V_{REF1}$	Reference voltage for $MOD_{SET}$		1.18	1.3	1.4	V
$V_{APCSET}$	Reference voltage for $APC_{SET}$			1.35		V

**Table 1-3. DC Characteristics**

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units
V <sub>BL</sub>	Bias_OK lower voltage threshold		0.88	1.0	1.05	V
V <sub>BH</sub>	Bias_OK upper voltage threshold		1.45	1.6	1.7	V
V <sub>FAULTL</sub> <sup>(4)</sup>	Lower voltage threshold for fault inputs IBOUT <sub>CA</sub> , OUT+, C <sub>APC</sub> , and MOD <sub>SET</sub>	FAIL asserts if any of these signals fall below this value.		300	400	mV
V <sub>FAULTH</sub> <sup>(5)</sup>	Upper voltage threshold for fault inputs IBOUT <sub>CC</sub>	FAIL asserts if any of these signals fall above this value.	2.6	2.9	V <sub>CC3</sub> - 0.2	V
V <sub>SELFL</sub>	Self bias voltage for IBOUT <sub>CA</sub> and OUT+	During disable state	0.5	1.65	2.0	V
V <sub>SELFH</sub>	Self bias voltage for IBOUT <sub>CC</sub>	During disable state		0	1.25	V

Notes:

1. Excludes bias and modulation currents.
2. Bias and modulation currents add directly to power supply current in 5V applications. The additional supply current noted excludes these currents.
3. V<sub>CC3</sub> "supply OK" circuitry monitors the internally regulated 3.3V supply when V<sub>CC</sub> = 5V (VCC3<sub>SEL</sub> =low). When V<sub>CC</sub> = 3.3V, V<sub>CC3</sub> "supply OK" circuitry monitors V<sub>CC</sub> (VCC3<sub>SEL</sub> =high).
4. A low level at IBOUT<sub>CA</sub> does not trigger a fault condition during common cathode operation.
5. A low level at IBOUT<sub>CC</sub> does not trigger a fault condition during common anode operation.

## 1.4 AC Characteristics

( $V_{CC} = 3.05\text{ V to }3.55\text{ V or }4.7\text{ V to }5.5\text{ V}$ ,  $T_A = -40\text{ }^\circ\text{C to }+95\text{ }^\circ\text{C}$ , unless otherwise noted)

Typical values for common anode are at  $V_{CC} = 3.3\text{ V}$ ,  $I_{BOUTC A} = 20\text{ mA}$ ,  $I_{MOD} = 20\text{ mA}$ ,  $T_A = 25\text{ }^\circ\text{C}$ , unless otherwise noted.

Typical values for common cathode are at  $V_{CC} = 3.3\text{ V}$ ,  $I_{BOUTC C} = 5\text{ mA}$ ,  $I_{MOD} = 5\text{ mA}$ ,  $T_A = 25\text{ }^\circ\text{C}$ , unless otherwise noted.

**Table 1-4. AC Characteristics**

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units.
$I_{MOD}$	Modulation current adjust range	Common Cathode Mode OUT+ and OUT- AC coupled into 50 $\Omega$ load.	1	–	15	mA
		Common Anode Mode OUT+ and OUT- DC coupled into 25 $\Omega$ load. <sup>(1)</sup>	3		45	
$I_{MOD(OFF)}$ <sup>(2)</sup>	Modulation current with output disabled	DIS = high	–	–	150	$\mu\text{A}$
	Ratio of modulation current to $MOD_{MON}$ current	$CC_{SEL} = \text{high}$ , $R_{LOAD} = 50\Omega$	–	30	–	A/A
		$CC_{SEL} = \text{low}$		65		
$I_{MOD-TC}$	Programmable range for modulation current temperature coefficient	Adjustable using $TC_{SLOPE}$	0	–	$10^4$	ppm/ $^\circ\text{C}$
$T_{TCSTART}$	Temperature at which modulation current TC compensation enables			20		$^\circ\text{C}$
$t_r$	Modulation output rise time	20% to 80% into 25 $\Omega$ <sup>(3)</sup> . For $I_{MOD}$ from 3 mA to 30 mA. Measured using 11110000 pattern at 2.5 Gbps	–	110	140	ps
$t_f$	Modulation output fall time		–	110	140	ps
$OS_{OFF}$	Overshoot of modulation output current in the "off" direction	into 25 $\Omega$ load	--	1	2	%
RJ	Random jitter		–	1.2	–	$\text{ps}_{rms}$
DJ	Deterministic jitter	Measured into 25 $\Omega$ load, $2^{31} - 1$ PRBS at 2.0 Gbps		20	45	$\text{ps}_{pp}$
		K28.5 pattern at 2.0 Gbps		20	45	
		(includes pulse width distortion)				

Notes:

1. Minimum voltage at OUT+ > 0.7V.
2. The current through the laser in this state can be made negligible by adding a 1k $\Omega$  or less resistor in parallel with the laser.
3. The M02094 is designed to drive 25 $\Omega$  loads. External resistance should be added in series or parallel to the Laser to create this load impedance. In common cathode mode, 50 $\Omega$  resistors internal to the M02094 are in parallel with the laser.



## 1.5 Safety Logic Timing

( $V_{CC} = 3.05\text{ V to }3.55\text{ V or }4.7\text{ V to }5.5\text{ V}$ ,  $T_A = -40\text{ }^\circ\text{C to }+95\text{ }^\circ\text{C}$ , unless otherwise noted)

Typical values for common anode are at  $V_{CC} = 3.3\text{ V}$ ,  $I_{BOUT_{CA}} = 20\text{ mA}$ ,  $I_{MOD} = 20\text{ mA}$ ,  $T_A = 25\text{ }^\circ\text{C}$ , unless otherwise noted.

Typical values for common cathode are at  $V_{CC} = 3.3\text{ V}$ ,  $I_{BOUT_{CC}} = 5\text{ mA}$ ,  $I_{MOD} = 5\text{ mA}$ ,  $T_A = 25\text{ }^\circ\text{C}$ , unless otherwise noted.

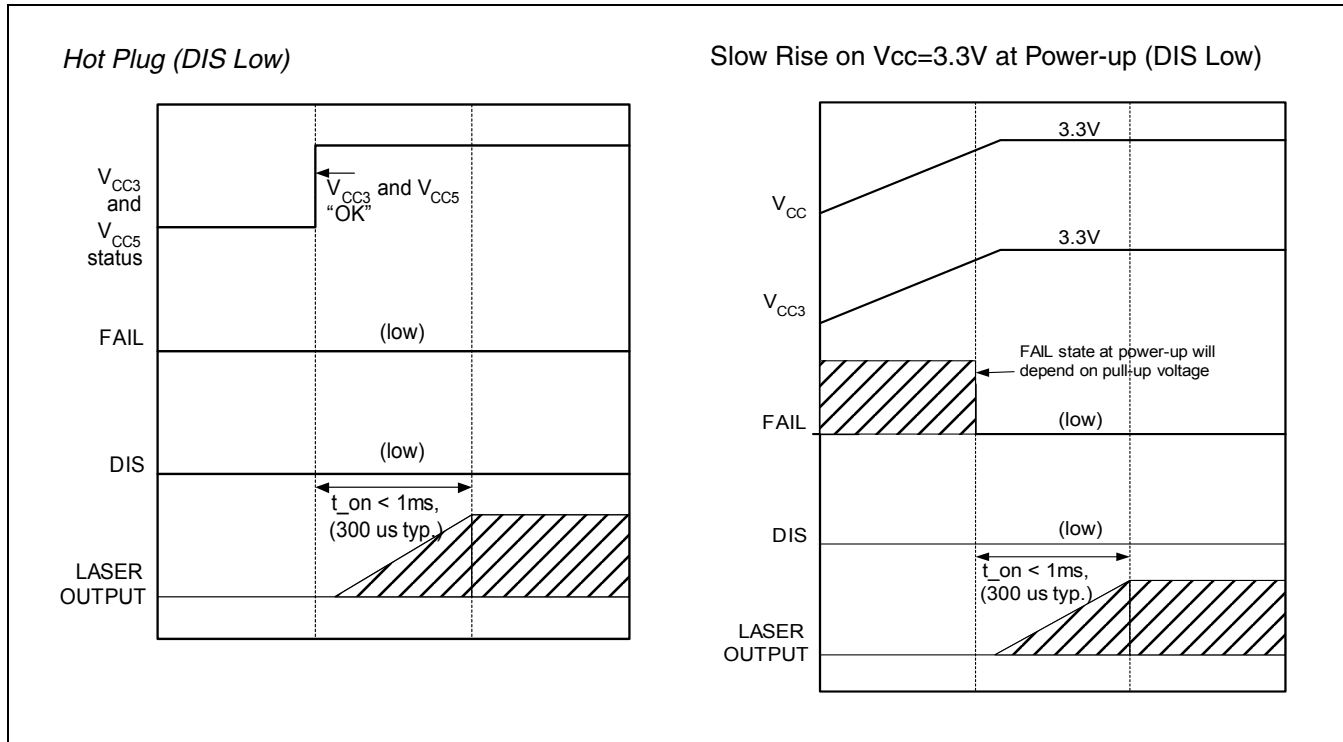
**Table 1-5. Safety Logic Timing**

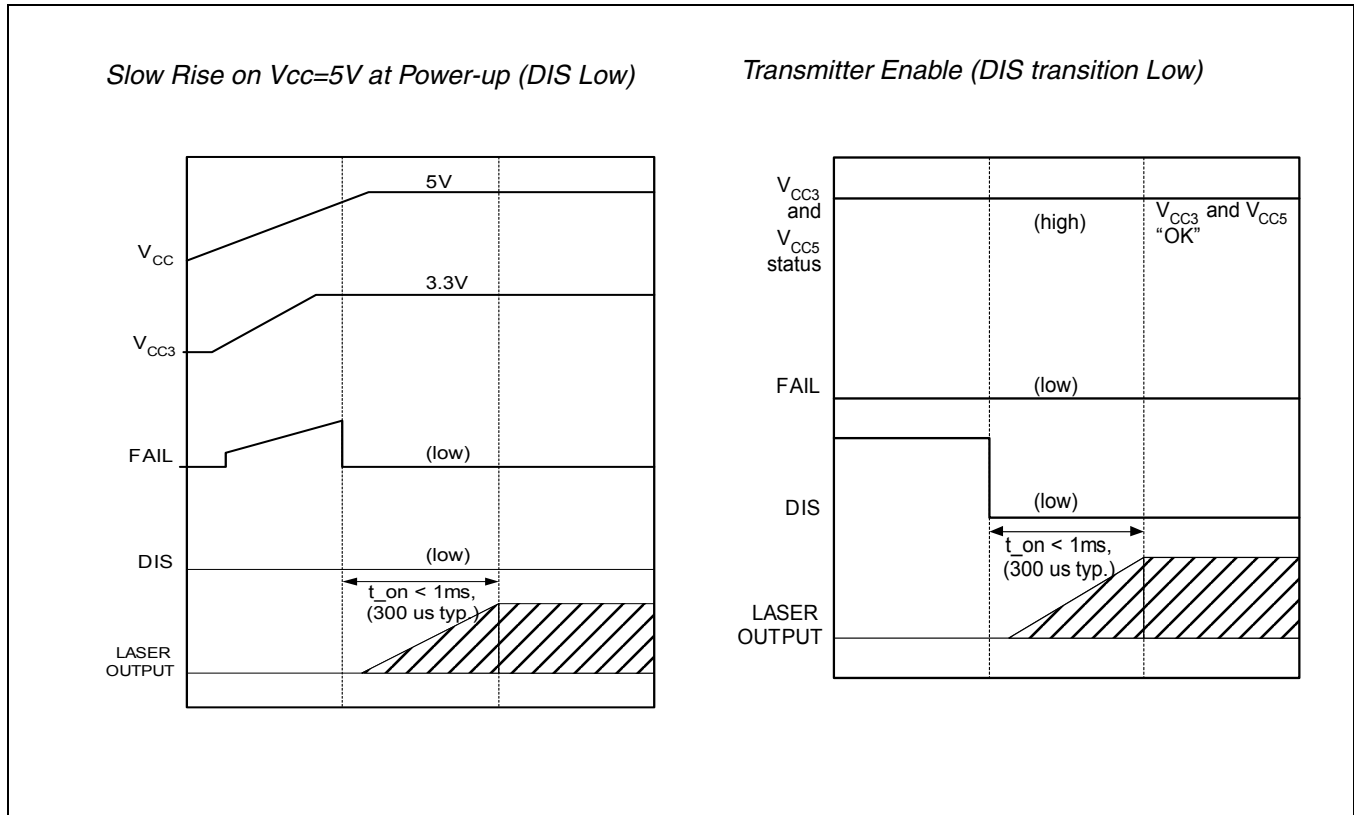
Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units.
t <sub>off</sub>	DIS assert time	Rising edge of DIS to fall of output signal below 10% of nominal <sup>(1)</sup>			10	μs
t <sub>on</sub>	DIS negate time	Falling edge of DIS to rise of output signal above 90% of nominal <sup>(1)</sup>			1	ms
t <sub>init</sub>	Time to initialize <sup>(2)</sup>	Includes reset of FAIL; from power on after Supply_OK or from negation of DIS during reset of FAIL condition	2	3	5	ms
t <sub>fault</sub>	Laser fault time - from fault condition to assertion of FAIL	From occurrence of fault condition or when Supply_OK is beyond specified range			100	μs
t <sub>reset</sub>	DIS time to start reset	DIS or ENA pulse width required to initialize safety circuitry or reset a latched fault			10 <sup>(3)</sup>	μs
t <sub>on<sub>BM</sub></sub>	DIS negate (turn-on) time during burst-mode operation	Using integrated switch at SVCC (3.3V operation) <sup>(4)</sup>		300	500	ns
t <sub>off<sub>BM</sub></sub>	DIS assert (turn-off) time during burst-mode operation	Using integrated switch at SVCC (3.3V operation) <sup>(4)</sup>		200	500	ns

