

## 16-Bit Low-Power Stereo Audio ADC With Microphone Bias and Microphone Amplifier

### FEATURES

- **Analog Front End:**
  - Stereo Single End Input With MUX
  - Mono Differential Input
  - Stereo Programmable Gain Amplifier
  - Microphone Boost Amplifier and Bias
- **Analog Performances Dynamic Range: 90 dB**
- **Power-Supply Voltage**
  - 1.71 V to 3.6 V for Digital I/O Section
  - 1.71 V to 3.6 V for Digital Core Section
  - 2.4 V to 3.6 V for Analog Section
- **Low Power Dissipation:**
  - 13 mW in Record, 1.8/2.4 V, 48 kHz, Stereo
  - 5.3 mW in Record, 1.8/2.4 V, 8 kHz, Mono
  - 3.3  $\mu$ W in All Power Down
- **Sampling Frequency: 5 kHz to 50 kHz**
- **Auto Level Control for Recording**
- **Operation by Single Clock Input Without PLL**
- **System Clock: Common Audio Clock (256 f<sub>s</sub>/384 f<sub>s</sub>), 12/24, 13/26, 13.5/27, 19.2/38.4, 19.68/39.36 MHz**
- **2 (I<sup>2</sup>C) or 3 (SPI) Wire Serial Control**
- **Programmable Function by Register Control:**
  - Digital Soft Mute
  - Hi-Z for LRCK, BCK and DOUT
  - Power Up/Down Control for Each Module
  - 30-dB to –12-dB Gain for Analog Inputs
  - 0/12/20-dB Boost for Microphone Input
  - Parameter Settings for ALC
  - Three-Band Tone Control and 3D Sound
  - High-Pass Filter and Two-Stage Notch Filter

- **Pop Noise Reduction Circuit**
- **Package:**
  - 24-Pin YZF (2.49 mm × 3.49 mm)
- **Operation Temperature Range: –40°C to 85°C**

### APPLICATIONS

- **Portable Audio Player, Cellular Phone**
- **Video Camcorder, Movie Digital Still Camera**
- **PMP/DMB, Voice Recorder**

### DESCRIPTION

The PCM1870A is a low-power stereo ADC designed for portable digital audio applications, with line-input amplifier, boost amplifier, microphone bias, programmable gain control, sound effects, and auto level control (ALC). It is available in a 24-Pin YZF (2.49-mm × 3.49-mm) package to save footprint. The PCM1870A accepts right-justified, left-justified, I<sup>2</sup>S, and DSP formats, providing easy interfacing to audio DSP and encoder chips. Sampling rates up to 50 kHz are supported. The user-programmable functions are accessible through a 2- or 3-wire serial control port.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

## ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

		PCM1870A	UNIT
$V_{DD}$ , $V_{IO}$ , $V_{CC}$	Supply voltage	–0.3 to 4	V
	Ground voltage differences: DGND, AGND, PGND	±0.1	V
	Input voltage	–0.3 to 4	V
	Input current (any pins except supplies)	±10	mA
	Ambient temperature under bias	–40 to 110	°C
	Storage temperature	–55 to 150	°C
	Junction temperature	150	°C
	Lead temperature (soldering)	260 / 5	°C / s
	Package temperature (reflow, peak)	260	°C

(1) Stresses beyond those listed under *absolute maximum ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *recommended operating conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

## RECOMMENDED OPERATING CONDITIONS

		MIN	NOM	MAX	UNIT	
$V_{CC}$	Analog supply voltage	2.4	3.3	3.6	V	
$V_{DD}$ , $V_{IO}$	Digital supply voltage	1.71	3.3	3.6	V	
	Digital input logic family	CMOS				
	Digital input clock frequency	SCKI system clock		18.432	MHz	
		LRCK sampling clock		48	kHz	
	Digital output load capacitance				10	pF
$T_A$	Operating free-air temperature	–40			85	°C

## ELECTRICAL CHARACTERISTICS

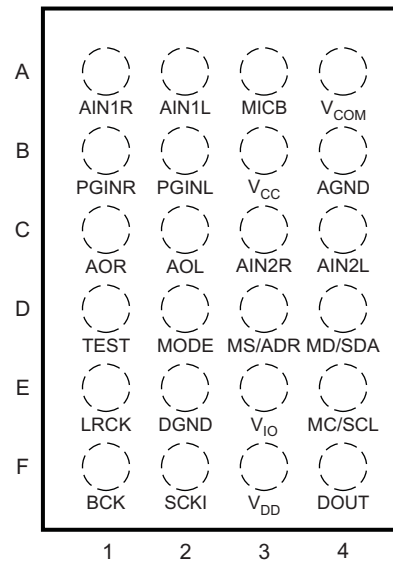
All specifications at  $T_A = 25^\circ\text{C}$ ,  $V_{DD} = V_{IO} = V_{CC} = V_{PA} = 3.3\text{ V}$ ,  $f_s = 48\text{ kHz}$ , system clock =  $256 f_s$ , and 16-bit data, unless otherwise noted

PARAMETER	TEST CONDITIONS	PCM1870AYZF			UNIT
		MIN	TYP	MAX	
<b>AUDIO DATA</b>					
<b>Data Format</b>					
Resolution		16			Bits
Audio data interface format		I <sup>2</sup> S, left-, right-justified, DSP			
Audio data bit length		16			Bits
Audio data format		MSB-first, 2s-complement			
$f_s$ Sampling frequency		5		50	kHz
System clock	$V_{DD} < 2\text{ V}$			27	MHz
	$V_{DD} > 2\text{ V}$			40	
<b>DIGITAL INPUT/OUTPUT</b>					
Logic family		CMOS compatible			
$V_{IH}$	Input logic level	0.7 $V_{IO}$			VDC
$V_{IL}$		0.3 $V_{IO}$			
$I_{IH}$	Input logic current	$V_{IN} = 3.3\text{ V}$	10		$\mu\text{A}$
$I_{IL}$		$V_{IN} = 0\text{ V}$	-10		
$V_{OH}$	Output logic level	$I_{OH} = -2\text{ mA}$	0.75 $V_{IO}$		VDC
$V_{OL}$		$I_{OL} = 2\text{ mA}$	0.25 $V_{IO}$		
<b>LINE INPUT TO DIGITAL OUTPUT THROUGH ADC (AIN1L/R, AIN2L/R AND PGINL/R—ALC = OFF, PG1 = PG2 = PG3 = PG4 = 0 dB)</b>					
<b>Dynamic Performance</b>					
Full-scale input voltage	0 dB	2.828			Vp-p
		1			Vrms
Dynamic range	EIAJ, A-weighted	90			dB
SNR	Signal-to-noise ratio	EIAJ, A-weighted	83	90	dB
Channel separation			87		dB
THD+N	Total harmonic distortion + noise	-1 dB	0.009%	0.017%	
<b>Analog Input</b>					
Center voltage		0.5 $V_{CC}$			V
Input impedance	AIN1L, AIN1R, AIN2L, and AIN2R	10	20		k $\Omega$
	PGINL and PGINR, PG3 = PG4 = -12 dB	70	142		
	PGINL and PGINR, PG3 = PG4 = 30 dB	4.7	9.5		
<b>ANALOG OUTPUTS (AOL AND AOR)</b>					
Center voltage		0.5 $V_{CC}$			V
Load resistance		10			k $\Omega$
Load capacitance		20			pF
<b>MICROPHONE BIAS—ALC = OFF, PG1 = PG2 = PG3 = PG4 = 0 dB</b>					
Bias voltage		0.75 $V_{CC}$			V
Bias source current		2			mA
Output noise		6.5			$\mu\text{V}$

**ELECTRICAL CHARACTERISTICS (continued)**

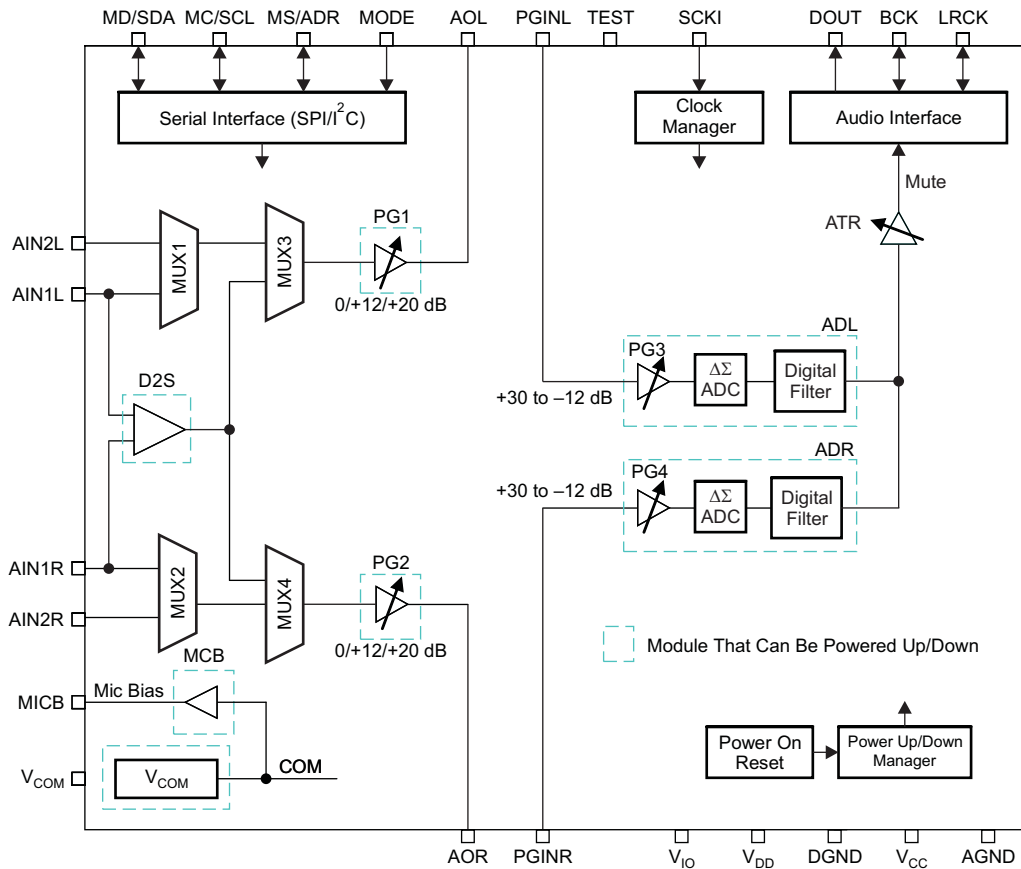
All specifications at  $T_A = 25^\circ\text{C}$ ,  $V_{DD} = V_{IO} = V_{CC} = V_{PA} = 3.3\text{ V}$ ,  $f_S = 48\text{ kHz}$ , system clock =  $256 f_S$ , and 16-bit data, unless otherwise noted