Notes:

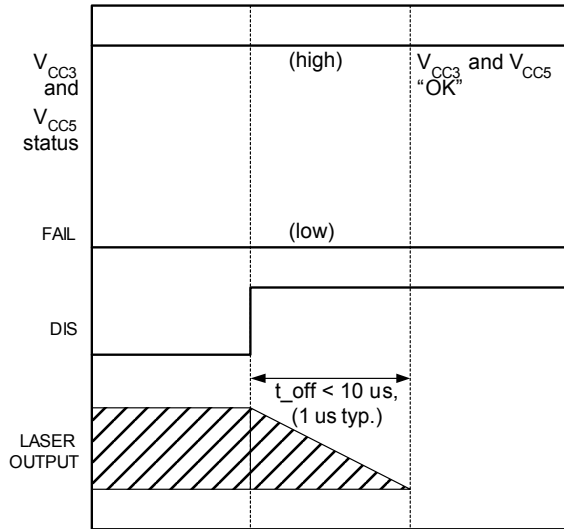
1. With CAPC = 2.2nF
2. User-adjustable. Specifications reflect timing with no external RESET capacitor.
3. With  $\leq 1\text{ nF}$  capacitor from RESET pin to ground.
4.  $I_{mod} > 12\text{ mA}$

**Figure 1-1. Safety Logic Timing Characteristics**

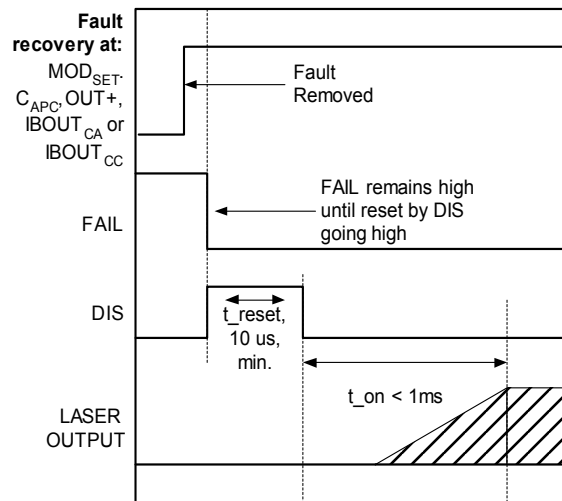




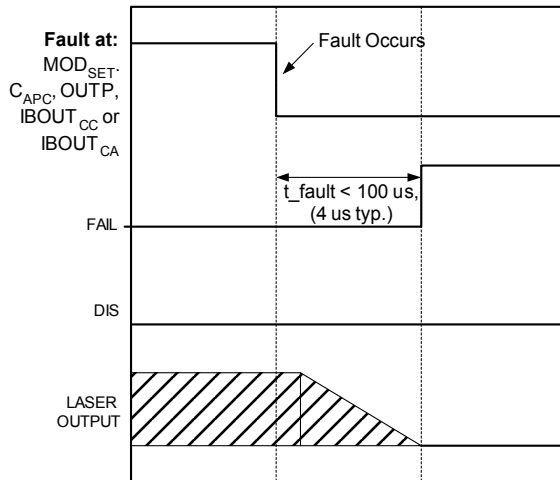
*Transmitter Disable (DIS transition high)*



*Fault Recovery Behaviour*



*Response to Fault*



*Unsuccessful Fault Reset Attempt*

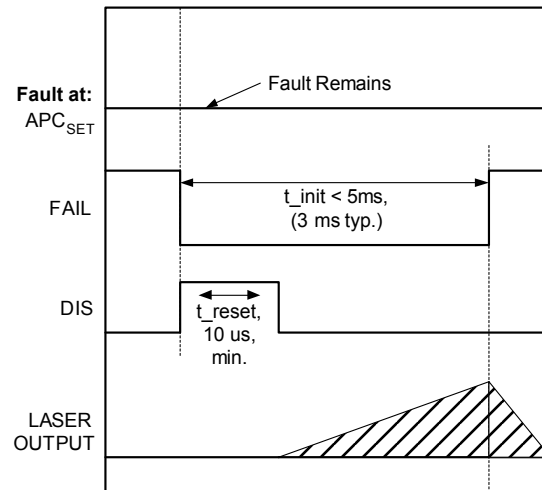
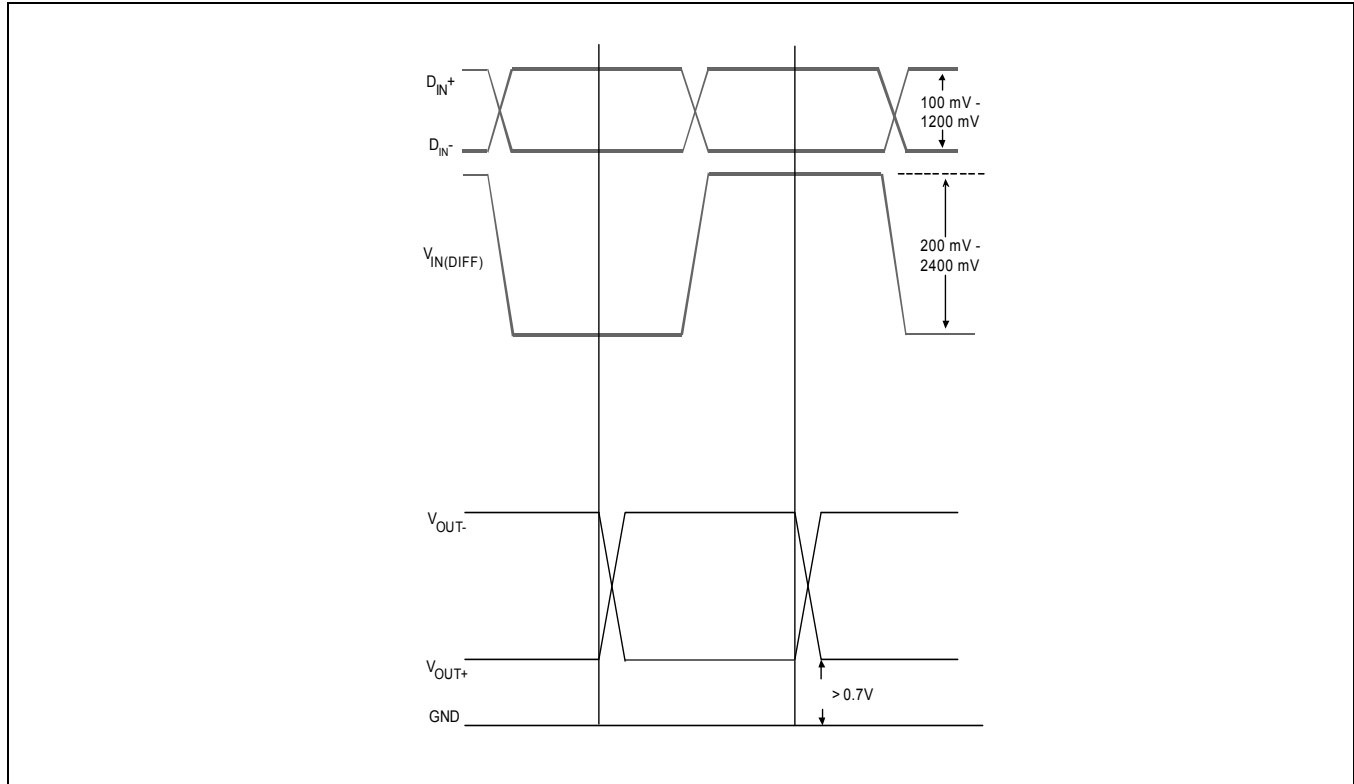


Figure 1-2. Relationship Between Data Inputs and Modulation Outputs



## 2.0 Pin Description

### 2.1 Pin Definitions

**Table 2-1. Pin Definitions**

Pin Number	Pin Name	Pin equivalent load	Function
1	V <sub>CC</sub>		Power supply, 5V or 3.3V.
2	D <sub>IN+</sub>		<p>Positive data input. Self biased. Compatible with AC coupled PECL, AC coupled CML, and DC-coupled PECL (when V<sub>CC</sub> = 3.3V).                      When D<sub>IN+</sub> is high, OUT+ will sink current.</p>
3	D <sub>IN-</sub>	See D <sub>IN+</sub> drawing	Negative data input. Self biased. Compatible with AC coupled PECL, AC coupled CML, and DC-coupled PECL (when V <sub>CC</sub> = 3.3V).

**Table 2-1. Pin Definitions**

Pin Number	Pin Name	Pin equivalent load	Function
4	VCC3 <sub>SEL</sub>		<p>3.3V V<sub>CC</sub> Select.            Connect to V<sub>CC</sub> for V<sub>CC</sub> = 3.3V operation.            Connect to GND for V<sub>CC</sub> = 5V operation.</p>
5	DIS		<p>Bias and modulation output disable (TTL/CMOS).            When high or left floating, the bias and modulation outputs are disabled. Set low for normal operation.</p>

**Table 2-1. Pin Definitions**

Pin Number	Pin Name	Pin equivalent load	Function
6	FAIL		<p>Safety circuit fault output (TTL/CMOS). Goes high when a safety logic fault is detected. This output will be low when DIS is high.</p> <p>Requires an external pull-up.</p>
7	RESET		<p>Safety circuit reset. Leave open for normal operation or add a capacitor to ground to extend the reset time.</p> <p>Connect to GND to disable window comparators at APC<sub>SET</sub></p>



**Table 2-1. Pin Definitions**

Pin Number	Pin Name	Pin equivalent load	Function
8	DISDLY		<p>Disable delay control. Connect to ground for normal operation.</p> <p>In burst mode operation add a capacitor from this pin to ground to set the maximum disable time. Disable times greater than this maximum will engage the “slow-start” circuitry.</p>
9	MOD <sub>MON</sub>		<p>Modulation Current Monitor. Connect to a resistor to GND.</p> <p>The current through this pin is typically 1/50th of the MODULATION current to the laser when CC<sub>SEL</sub> is low or 1/25th the MODULATION current when CC<sub>SEL</sub> is high.</p> <p>This pin may be left open if the feature is not needed and the M02094 current consumption will be reduced by 0.5mA typically.</p>
10	BIAS <sub>MON</sub>	See MOD <sub>MON</sub> drawing	<p>Bias Current Monitor. Connect to a resistor to GND.</p> <p>The current through this pin is typically:</p> <ul style="list-style-type: none"> <li>1/45.7th of the BIAS current to the laser when CC<sub>SEL</sub> is low</li> <li>1/13.5th of the BIAS current to the laser when CC<sub>SEL</sub> is high</li> </ul> <p>This pin may be left open if the feature is not needed and the M02094 current consumption will be reduced by 0.5mA typically.</p>

**Table 2-1. Pin Definitions**

Pin Number	Pin Name	Pin equivalent load	Function
11	TxPwr <sub>MON</sub>		<p>Transmit Power Monitor. Connect to a resistor to GND.</p> <p>The current through this pin is approximately the same as the photodiode current into I<sub>PIN</sub>. The current out of this pin is low pass filtered (no external filtering required).</p> <p>This pin may be left open if the feature is not needed and the M02094 current consumption will be reduced by the I<sub>PIN</sub> current.</p>
12	APC <sub>SET</sub>		<p>Average Power Control, laser bias current adjustment. Connect to a resistor between this pin and ground to set the bias current to the laser.</p> <p>The APC loop will adjust the laser bias current to maintain a voltage at APC<sub>SET</sub> of approximately 1.3V. The current sourced from this pin is approximately the same as the current into I<sub>PIN</sub>.</p>
13	I <sub>PIN</sub>		<p>For CC<sub>SEL</sub> low - Current input from monitor photodiode anode.</p> <p>The APC loop will adjust the laser bias current to maintain a voltage at APC<sub>SET</sub> of approximately 1.3V and at this pin of approximately one V<sub>GS</sub>.</p> <p>For CC<sub>SEL</sub> high - Current source for monitor photodiode cathode.</p> <p>The APC loop will adjust the laser bias current to maintain a voltage at APC<sub>SET</sub> of approximately 1.3V and at this pin of approximately one V<sub>GS</sub> below V<sub>CC3</sub>.</p>