PARAMETER	TEST CONDITIONS	PCM1870AYZF			UNIT	
		MIN	TYP	MAX		
<b>FILTER CHARACTERISTICS</b>						
<b>Decimation Filter for ADC</b>						
Pass band				0.408 $f_S$		
Stop band		0.591 $f_S$				
Pass-band ripple				$\pm 0.02$	dB	
Stop-band attenuation	$f < 3.268 f_S$	-60			dB	
Group delay				$17/f_S$	s	
<b>High-Pass Filter for ADC</b>						
Frequency response ( $f_c = 4\text{ Hz}$ )	-3 dB		3.74		Hz	
	-0.5 dB		10.66			
	-0.1 dB		24.20			
Frequency response ( $f_c = 120\text{ Hz}$ )	-3 dB		118.77		Hz	
	-0.5 dB		321.75			
	-0.1 dB		605.52			
Frequency response ( $f_c = 240\text{ Hz}$ )	-3 dB		235.68		Hz	
	-0.5 dB		609.95			
	-0.1 dB		2601.2			
<b>POWER SUPPLY AND SUPPLY CURRENT</b>						
$V_{IO}$	Voltage range	$V_{IO}$	1.71	3.3	3.6	VDC
$V_{DD}$		$V_{DD}$	1.71	3.3	3.6	
$V_{CC}$		$V_{CC}$	2.4	3.3	3.6	
Supply current		BPZ input, all active, no load		8	12	mA
		All inputs are held static.		1	10	$\mu\text{A}$
Power dissipation		BPZ input		26.4	39.6	mW
		All inputs are held static.		3.3	33	$\mu\text{W}$
<b>TEMPERATURE CONDITION</b>						
Operation temperature			-40		85	$^\circ\text{C}$
$\theta_{JA}$	Thermal resistance			35		$^\circ\text{C/W}$

**PIN ASSIGNMENTS**
**PCM1870AYZF  
(Top View)**

**Table 1. TERMINAL FUNCTIONS**

TERMINAL		I/O	DESCRIPTION
NAME	NO. YZF		
AGND	B4	–	Ground for analog
AIN1L	A2	I	Analog input 1 for L-channel
AIN1R	A1	I	Analog input 1 for R-channel
AIN2L	C4	I	Analog input 2 for L-channel
AIN2R	C3	I	Analog input 2 for R-channel
AOL	C2	O	Microphone amplifier output for L-channel
AOR	C1	O	Microphone amplifier output for R-channel
BCK	F1	I/O	Serial bit clock
DGND	E2	–	Ground for digital
DOUT	F4	O	Serial audio data output
LRCK	E1	I/O	Left- and right-channel clock
MC/SCL	E4	I	Mode control clock for 3-wire / 2-wire interface
MD/SDA	D4	I/O	Mode control data for 3-wire / 2-wire interface
MICB	A3	O	Microphone bias source output
MODE	D2	I	2- or 3-wire interface selection (LOW: SPI, HIGH: I <sup>2</sup> C)
MS/ADR	D3	I	Mode control select for 3-wire / 2-wire interface
PGINL	B2	I	Analog input to gain amplifier for L-channel
PGINR	B1	I	Analog input to gain amplifier for R-channel
SCKI	F2	I	System clock
TEST	D1	I	Test Pin. Should be connected to ground.
V <sub>CC</sub>	B3	–	Power supply for analog
V <sub>COM</sub>	A4	–	Common voltage for analog
V <sub>DD</sub>	F3	–	Power supply for digital core
V <sub>IO</sub>	E3	–	Power supply for digital I/O

**FUNCTIONAL BLOCK DIAGRAM**



B0231-01

**TYPICAL PERFORMANCE CURVES**

All specifications at  $T_A = 25^\circ\text{C}$ ,  $V_{DD} = V_{IO} = V_{CC} = V_{PA} = 3.3\text{ V}$ ,  $f_s = 8\text{ kHz to }48\text{ kHz}$ , system clock =  $256 f_s$  and 16-bit data, unless otherwise noted.

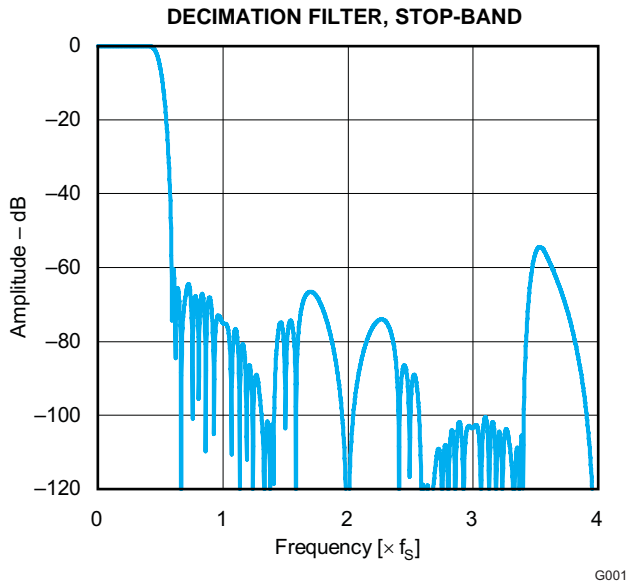


Figure 1.

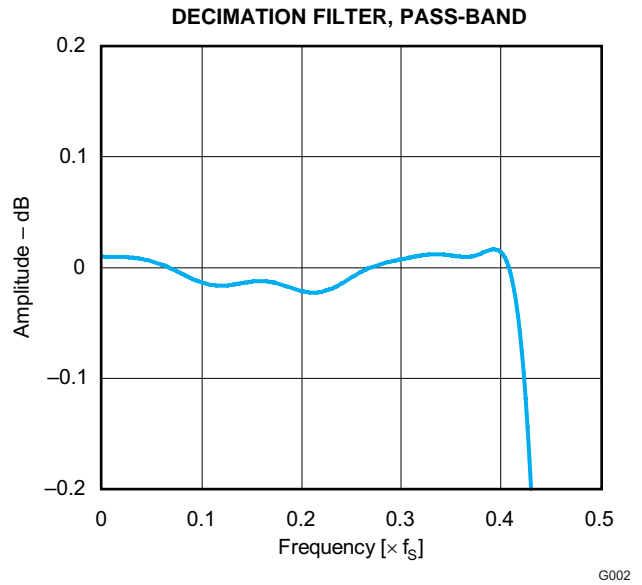


Figure 2.

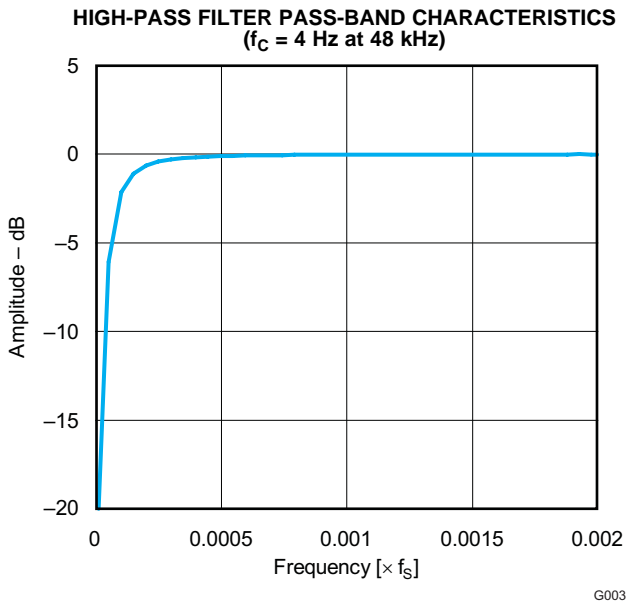


Figure 3.

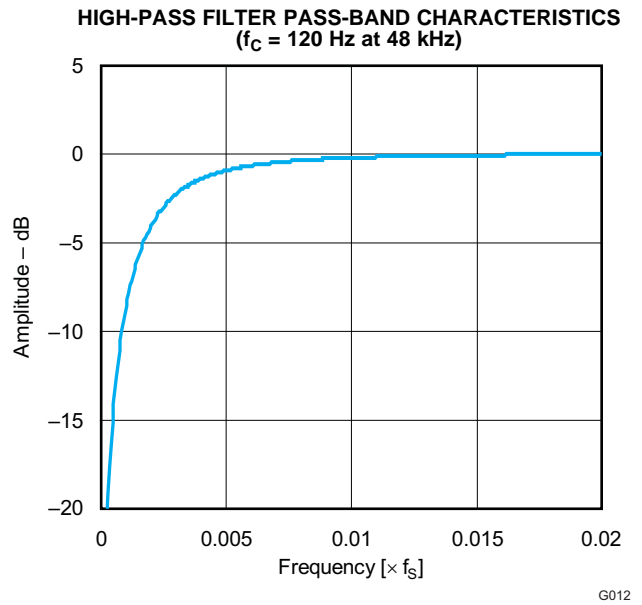
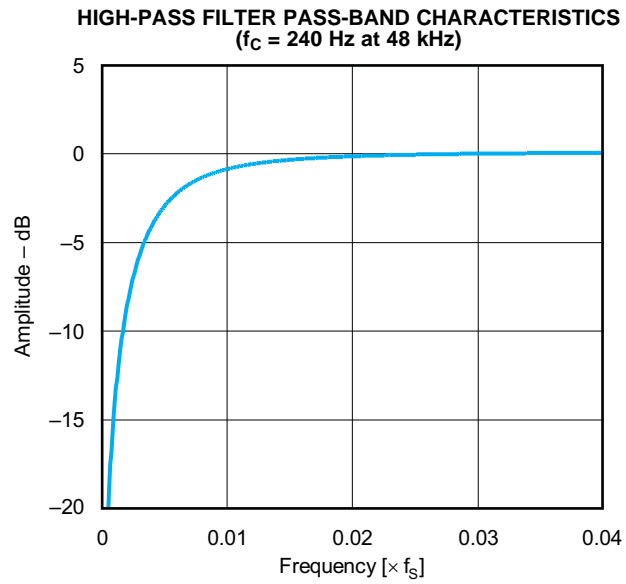


Figure 4.

**TYPICAL PERFORMANCE CURVES (continued)**

All specifications at  $T_A = 25^\circ\text{C}$ ,  $V_{DD} = V_{IO} = V_{CC} = V_{PA} = 3.3\text{ V}$ ,  $f_S = 8\text{ kHz to } 48\text{ kHz}$ , system clock =  $256 f_S$  and 16-bit data, unless otherwise noted.

**Figure 5.**



**TYPICAL PERFORMANCE CURVES (continued)**

All specifications at  $T_A = 25^\circ\text{C}$ ,  $V_{DD} = V_{IO} = V_{CC} = V_{PA} = 3.3\text{ V}$ ,  $f_S = 44.1\text{ kHz}$ , system clock =  $256 f_S$  and 16-bit data, unless otherwise noted.