**Table 2-1. Pin Definitions**

Pin Number	Pin Name	Pin equivalent load	Function
14	IBOUT <sub>CA</sub> (CC <sub>SEL</sub> = low)		<p>Laser bias current output for common anode lasers (CC<sub>SEL</sub> must be low).</p> <p>Connect directly to laser cathode or at higher bit rates through a ferrite or a resistor to isolate the capacitance of this pin from the modulation drive, (~ 6pF).</p> <p>Maintain a voltage at least 0.7V above GND at this pin</p>
	SGND (CC <sub>SEL</sub> = high)		<p>Switched ground connection for common cathode lasers (CC<sub>SEL</sub> must be high).</p> <p>Provides redundant shutdown during a disable or fault condition. This switch is disabled during common anode operation.</p>
15	GND <sub>O</sub>		<p>Ground for output stage. May be connected directly to circuit board ground. At high bit rates (&gt;2Gb/s) an optional inductor or ferrite may be added to reduce switching transients.</p>

**Table 2-1. Pin Definitions**

Pin Number	Pin Name	Pin equivalent load	Function
16	OUT+		<p>Positive modulation current output (AC or DC coupled to cathode of laser in common anode designs). Sinks current when <math>D_{IN+}</math> is HIGH.</p> <p>Maintain a voltage <math>\geq 0.7V</math> at this pin.</p>
17	OUT-	See OUT+ drawing	<p>Negative modulation current output (AC coupled to anode of laser in common cathode designs). Sinks current when <math>D_{IN-}</math> is HIGH</p> <p>Maintain a voltage <math>\geq 0.7V</math> at this pin.</p>
18	$SV_{CC}$ ( $CC_{SEL} = \text{low}$ )		<p>Switched VCC. Supplies laser current for common anode designs. (<math>CC_{SEL}</math> must be LOW).</p> <p>Provides redundant shutdown during a disable or fault condition. This switch is disabled during common cathode operation.</p>
	$IBOUT_{CC}$ ( $CC_{SEL} = \text{high}$ )		<p>Laser bias current output for common cathode laser designs (<math>CC_{SEL}</math> must be high).</p> <p>Connect directly to laser anode or at higher bit rates through a ferrite or a resistor to isolate the capacitance of this pin from the modulation drive.</p> <p>Maintain a voltage of <math>&lt; 2.5V</math> at this pin.</p>

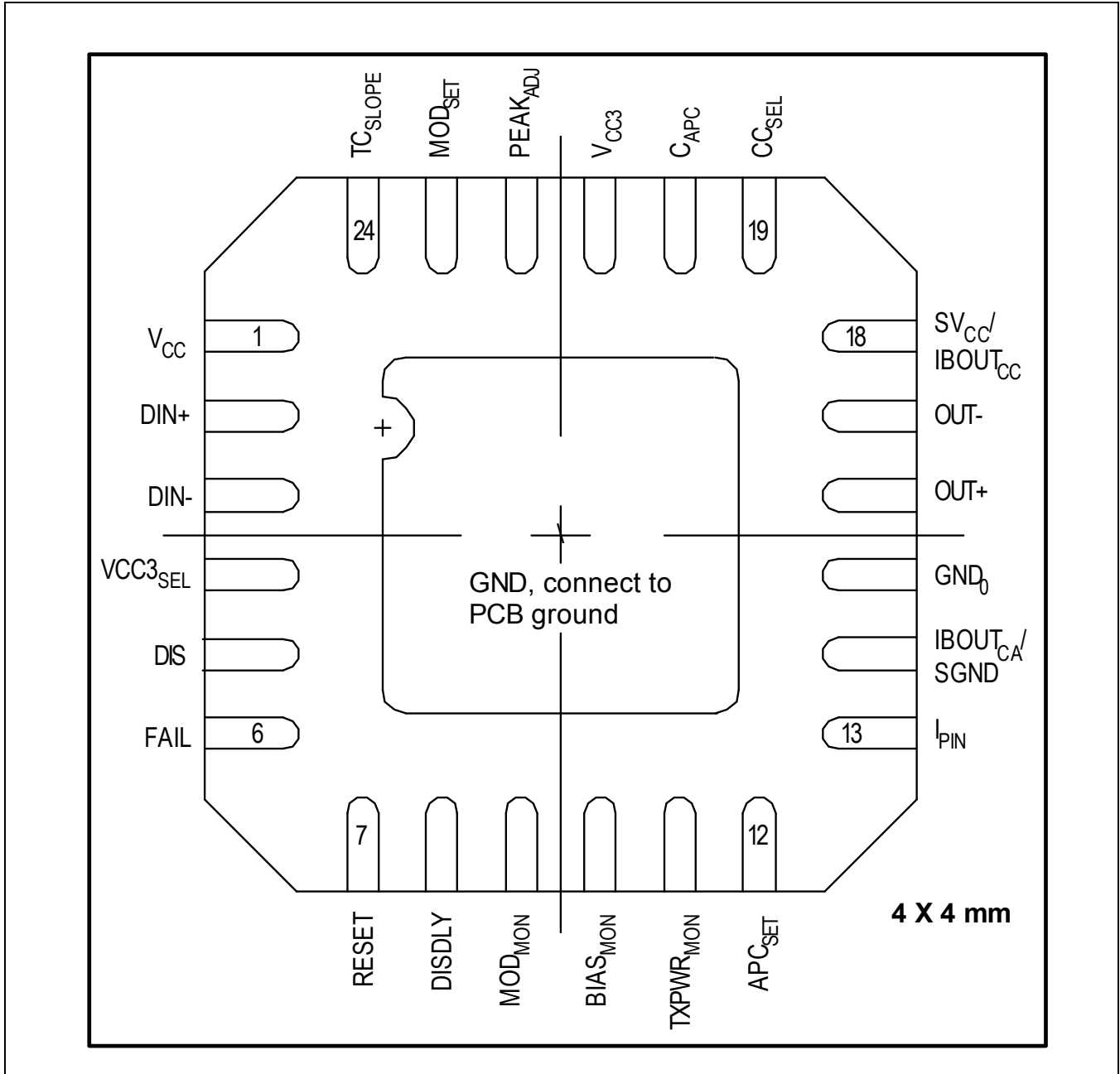
**Table 2-1. Pin Definitions**

Pin Number	Pin Name	Pin equivalent load	Function
19	CC <sub>SEL</sub>		<p>Laser select input.</p> <p>When high, common cathode operation is selected; IBOU<sub>CC</sub> and SGND are functional and internal 50Ω output terminations are switched in at the modulation outputs OUT+ and OUT-.</p> <p>When low or floating, common anode operation is selected; IBOU<sub>CA</sub> and SV<sub>CC</sub> are functional and the 50Ω internal termination resistors at OUT+ and OUT- are disconnected.</p>
20	C <sub>APC</sub>		<p>Automatic power control loop dominant pole capacitor. (Connect a capacitor between this pin and V<sub>CC3</sub>.)</p> <p>A nominal capacitance of 2.2nF will give a bias current enable time of less than 1 ms.</p>
21	V <sub>CC3</sub>		<p>3.3V applications - Power supply input. Connect to V<sub>CC</sub>.</p> <p>5V applications - Do not connect to 5V. Internally generated 3.3V power supply output. Do not attach to non-M02094 circuitry.</p>

**Table 2-1. Pin Definitions**

Pin Number	Pin Name	Pin equivalent load	Function
22	PEAK <sub>ADJ</sub>		<p>Peaking adjustment input. A resistor (2kΩ to 20kΩ) between this pin and ground sets the amount of peaking current on OUT- to improve the fall time of the laser output. The peaking current is approximately <math>4 * (1.3V / \text{resistance to ground})</math>.</p> <p>Connect to V<sub>CC3</sub> to disable peaking control.</p>
23	MOD <sub>SET</sub>		<p>Modulation current control. Connect a resistor to ground to set the modulation current.</p>
24	TC <sub>SLOPE</sub>	See MOD <sub>SET</sub> drawing	<p>Modulation current temperature compensation slope adjustment. Connect a resistor to ground to set the temperature coefficient. Leave open to minimize the temperature compensation coefficient.</p>
CENTER PAD	GND		<p>Ground. Must be connected to ground for proper operation. This is the only package ground connection.</p>

Figure 2-1. Pin Assignments for M02094 Device



## 3.0 Functional Description

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### 3.1 Overview

The M02094 is a highly integrated laser driver intended for applications to 2.0 Gbps.

Many features are user-adjustable, including common anode or common cathode laser mode, the APC (automatic power control) loop bias control (via a monitor photodiode), modulation current, temperature compensation control of modulation current, and peaking adjustment. The part may be operated from a 3.3V or 5V supply.

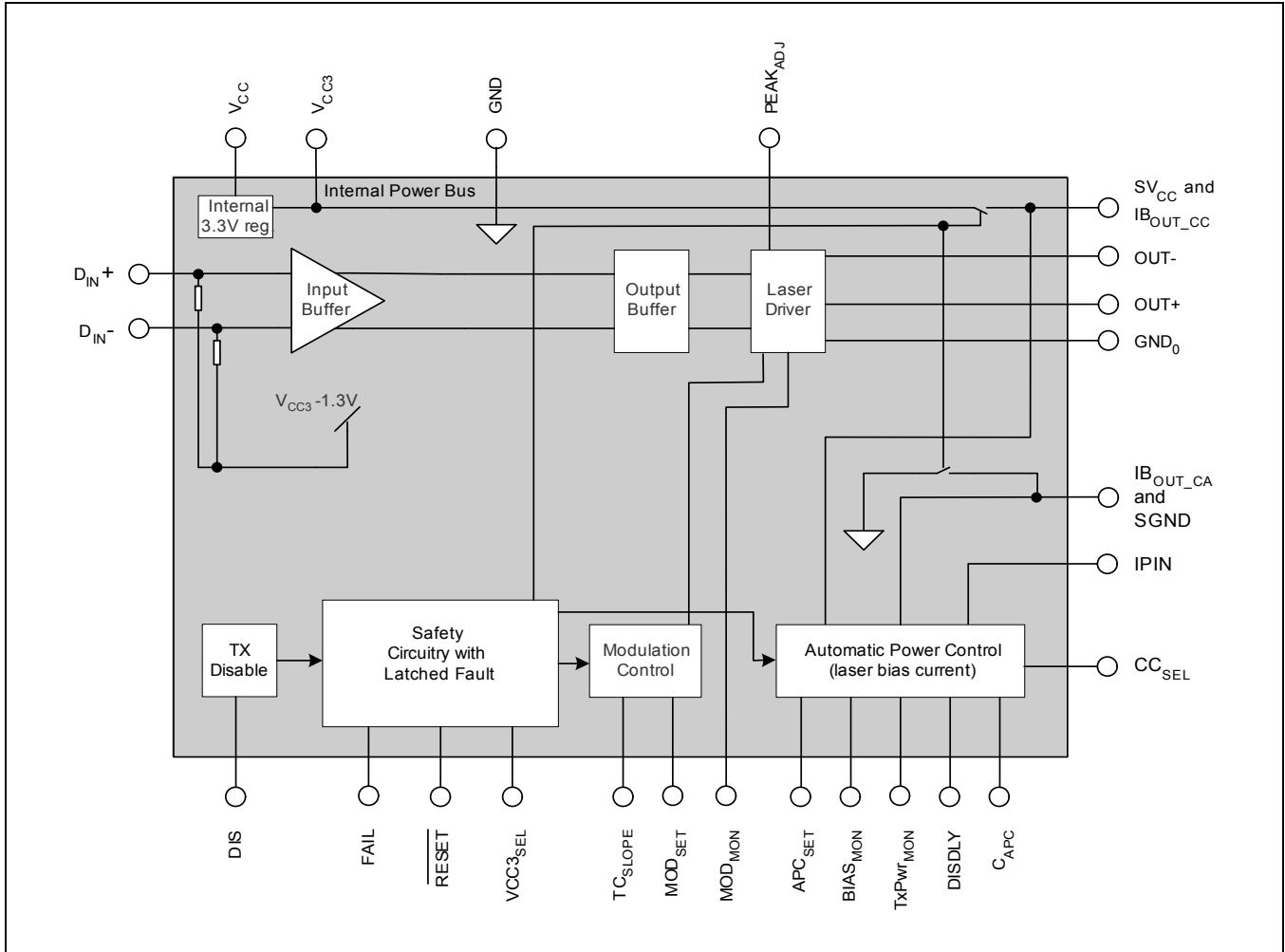
For E-PON and other burst-mode applications, the part supports fast and accurate turn-on and turn-off of the laser bias and modulation currents.