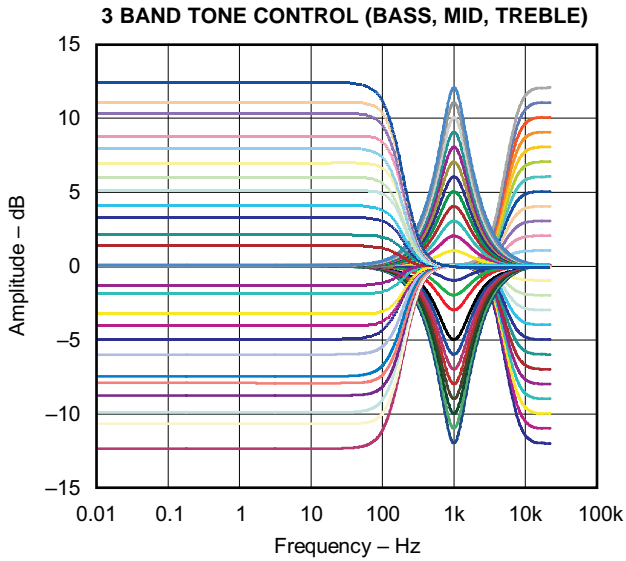


Figure 6.

G005

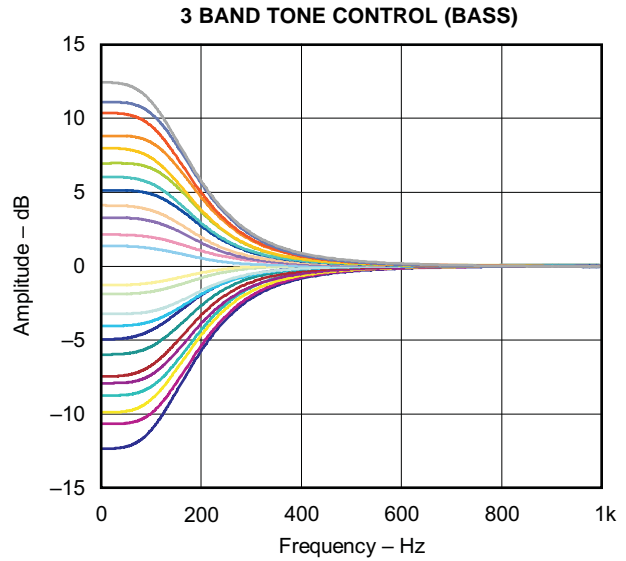


Figure 7.

G006

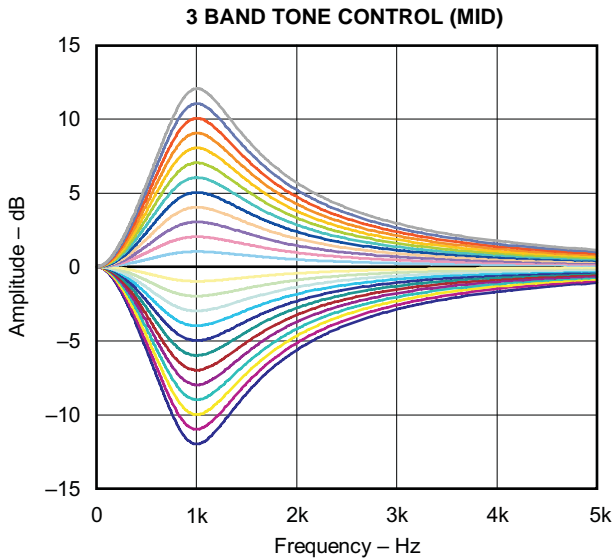


Figure 8.

G007

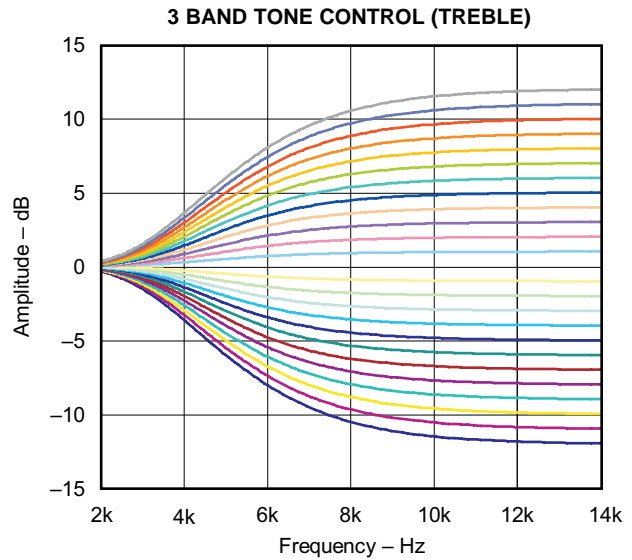


Figure 9.

G008

**TYPICAL PERFORMANCE CURVES (continued)**

All specifications at  $T_A = 25^\circ\text{C}$ ,  $V_{DD} = V_{IO} = V_{CC} = V_{PA} = 3.3\text{ V}$ ,  $f_S = 48\text{ kHz}$ , system clock =  $256 f_S$  and 16-bit data, unless otherwise noted.