Safety circuitry is also included to provide a latched shut-down of laser bias and modulation current if a fault condition occurs. An internal VCC switch provides redundant shutdown when operating the device in common anode configuration. An internal ground switch provides redundant shutdown when operating the device in common cathode configuration.

Modulation, bias, and transmit power monitor current mirrors are provided for DDMI applications and allow monitoring without disturbing the analog signal path.



Figure 3-1. M02094 Block Diagram Example



## 3.2 Features

- High speed operation; suitable for SFP/SFF applications from 155Mbps to 2.0 Gbps. Typical rise/fall times < 110 ps.
- Supports Common Anode VCSEL, Common Cathode VCSEL, or FP LASER. May be used with or without a monitor photodiode.
- Programmable temperature compensation. Modulation output and bias output can be controlled using a few discrete resistors.
- Supports DDMI (SFF-8472) diagnostics.
- DC or AC coupled modulation drive.
- Peaking circuit to optimize VCSEL response.
- Low overshoot allows high extinction ratio with low jitter.
- Supports E-PON burst mode with no extra components
- Automatic Laser Power Control, with “Slow-Start”.
- 3.3V or 5V operation

## 3.3 General Description

### 3.3.1 Detailed Description

The M02094 laser driver consists of the following circuitry: an internal regulator, common anode/common cathode configuration control, bias current generator and automatic power control, data inputs, peaking adjust, modulation current control, modulator output, laser fail indication, disable control, and monitor outputs for the bias current, modulation current, and transmitted power.

### 3.3.2 Internal Regulator

The M02094 contains an internal 3.3V regulator so high bit rate performance can be achieved with 5V or 3.3V power supply.

When operating from a 5V supply ( $V_{CC}$  is connected to +5V), an internal regulator provides a voltage of approximately 3.3V to the majority of the on-chip circuitry. The on-chip regulator is internally compensated, requiring no external components. When a 3.3V supply is used ( $V_{CC}$  and  $V_{CC3}$  connected to 3.3V) the regulator is switched off and the internal circuitry is powered directly through the  $V_{CC3}$  supply pin. The decision as to whether or not the internal regulator is required is made via the  $VCC3_{SEL}$  pin, which also determines whether the safety circuitry needs to monitor for proper +5V supply voltage.

$SV_{CC}$  is sourced from  $V_{CC3}$  through a switch for common anode applications (this pin becomes  $IB_{OUT\_CC}$  in common cathode applications). When a fault condition is present, FAIL will assert and the switch sourcing  $SV_{CC}$  will open so no current can pass through the laser.  $SV_{CC}$  does not need any external capacitance, if capacitance to ground is added at  $SV_{CC}$  it should be less than or equal to 100 pF.

$V_{CC}$  and  $V_{CC3}$  status are internally monitored by the M02094 during power-up and normal operation. During power-up the “slow-start” circuitry requires that  $V_{CC}$  and  $V_{CC3}$  each reach an acceptable level before enabling bias or modulation current.

**Table 3-1. Pin Connection for 3.3V and 5V  $V_{CC}$**

		Pin Connection For:	
		$V_{CC} = 3.3V$	$V_{CC} = 5V$
Pins Dependent on $V_{CC}$ Voltage	$V_{CC3}$ (pin 21)	Connect to $V_{CC}$	Reference for $C_{APC}$ and $PEAK_{ADJ}$
	$C_{APC}$ (pin 20)	Capacitor between $C_{APC}$ and $V_{CC3}$ or $V_{CC}$	Capacitor between $C_{APC}$ and $V_{CC3}$ (not $V_{CC}$ )
	$PEAK_{ADJ}$ (pin 22)	Connect to $V_{CC3}$ or $V_{CC}$ to disable	Connect to $V_{CC3}$ to disable (not $V_{CC}$ )
	$VCC3_{SEL}$ (pin 4)	Connect to $V_{CC3}$ or $V_{CC}$	Connect to GND

### 3.3.3 Common Cathode/Common Anode Configuration Control

When  $CC_{SEL}$  is programmed high, the M02094 is configured for common cathode lasers. When  $CC_{SEL}$  is low, the M02094 is configured for common anode lasers.

The state of the  $CC_{SEL}$  pin determines:

1. whether bias current is sourced or sunk
2. whether monitor photodiode current is sunk or sourced
3. whether internal termination resistors at  $OUT+$  and  $OUT-$  are active
4. whether the redundant safety switch disconnects  $V_{CC}$  or GND from the output circuitry.

The affected pins are  $OUT+$ ,  $OUT-$ ,  $SV_{CC}/IB_{OUTCC}$ , and  $SGND/IB_{OUTCA}$ . The [Table 3-2](#) below shows the configuration of each pin for the 2 states of  $CC_{SEL}$ .

**Table 3-2. Pin Connection for Common Anode and Common Cathode Laser Modes**

			Pin Connection When:	
			CC <sub>SEL</sub> = High	CC <sub>SEL</sub> = Low
Pins Dependent on CC <sub>SEL</sub> Setting	Pin 18	SV <sub>CC</sub>	Inactive	Supply for laser and all output load components.
		IB <sub>OUTCC</sub>	Laser bias source current.	Inactive
	Pin 14	SGND	Ground for laser and all output load components	Inactive
		IB <sub>OUTCA</sub>	Inactive	Laser bias sink current.
	Pin 13	I <sub>PIN</sub>	Monitor photodiode source current	Monitor photodiode sink current
	Pin 22	PEAK <sub>ADJ</sub>	Controls Negative going edge of OUT-	Controls Negative going edge of OUT-, (do not use)
	Internal 50Ω pull-up resistors on OUT+ and OUT-		Active	Inactive
Ratio of Bias current to BIAS <sub>MON</sub> current		13.5:1	45.7:1	
Ratio of Modulation current to MOD <sub>MON</sub> current		30:1	65:1	

### 3.3.4 Bias Current Generator and Automatic Power Control

The M02094 can either source or sink bias current for the laser diode depending on whether it is in common anode or common cathode mode. In common cathode mode (CC<sub>SEL</sub> high) IB<sub>OUTCC</sub> will source current. In common anode mode (CC<sub>SEL</sub> low) IB<sub>OUTCA</sub> will sink current.

Regardless of whether the M02094 is configured for common anode or common cathode mode, the following information applies.

To maintain constant average optical power, the M02094 incorporates a control loop to compensate for the changes in laser threshold current over temperature and lifetime. The bias current will be determined by the value of the external resistor R<sub>APCSET</sub> and the transfer efficiency between the laser and monitor photodiode.

The photo current from the monitor photodiode mounted in the laser package is sunk or sourced at I<sub>PIN</sub>. This photo current is mirrored and an equivalent current is sourced from pins TxPwr<sub>MON</sub> and APC<sub>SET</sub>. The APC loop adjusts the laser bias current (hence the monitor diode photo current) to maintain a voltage of 1.3V at APC<sub>SET</sub>.

$$R_{APCSET} * I_{PIN} = 1.3 \text{ V}$$

The APC loop has a time constant determined by C<sub>APC</sub>, R<sub>APCSET</sub> and the transfer efficiency between the laser and monitor photodiode. The larger the C<sub>APC</sub> capacitor the lower the bandwidth of the loop and the larger the R<sub>APCSET</sub> resistor the lower the bandwidth of the loop.

In general, it is recommended that at least 2.2 nF of external capacitance be added externally between  $C_{APC}$  and  $V_{CC3}$  to assure loop stability. With use of a 2.2 nF capacitor, the bias current can reach 90% of its final value within 1 millisecond.

In Common Anode mode with a 2.2nF  $C_{APC}$  capacitor the APC loop bandwidth is less than 30 kHz for almost all combinations of  $R_{APCSET}$  and transfer efficiency., which should be adequate for bit rates of 155Mbps. (and all higher bit rates).

In Common Cathode mode with a 2.2nF  $C_{APC}$  capacitor the APC loop bandwidth will be slightly higher, but should be less than 40 kHz for almost all combinations of  $R_{APCSET}$  and transfer efficiency. Contact the factory with your specific values of  $C_{APC}$ ,  $R_{APCSET}$ , and transfer efficiency to determine the maximum APC loop bandwidth in your application.

The bias generator also includes a bias current monitor mirror ( $BIAS_{MON}$ ), whose output current is typically 1/45.7th of the bias current in common anode mode ( $CCSEL = low$ ) or 1/13.5th of the bias current in common cathode mode ( $CCSEL = high$ ). This pin can be connected through a resistor to ground. If this function is not needed this pin can be left open.

The M02094 can be used without a monitor photodiode by connecting  $BIAS_{MON}$  to  $APC_{SET}$  (see [Figure 3-7](#) and [Figure 4-3](#)). In this case the M02094 will increase the bias current (hence the  $BIAS_{MON}$  current) to the laser until the voltage at  $APC_{SET}$  is approximately 1.3V.

### 3.3.5 Data Inputs

The inputs to the data buffers are self-biased through 4 k $\Omega$  resistors to an internal voltage  $V_{TT}$  which is approximately  $V_{CC3} - 1.3V$ . Both CML and PECL inputs signals can be AC coupled to the M02094, or in 3.3V applications PECL inputs can be DC coupled to the data inputs. In most applications the data inputs are AC coupled with controlled impedance pcb traces which will need to be terminated externally with a 100 $\Omega$  or 150 $\Omega$  resistor between the + and - inputs.

### 3.3.6 Peak Adjust

Some VCSELs do not turn off quickly without peaking the negative going edge.

In common cathode applications, peaking on this edge can be added with a resistor connected between the  $PEAK_{ADJ}$  input and GND. The amount of peaking is approximately

Peaking current =  $4 * (1.3V / \text{resistance to ground})$ .

The resistance to ground should be between 2k $\Omega$  and 20k $\Omega$ . (Which will result in a peaking currents from 2.6mA to 260 $\mu A$ .)

Peaking control can be disabled by connecting  $PEAK_{ADJ}$  to  $V_{CC3}$ , resulting in no peaking current and reducing supply current by approximately 2mA.

In common anode configuration the  $PEAK_{ADJ}$  pin should be connected to  $V_{CC3}$ .

*Note:* Unlike the rest of the signal currents in the M02094, the output Peak Adjust current is unbalanced (single-sided drive). The designer should be aware that the use of peaking may result in unwanted EMI emissions. If EMI problems are traced to the use of peaking, high frequency decoupling (10pF capacitor or smaller) may be needed on the  $V_{CC}$  line.

### 3.3.7 Modulation Control

There are two programmable control lines for controlling the modulation current and its temperature compensation. These inputs can be programmed simply with a resistor to ground .

The modulation current amplitude is controlled by the MOD<sub>SET</sub> input pin. The modulation current is temperature compensated by the TC<sub>SLOPE</sub> input.

If the temperature compensation at TC<sub>SLOPE</sub> is disabled, the modulation output current is simply:

$$I_{OUT} = 42 \times (1.3V / R_{MODSET}) \text{ when } CC_{SEL} \text{ is low}$$

and

$$I_{OUT} = 22 \times (1.3V / R_{MODSET}) \text{ when } CC_{SEL} \text{ is high and a } 50\Omega \text{ VCSEL is used.}$$

Where R<sub>MODSET</sub> is the resistance from pin MOD<sub>SET</sub> to ground.

To temperature compensate the modulation current, choose R<sub>TC<sub>SLOPE</sub></sub> to meet the following relationship:

$R_{TC_{SLOPE}} = 19.5 \times (TC)^{-1.5}$ , where TC is the desired slope of the modulation current from 25°C to 85°C in%/°C and R<sub>TC<sub>SLOPE</sub></sub> is in kΩ. If no temperature compensation is desired, leave R<sub>TC<sub>SLOPE</sub></sub> open.

In any case, R<sub>TC<sub>SLOPE</sub></sub> will have negligible effect at M02094 case temperatures below 10°C.