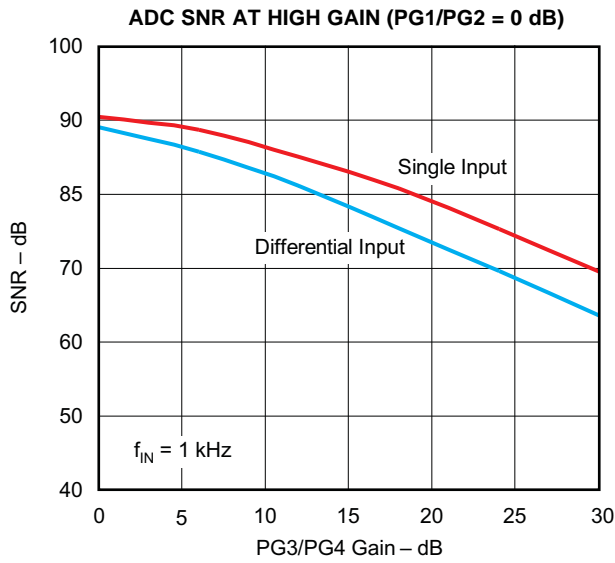


Figure 10.

G009

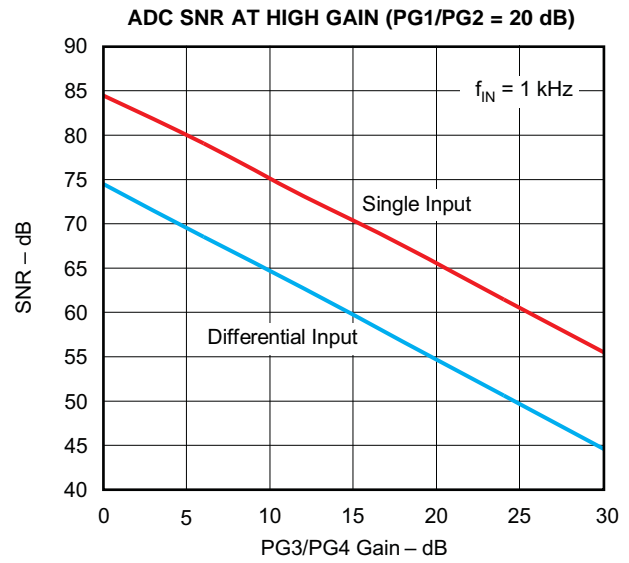


Figure 11.

G010

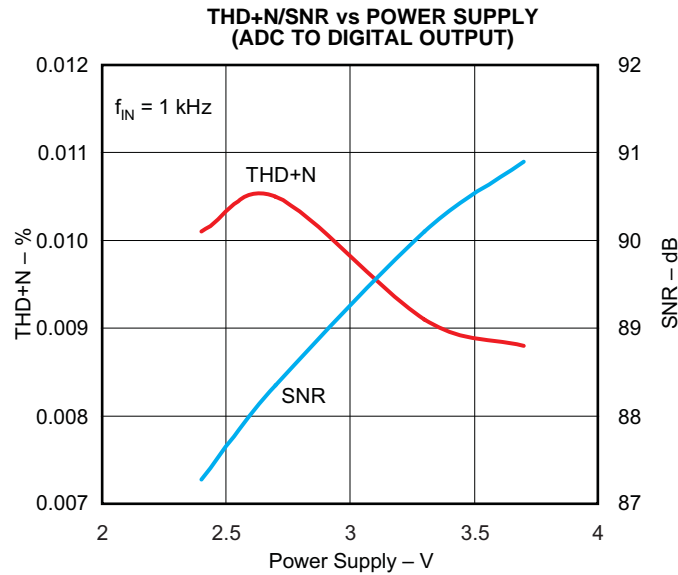


Figure 12.

G011

## PCM1870A DESCRIPTION

### Analog Input

The AIN1L, AIN1R, AIN2L and AIN2R pins can be used as microphone or line inputs with selectable 0- or 20-dB boost and 1-V<sub>rms</sub> input. All analog inputs have high input impedance (20 k $\Omega$ ), which is not changed by gain settings. One pair of inputs is selected by register 87 (AIL[1:0], AIR[1:0]). AIN1L and AIN1R can also be used as a monaural differential input.

### Gain Setting for Analog Input

Analog signals can be adjusted from 30 dB to –12 dB in 1-dB steps after the 0-, 12- or 20-dB boost amplifier. Gain level can be set for each channel by register 79, 80 (ALV[5:0], ARV[5:0]).

### A/D Converter

The ADC includes a multilevel delta-sigma modulator, aliasing filter, decimation filter, high-pass filter and notch filter and can accept 1 V<sub>rms</sub> as full-scale input voltage. The decimation filter has a digital soft mute controlled by register 81 (RMUL, RMUR), and the high-pass and notch filters can be disabled by register 81 (HPF[1:0]) and registers 96 through 104 if they are not needed to cancel dc offset or avoid wind noise.

### Common Voltage

Unbuffered common voltage. The V<sub>COM</sub> pin is normally biased to 0.5 V<sub>CC</sub>, and it provides common voltage to internal circuitry. Connecting a 4.7- $\mu$ F capacitor to this pin is recommended to optimize analog performance.

### Microphone Bias

The MICB pin is a microphone bias source for an external microphone, which can provide 2 mA (typ) bias current.

### Auto Level Control (ALC) for Recording

The sound when microphone recording should be adjusted to a suitable level without saturation. The digitally controlled auto level control (ALC) automatically expands small input signals and compresses large input signals while recording. Expansion level, compression level, attack time, and recovery time can be selected by register 83. See the bit descriptions of [register 83](#) for detailed settings.

### 3D Sound

A 3D sound effect is provided by mixing L-channel and R-channel data through a band-pass filter with two control parameters, mixing ratio and band-pass filter characteristic. The control parameters are set in register 95 (3DP[3:0], 3FLO). The 3D sound effect is applied to the ADC digital output.