For example:

Given a common cathode VCSEL with a desired modulation current at low temperatures of 10mA and a temperature coefficient of -0.5%/°C at high temperatures (which will require a laser driver temperature coefficient of +0.5%).

$$\text{Choose } R_{MODSET} = 22 \times (1.3V / 10mA) = 2.8k\Omega.$$

$$\text{Choose } R_{TC_{SLOPE}} = 19.5 \times (0.5)^{-1.5} k\Omega = 55k\Omega.$$

**Figure 3-2. Selecting  $TC_{SLOPE}$  Resistance in Common Anode Configuration**

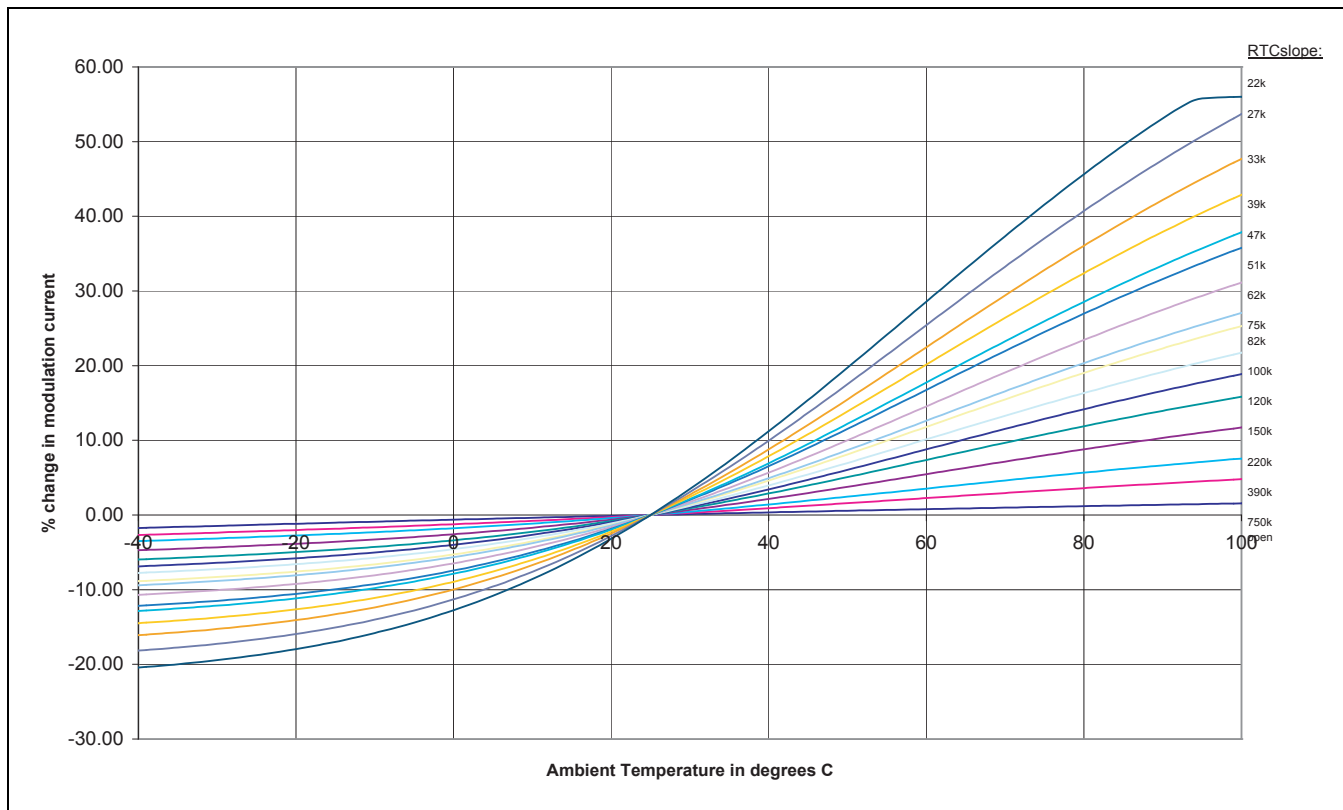
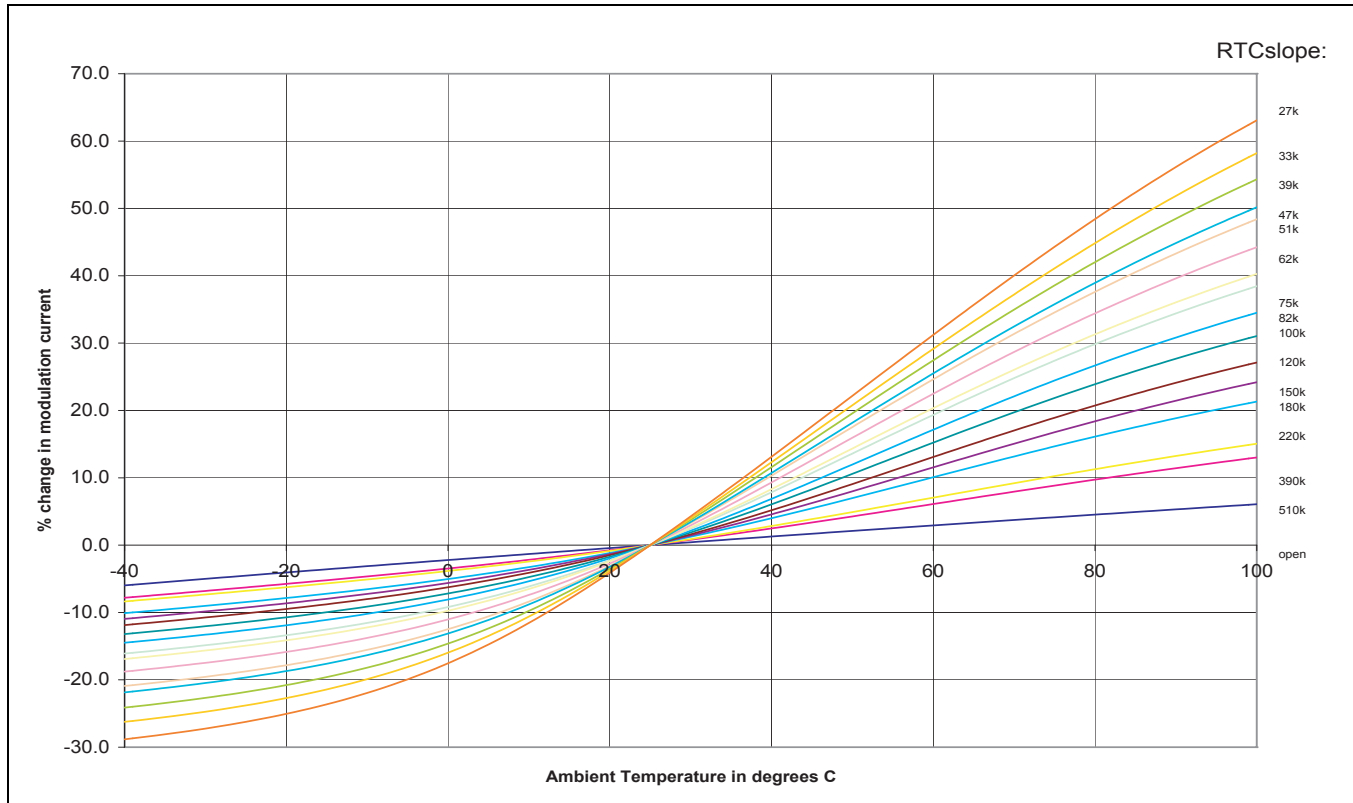


Figure 3-3. Selecting  $TC_{SLOPE}$  Resistance in Common Cathode Configuration

### 3.3.8 Modulator (Figure 3-4)

The output stages OUT+ and OUT- are designed to drive 25Ω output loads over a wide range of currents and circuit architectures. The VCSEL may be a common anode or common cathode device. The output can be AC, DC, or Differentially coupled depending on the supply voltage and laser configuration.

In a common anode configuration with a VCSEL (Figure 4-1), OUT+ should be connected through a capacitor to the VCSEL. A pull-up resistor should be added in parallel to the VCSEL from SVCC to the OUT+ output. The dynamic impedance of the parallel combination of the VCSEL and pull-up resistor should be roughly 25 ohms. A 24 ohm pull-up resistor should also be added from SVCC to OUT- so the currents and voltage swings in the two outputs are balanced.

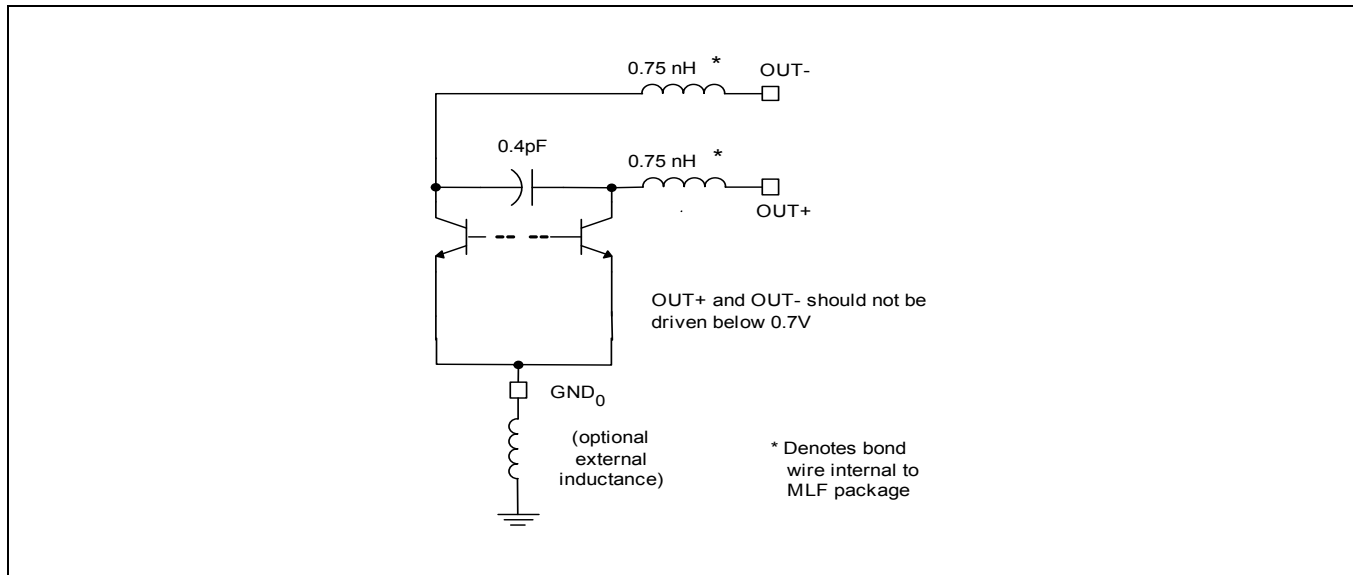
In a common anode configuration with a Fabry-Perot laser (Figure 4-4), OUT+ may be AC, DC, or Differentially coupled to the laser cathode. A resistor should be added in series with the laser such that the dynamic impedance of the series combination of the laser and resistor should be roughly 25 ohms. A 24 ohm pull-up resistor to SVCC is needed on the OUT- output.

For common cathode operation with a VCSEL (Figure 4-3), internal 50 ohm terminations are switched in between the OUT+ and OUT- outputs and VCC3. VCSELs with impedances from 25-75Ω can be simply AC coupled to the OUT- output with no additional load matching resistors. In this case OUT+ should be AC coupled to ground through 50Ω.



The VCSEL driver output stage is separately grounded from the rest of the circuitry (through  $GND_0$ ). At higher data rates (above 2Gb/s)  $GND_0$  may be connected to ground through a minimum of 2 nH of inductance to improve the transient response. A ferrite can also provide the extra isolation (Murata BLM18HG471SN1 or equivalent recommended).

**Figure 3-4. Modulator Output**



### 3.3.9 Fail Output

The M02094 has a FAIL alarm output which is compatible with the TX\_FAULT signalling requirements of common pluggable module standards.

The ESD protection on this pin provides a true open collector output that can withstand significant variation in  $V_{CC}$  when signalling between circuit boards. Also, if the M02094 loses power the FAIL output will continue to pull up and signal a fail condition. In a simple static protection scheme used by other ICs the protection diodes would clamp the FAIL signal to ground when the chip loses power.

### 3.3.10 TX Disable and Disable Delay Control

The DIS pin is used to disable the transmit signal. When the transmit is disabled both the bias and modulation currents are off.

The DIS input is compatible with TTL levels regardless of whether  $V_{CC} = 5V$  or  $V_{CC} = 3.3V$ . In most module applications a pull-up resistor to  $V_{CC}$  between 4.7 k $\Omega$  and 10 k $\Omega$  is required. Because this pin has an internal 7 k $\Omega$  resistor to  $V_{CC}$ , no external pull-up resistor is required.

The DISDLY pin is used in conjunction with the DIS pin to control bias current enable time. (The modulation current enable time is always less than 600 ns). Unless the DISDLY pin is programmed for burst mode, the APC loop enable time will be slow (less than 1 ms with a  $C_{APC} = 2.2$  nF).