### 3-Band Tone Control

Tone control has bass, midrange, and treble controls that can be adjusted from 12 dB to –12 dB in 1-dB steps by register 92 to 94 (LGA[4:0], MGA[4:0] and HGA[4:0]). Register 92 (LPAE), which attenuates the digital input signal automatically, can prevent clipping of the output signal at settings higher than 0 dB for bass control. LPAE is not effective for midrange and treble control.

### High-Pass Filter and Notch Filter

The high-pass filter cuts dc offset in the analog section of the ADC and can be set to 4 Hz, 120 Hz, or 240 Hz at 48-kHz sampling by register 81 (HPF[1:0]).

Notch filters remove noise at particular frequencies, CCD noise, motor noise and other mechanical noise in an application. The PCM1870A has two notch filters, whose center frequency and frequency bandwidth can be programmed by registers 96 to 104.

### Digital Monaural Mixing

The audio data can be mixed to monaural digital data from stereo digital data in the internal audio interface section by register 96 (MXEN).

### Zero-Cross Detection

Zero-cross detection minimizes audible zipper noise while changing analog volume and digital attenuation. This function can be applied to digital input or digital output by register 86 (ZCRS).

### Power Up/Down for Each Module

Using register 73 (PBIS), register 82 (PAIR, PAIL, PADS, PMCB, PADR, PADL), and register 90 (PCOM), unused modules can be powered down to minimize power consumption (13 mW when recording only).

### Digital Interface

All digital I/O pins can interface at various power supply voltages. The  $V_{IO}$  pin can be connected to a 1.71-V to 3.6-V power supply.

### Power Supply

The  $V_{CC}$  pin can be connected to 2.4 V to 3.6 V. The  $V_{DD}$  pin and  $V_{IO}$  pin can be connected to 1.71 to 3.6 V. A different voltage can be applied to each of these pins (for example,  $V_{DD} = 1.8$  V,  $V_{IO} = 3.3$  V).

### High Impedance Mode for LRCK, BCK and DOUT Pins

These pins can be set to high impedance by register 84 (BLHZ, DOHZ) to use LRCK, BCK and DOUT as audio interface bus. If other devices have high impedance mode, audio system can share two or more audio interface signals and reduce audio interface port of main DSP.

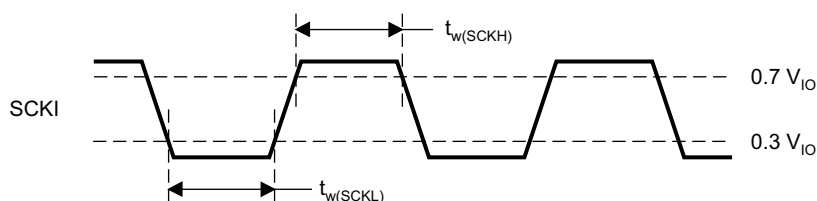
## DESCRIPTION OF OPERATION

### System Clock Input

The PCM1870A can accept input clocks of various frequencies without a PLL. The clocks are used for clocking of the digital filters, auto level control, and delta-sigma modulators, and classified into common-audio and application-specific clocks. Table 2 shows frequencies of the common audio clock and the application-specific clock. Figure 13 shows timing requirements for system clock inputs. The sampling rate and frequency of the system clock are determined by settings in register 86 (MSR[2:0]) and register 85 (NPR[5:0]). Note that the sampling rate of the application-specific clock has a little sampling error. The details are shown in Table 8.

**Table 2. Frequency of Common Audio Clock**

	FREQUENCY
Common audio clock	11.2896, 12.288, 16.9344, 18.432 MHz
Application-specific clock	12, 13, 13.5, 24, 26, 27, 19.2, 19.68, 38.4, 39.36 MHz



T0005-12

PARAMETERS	SYMBOL	MIN	UNIT
System clock pulse duration, high	$t_w(SCKH)$	14	ns
System clock pulse duration, low	$t_w(SCKL)$	14	ns

**Figure 13. System Clock Timing**

### Power-On Reset and System Reset

The power-on-reset circuit outputs reset signal, typically at  $V_{DD} = 1.2 V$ , and this circuit does not depend on the voltage of other power-supplies ( $V_{CC}$ ,  $V_{PA}$ , and  $V_{IO}$ ). Internal circuits are cleared to default status, then signals are removed from all analog and digital outputs. The PCM1870A does not require any power supply sequencing. Register data must be written after turning all power supplies on.

System reset is enabled by setting register 85 (SRST), and all registers are cleared automatically. All circuits are reset to their default status at once. Note that the PCM1870A has audible pop noise on the analog outputs when enabling SRST.

### Power On/Off Sequence

To reduce audible pop noise, a sequence of register settings is required after turning all power supplies on when powering up, or before turning the power supplies off when powering down. If some modules are not required for a particular application or operation, they should be placed in the power-down state after performing the power-on sequence. The recommended power-on and power-off sequences are shown in Table 3 and Table 4, respectively.

**Table 3. Recommended Power-On Sequence**

STEP	REGISTER SETTINGS	NOTE
1	—	Turn ON all power supplies <sup>(1)</sup>
2	5102 <sub>H</sub>	ADC audio interface format (left-justified) <sup>(2)</sup>
3	5A00 <sub>H</sub>	PG1, PG2 gain control (0 dB)
4	4980 <sub>H</sub>	Analog bias power up
5	5601 <sub>H</sub>	Zero-cross detection enable
6	4A01 <sub>H</sub>	V <sub>COM</sub> power up
7	523F <sub>H</sub>	Analog front end (ADL, ADR, D2S, MCB, PG1, 2, 5, 6) power up
8	5711 <sub>H</sub>	Analog input (MUX3, MUX4) select. Analog input (MUX1, MUX2) select
9	4F0C <sub>H</sub>	Analog input L-ch (PG3) volume (0 dB) <sup>(3)</sup>
10	500C <sub>H</sub>	Analog input R-ch (PG4) volume (0 dB) <sup>(3)</sup>

- (1) V<sub>DD</sub> should be turned on first or at the same time with other power supplies. It is recommended to set the register data after turning on all power supplies and while the system clock is running.
- (2) The audio interface format should be set to match the DSP or decoder being used.
- (3) Any level is acceptable for volume or attenuation. The level should return to that recorded in the register data when system was last powered off.