When a capacitor C is added to the DISDLY pin, the slow-start circuitry is disabled for typically

$$T = 3 * 10^6 \text{ (sec/F)} * C \text{ (F)}$$

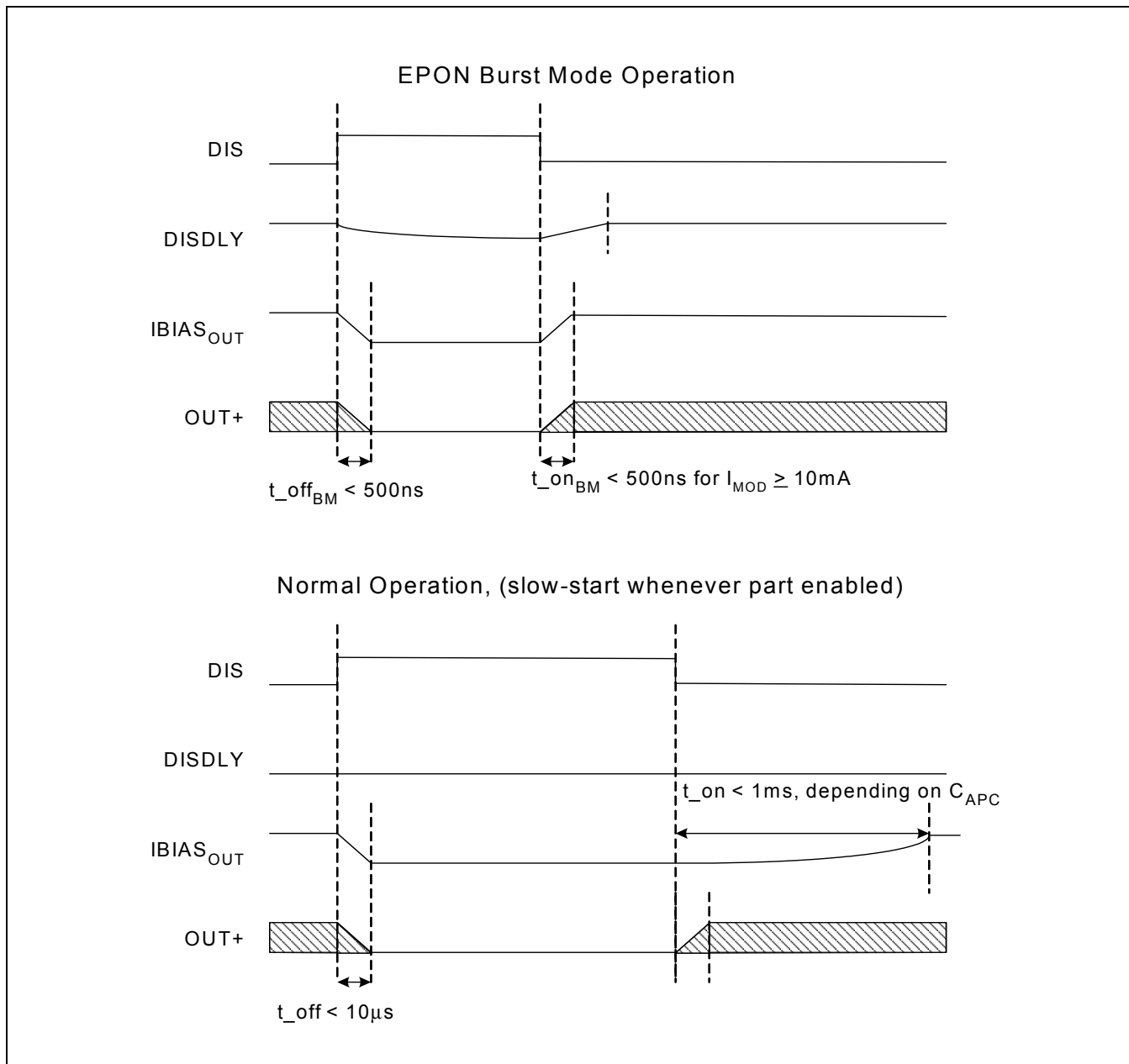
following the DIS high transition. If DIS transitions low during this time, the bias current will quickly return to within 90% of its final value (within less than 500 ns). If DIS transitions low after this time the slow-start circuitry will engage and the bias current will not return to its final value for approximately 1ms (depending on the C<sub>APC</sub> capacitor).

### 3.3.11 Burst Mode Operation (Figure 3-5)

The M02094 will meet the timing requirements of EPON with the addition of a capacitor at DISDLY (see paragraph above and Figure 3-5).

As shown in Figure 4-4, the laser should be DC coupled to OUT+. V<sub>CC</sub> may be 3.3V or 5V.

Figure 3-5. DIS and DISLY Timing



### 3.3.12 Video Operation

The M02094 can be used to transmit video optical data even in the presence of the pathological pattern. This is done by fully DC coupling the signal from the input to the laser output.

In most data communications applications, AC coupling occurs at three points in a laser driver schematic: the data inputs, the APC control, and coupling the modulation current to the laser. In the M02069, DC coupling can be used at all three of these points. With DC coupling the laser output will not be a function of the data input one/zero density.

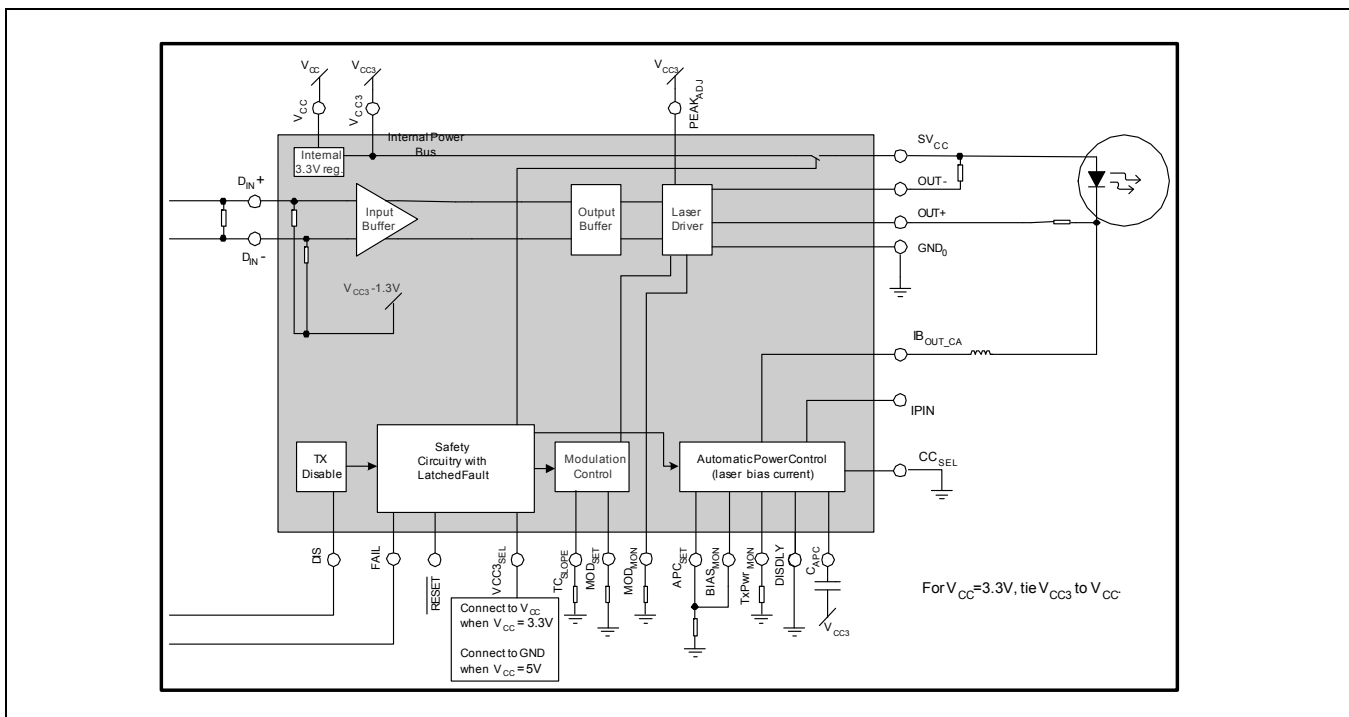
The data inputs can be DC coupled using PECL or CML levels.

The APC of the bias current is controlled by feedback from the monitor photodiode in the laser package in most communications applications. In video applications this monitor photodiode should not be used if the pathological pattern may occur. Instead, the APC should be controlled in an "open loop" configuration. (Open loop simply means a monitor photodiode is not used). In the open loop configuration the APC is controlled by a resistor or a thermistor network or a look-up table and DAC. This removes AC time constants from the bias current. In [Figure 3-6](#) the BIASmon pin is connected to the APCset pin. In this case the bias current is:

$$I_{BIAS} = 45.7 \times (1.35V / R_{APC_{SET}})$$

The modulation current output OUT+ can be DC coupled to the laser as shown in [Figure 3-6](#). There are no AC time constants in the modulation current amplitude in this configuration.

**Figure 3-6. Video Application Block Diagram**



### 3.3.13 Current Monitors

To facilitate complying with laser safety and DDMI<sup>1</sup> requirements, output monitors are provided for transmit power ( $TxPwr_{MON}$ ), bias ( $BIAS_{MON}$ ), and modulation current ( $MOD_{MON}$ ).

These outputs will source current proportional to the emitted optical power ( $TxPwr_{MON}$ ) the bias current ( $BIAS_{MON}$ ) and modulation current ( $MOD_{MON}$ ).

To use these pins they should be terminated with a resistor to ground that sets the desired full-scale voltage (not to exceed 2.5V).

If the outputs of these monitors are not needed,  $TxPwr_{MON}$ ,  $BIAS_{MON}$ , and  $MOD_{MON}$  can all be left floating and the chip current consumption will be reduced by the value of the monitor currents.

## 3.4 Laser Eye Safety

Using this laser driver in the manner described herein does not ensure that the resulting laser transmitter complies with established standards such as IEC 825. Users must take the necessary precautions to ensure that eye safety and other applicable standards are met. Note that determining and implementing the level of fault tolerance required by the applications that this part is going into is the responsibility of the transmitter designer and manufacturer since the application of this device cannot be controlled by MACOM.

<sup>1</sup> Digital Diagnostic Monitoring Interface for Optical Transceivers, defined in SFF-8472.

### 3.4.1 Safety Circuitry

Safety Circuitry in the M02094 will disable the modulation and bias current and assert the FAIL output immediately upon detecting a fault condition. In addition, the supply voltage that sources or sinks the laser current (SV<sub>CC</sub> or SGND) will immediately go open circuit and prevent any current from passing through the laser.

Fault conditions checked by the M02094 include shorts to ground or V<sub>CC</sub> of all pins which can increase the laser modulation or bias current.

For an initialization or power-up sequence to be successful, all the fault detection monitors must signal that the chip is “healthy”.

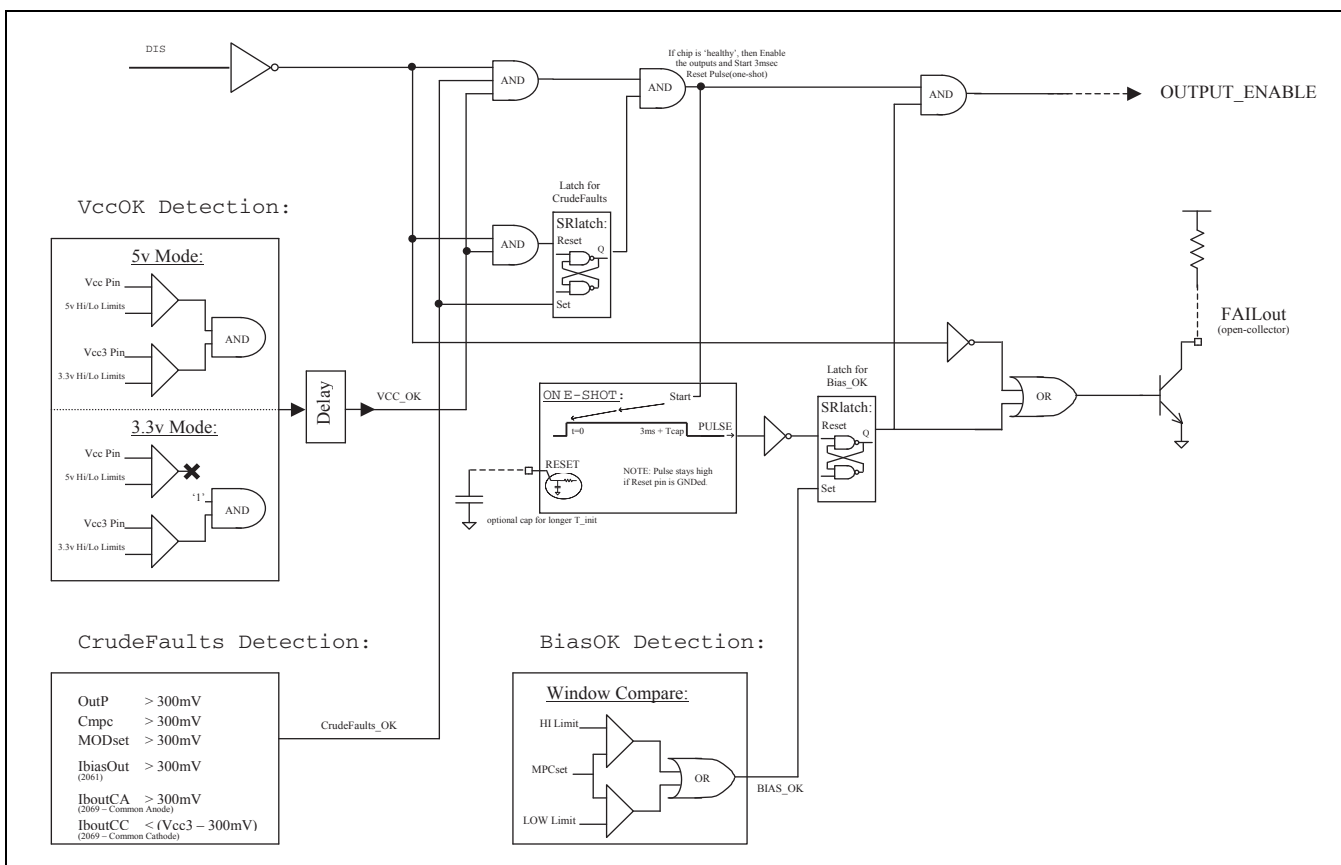
When DIS goes low, pins are checked for shorts to ground or V<sub>CC</sub> and a FAIL condition is latched if there is a fault.

If the state of the pins is OK, a one-shot at the reset pin begins a countdown which will latch a FAIL condition if the bias current has not stabilized to an acceptable level during the one-shot time. The one-shot can be extended with an external capacitor connected from the RESET pin to ground.

The one-shot<sup>1</sup> width is approximately

$$T_{\text{ONE-SHOT}} = 3 \text{ ms} + (0.3 \text{ ms/pF}) \times (\text{external capacitance}).$$

**Figure 3-7. Safety Circuit Block Diagram**



1. The one-shot is actually comprised of an oscillator and 10-bit counter.

## 3.5 Fault Conditions

This section describes the M02094 operating modes during fault conditions. Over voltage, under voltage, pins shorted to  $V_{CC}$  and pins shorted to ground are included in the fault [Table 3-3](#).

**Table 3-3. Circuit Response to Single-Point Fault Conditions**

Pin Name	Circuit Response to Over-voltage Condition or Short to $V_{CC}$	Circuit Response to Under-Voltage Condition or Short to Ground
$V_{CC}$	Bias and modulation outputs are disabled once $V_{CC}$ rises above the supply detection (high voltage) threshold (see <a href="#">Table 1-3</a> )	Bias and modulation outputs are disabled once $V_{CC}$ drops below the supply detection (low voltage) threshold
DIN+, DIN-	The APC loop will attempt to compensate for the change in output power. If the APC loop can not maintain the set average power, a fault state occurs. <sup>(1,2,3)</sup>	The APC loop will attempt to compensate for the change in output power. If the APC loop can not maintain the set average power, a fault state occurs. <sup>(1,2,3)</sup>
$V_{CC3_{SEL}}$	Does not affect laser power.	Does not affect laser power.
DIS	Bias and modulation outputs are disabled. $SV_{CC}$ is opened when $CC_{SEL}$ is low or floating (or SGND is opened when $CC_{SEL}$ is high)	Does not affect laser power (normal condition for circuit operation).
FAIL	Does not affect laser power.	Does not affect laser power.
RESET	Does not affect laser power.	Does not affect laser power.
MOD <sub>MON</sub>	Does not affect laser power.	Does not affect laser power.
BIAS <sub>MON</sub>	Does not affect laser power.	Does not affect laser power.
TxPWR <sub>MON</sub>	Does not affect laser power.	Does not affect laser power.
APC <sub>SET</sub>	A fault state occurs. <sup>(1)</sup>	A fault state occurs. <sup>(1)</sup>
$I_{PIN}$	A fault state occurs. <sup>(1)</sup>	A fault state occurs. <sup>(1)</sup>
IBOUT <sub>CA</sub> <sup>(3)</sup>	The laser will be turned off, then a fault state occurs. <sup>(1)</sup>	A fault state occurs. <sup>(1)</sup>
IBOUT <sub>CC</sub> <sup>(4)</sup>	A fault state occurs. <sup>(1)</sup>	The laser will be turned off, then a fault state occurs. <sup>(1)</sup>
OUT+ <sup>(3)</sup>	Laser modulation is prevented; the APC loop will increase the bias current to compensate for the drop in laser power if it is DC coupled. If the set output power can not be obtained, a fault state occurs. <sup>(1,2)</sup>	A fault state occurs. <sup>(1)</sup>
OUT- <sup>(4)</sup>	Does not affect laser power during common cathode operation because output is AC coupled.	Does not affect laser power during common cathode operation because output is AC coupled.
$SV_{CC}$ <sup>(3)</sup>	Does not affect laser power.	Laser bias current will be shut off and a fault state occurs. <sup>(1)</sup>
$C_{APC}$	Laser bias current will be shut off, then a fault state occurs. <sup>(1)</sup>	A fault state occurs. <sup>(1)</sup>
$V_{CC3}$	Bias and modulation outputs are disabled once $V_{CC3}$ rises above the supply detection (high voltage) threshold	Bias and modulation outputs are disabled once $V_{CC3}$ drops below the supply detection (low voltage) threshold
PEAK <sub>ADJ</sub>	Does not affect laser power.	Does not affect laser power
$CC_{SEL}$	Normal operation for common cathode configuration. If the M02094 is configured for common anode drive, a fault state will occur. <sup>(1)</sup>	Normal operation for common anode configuration. If the M02094 is configured for common cathode drive, a fault state will occur. <sup>(1)</sup>

**Table 3-3. Circuit Response to Single-Point Fault Conditions**

Pin Name	Circuit Response to Over-voltage Condition or Short to $V_{CC}$	Circuit Response to Under-Voltage Condition or Short to Ground
MOD <sub>SET</sub>	When the laser is DC coupled in common anode configuration, the APC loop will attempt to compensate for the drop in output power. If the APC loop can not maintain the set average power, a fault state occurs. <sup>(1,2)</sup>	A fault state occurs. <sup>(1)</sup>
TC <sub>SLOPE</sub>	Does not affect laser power.	When the laser is DC coupled in common anode configuration, the APC loop will attempt to compensate for any change in output power. If the APC loop can not maintain the set average power, a fault state occurs. <sup>(1,2)</sup>
DISDLY	Does not affect laser power.	Does not affect laser power.
Notes: 1. A fault state will assert the FAIL output, disable bias and modulation outputs and will either open the switch at $SV_{CC}$ ( $CC_{SEL}$ =high) or $S_{GND}$ ( $CC_{SEL}$ =low). 2. Does not affect laser power when the output is AC coupled to the laser. 3. Does not affect laser power during common cathode operation. 4. Does not affect laser power during common anode operation.		



# 4.0 Applications

## 4.1 Applications

- EPON FTTH modules
- Gigabit Ethernet modules
- 1G/2G Fibre Channel modules
- Digital Video

**Figure 4-1. Application Diagram, Common Anode VCSEL (may be DC coupled, but extra bias current will be required for the pull-up resistor on OUT+)**

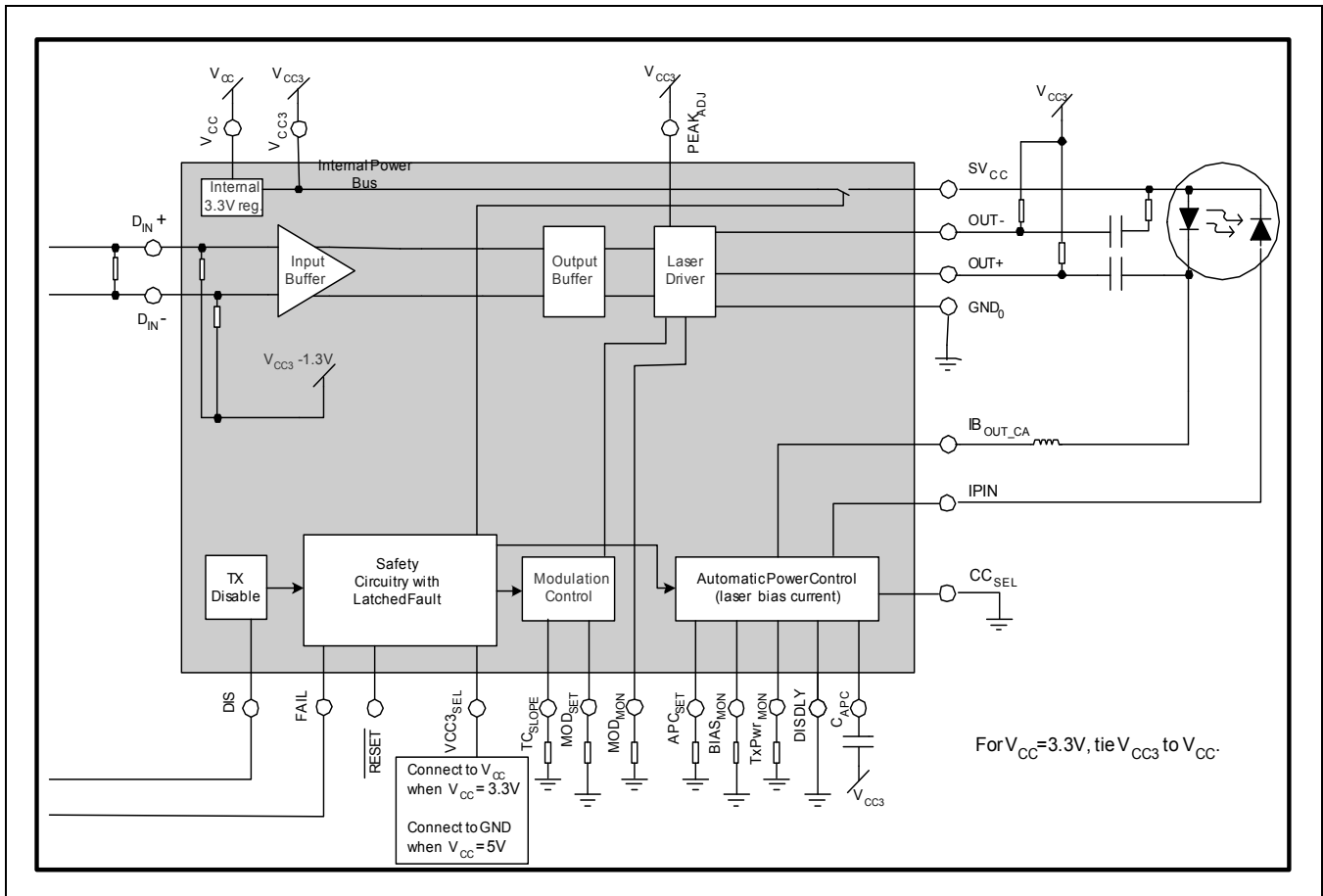


Figure 4-2. Application Diagram, Common Cathode VCSEL

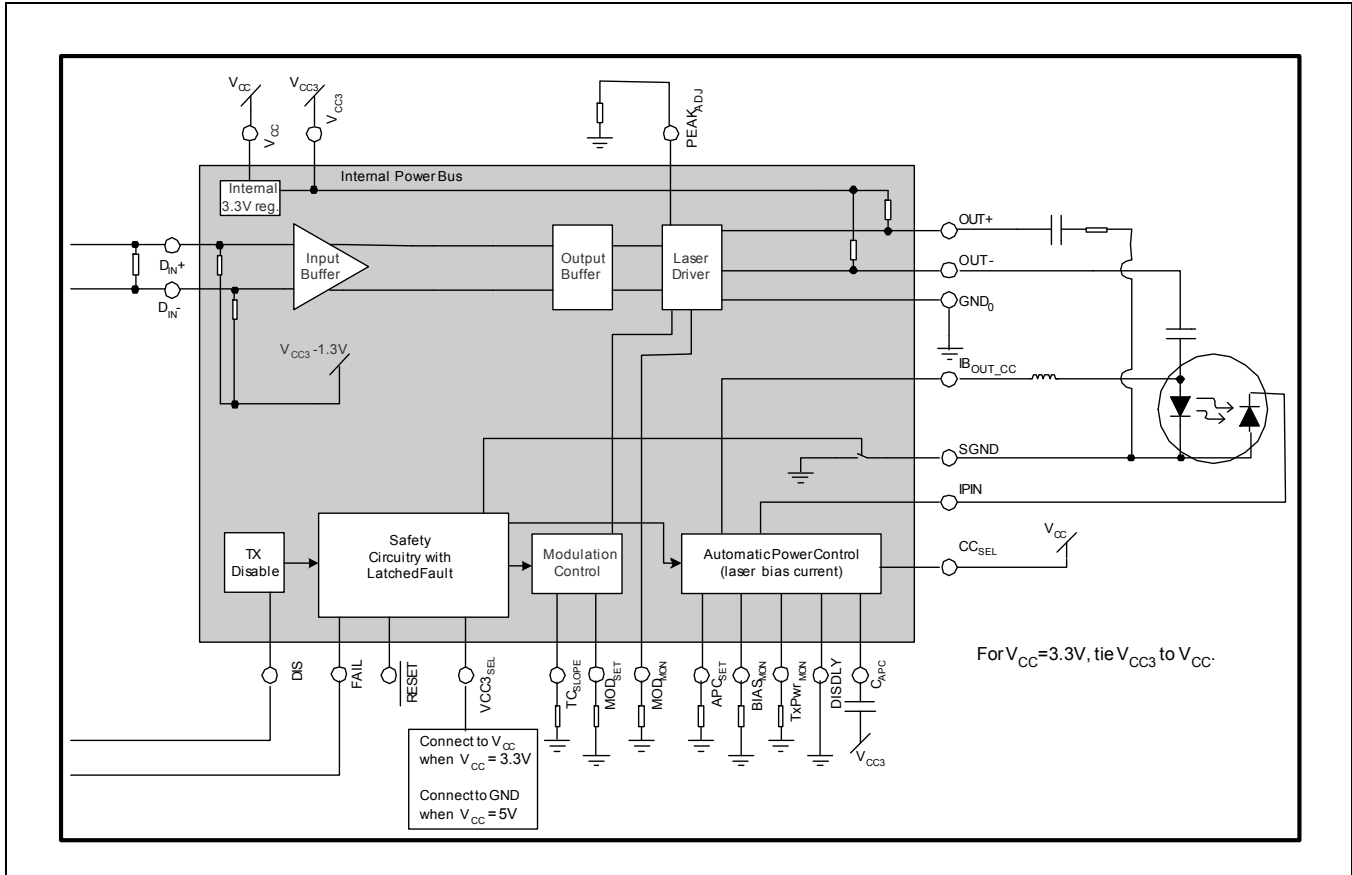
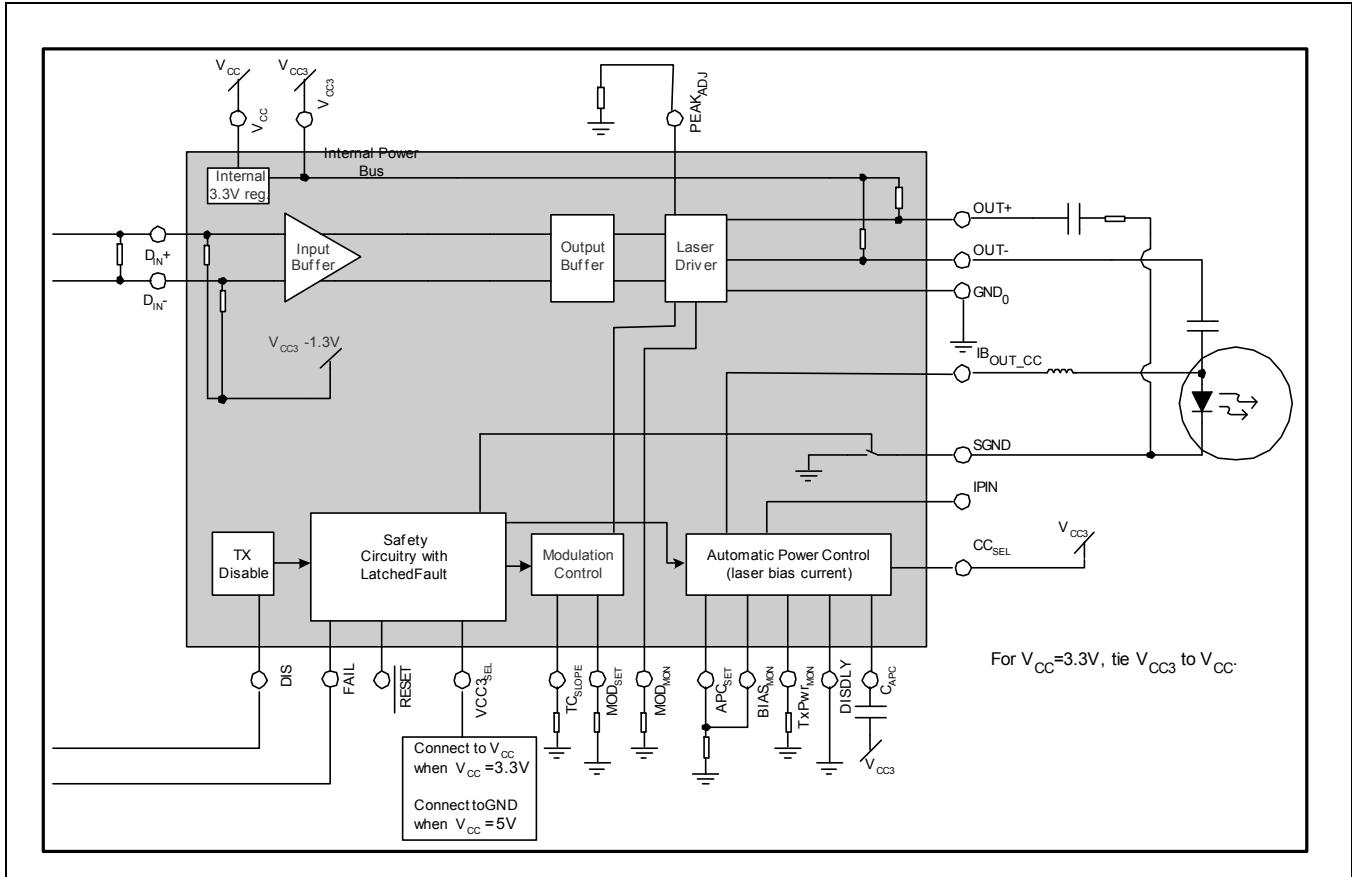
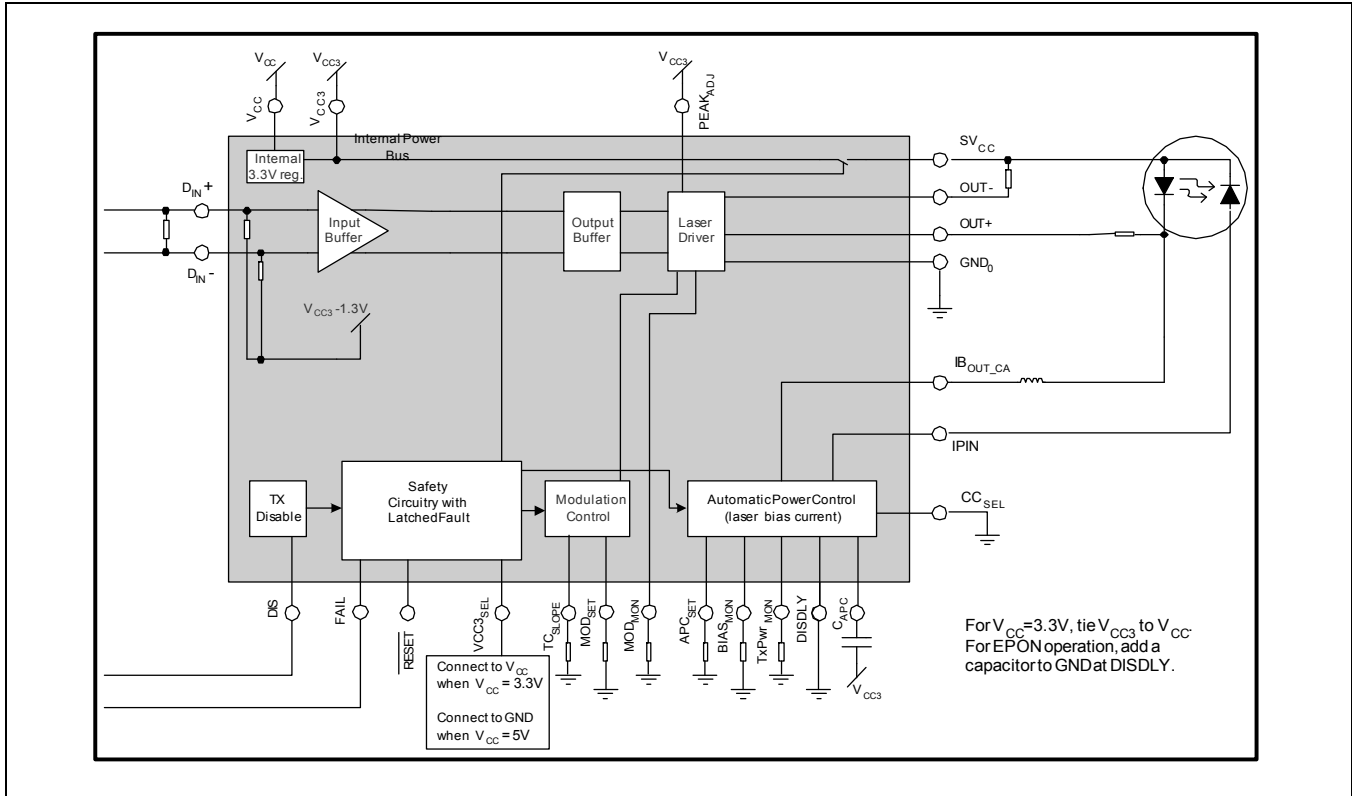


Figure 4-3. Application Diagram, Common Cathode VCSEL w/o Monitor Diode



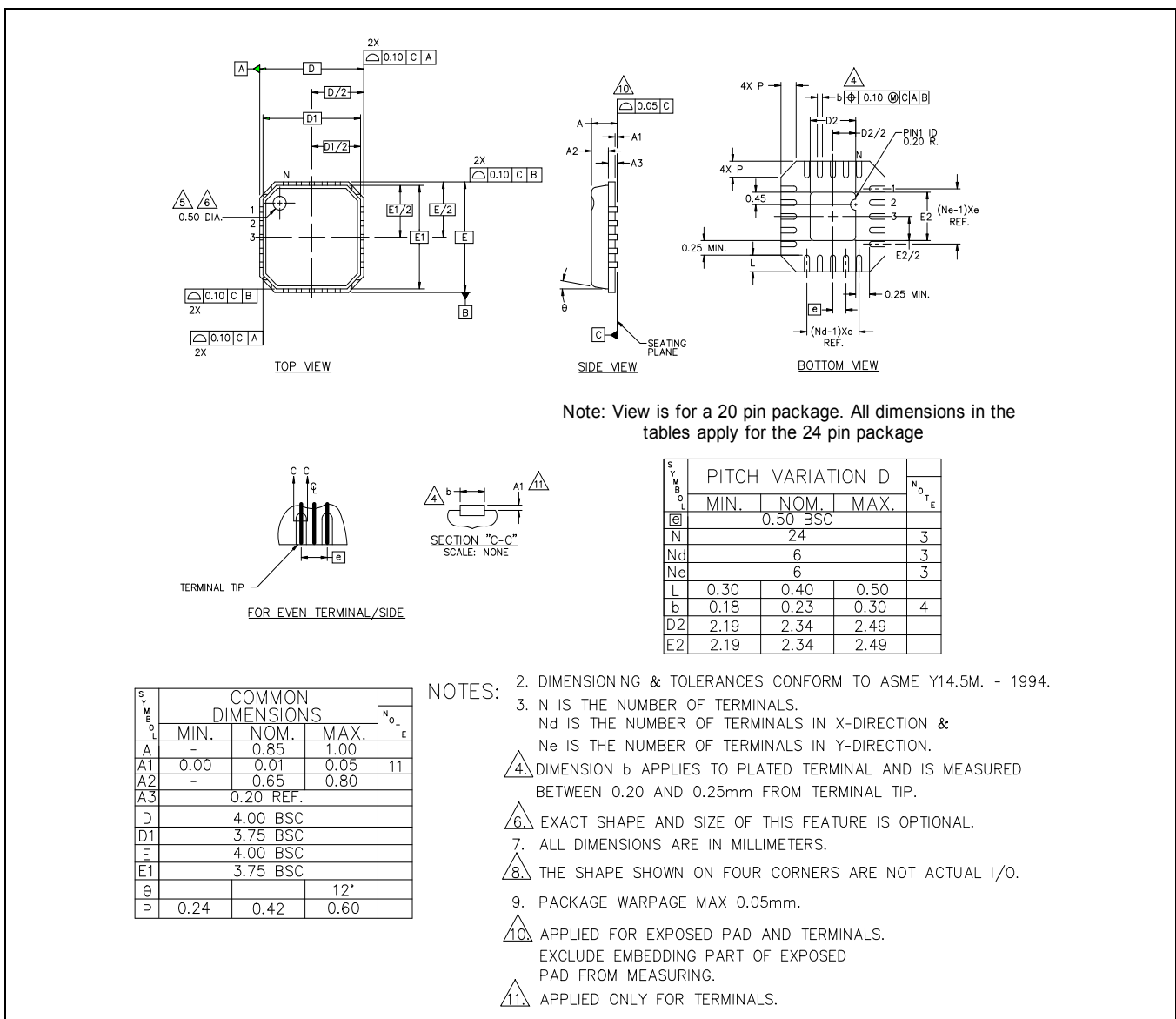
**Figure 4-4. Application Diagram, Common Anode FP Laser, DC Coupling**



# 5.0 Packaging Specification

## 5.1 Package Specification

Figure 5-1. QFN24 Package Information



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