**Table 4. Recommended Power-Off Sequence**

STEP	REGISTER SETTINGS	NOTE
1	5132 <sub>H</sub>	ADC L-ch/R-ch digital soft mute enable, ADC audio interface format (left-justified) <sup>(1)</sup>
2	5200 <sub>H</sub>	Analog front end (ADL, ADR, D2S, MCB, PG1, 2, 5, 6) power down
3	4A00 <sub>H</sub>	V <sub>COM</sub> power down
4	4900 <sub>H</sub>	Analog bias power down
5	—	Turn OFF all power supplies. <sup>(2)</sup>

- (1) The audio interface format should be set to match the DSP or decoder being used.
- (2) Power-supply sequencing is not required. It is recommended to make the required register settings while the system clock is running, then turn off all power supplies.

## Power-Supply Current

The current consumption of the PCM1870A depends on the power-up/down status of each circuit module. In order to save power consumption, disabling each module is recommended when it is not used in an application or operation. [Table 5](#) shows current consumption in some states.

**Table 5. Power Consumption Table**

OPERATION MODE	POWER SUPPLY CURRENT [mA]				PD [mW]	
	V <sub>DD</sub> (1.8 V)	V <sub>DD</sub> (3.3 V)	V <sub>CC</sub> (3.3 V)	V <sub>IO</sub> (3.3 V)	TOTAL (V <sub>DD</sub> = 1.8 V)	TOTAL (V <sub>DD</sub> = 3.3 V)
<b>ALL POWER DOWN</b>	0.000	0.000	0.001	0.000	0.003	0.003
<b>Recording (f<sub>S</sub> = 48 kHz)</b>						
Line input (AIN2L/AIN2R)	1.78	3.71	4.58	0.10	18.3	27.7
Mic input (AIN1L/AIN1R, 20 dB)	1.79	3.71	5.06	0.10	19.9	29.3
Mic input (AIN1L/AIN1R, 20 dB) with ALC	2.73	5.59	5.06	0.10	21.6	35.5
Mono mic input (AIN1L, 20 dB)	1.33	2.80	3.56	0.10	14.1	21.3
Mono mic input (AIN1L, 20 dB) with ALC	2.21	4.60	3.56	0.10	15.7	27.3
Mono diff mic input (AIN1L/AIN1R, 20 dB)	1.33	2.80	3.88	0.10	15.2	22.4
Mono diff mic input (AIN1L/AIN1R, 20 dB) with ALC	2.21	4.60	3.88	0.10	16.8	28.3
<b>Recording (f<sub>S</sub> = 22.05 kHz)</b>						
Line input (AIN2L/AIN2R)	0.82	1.66	3.71	0.10	13.7	18.1
Mic input (AIN1L/AIN1R, 20 dB)	0.82	1.66	4.20	0.10	15.3	19.7
Mic input (AIN1L/AIN1R, 20 dB) with ALC	1.26	2.55	4.20	0.10	16.1	22.6
Mono mic input (AIN1L, 20 dB)	0.61	1.23	2.74	0.10	10.1	13.4
Mono mic input (AIN1L, 20 dB) with ALC	1.03	2.10	2.74	0.10	10.9	1.63
Mono diff mic input (AIN1L/AIN1R, 20 dB)	0.61	1.23	3.06	0.10	11.2	14.5
Mono diff mic input (AIN1L/AIN1R, 20 dB) with ALC	1.02	2.08	3.06	0.10	11.9	17.3
<b>Recording (f<sub>S</sub> = 16 kHz)</b>						
Line input (AIN2L/AIN2R)	0.59	1.18	3.51	0.10	12.7	15.8
Mic input (AIN1L/AIN1R, 20 dB)	0.59	1.18	3.99	0.10	14.2	17.4
Mic input (AIN1L/AIN1R, 20 dB) with ALC	0.91	1.85	3.99	0.10	14.8	19.6
Mono mic input (AIN1L, 20 dB)	0.44	0.87	2.55	0.10	9.2	11.6
Mono mic input (AIN1L, 20 dB) with ALC	0.75	1.52	2.55	0.10	9.8	13.8
Mono diff mic input (AIN1L/AIN1R, 20 dB)	0.44	0.87	2.87	0.10	10.3	12.7
Mono diff mic input (AIN1L/AIN1R, 20 dB) with ALC	0.74	1.50	2.87	0.10	10.8	14.8
<b>Recording (f<sub>S</sub> = 8 kHz)</b>						
Line input (AIN2L/AIN2R)	0.29	0.54	3.23	0.10	11.2	12.8
Mic input (AIN1L/AIN1R, 20 dB)	0.29	0.54	3.72	0.10	12.8	14.4
Mic input (AIN1L/AIN1R, 20 dB) with ALC	0.46	0.88	3.72	0.10	13.1	15.5
Mono mic input (AIN1L, 20 dB)	0.22	0.39	2.29	0.10	8.0	9.2
Mono mic input (AIN1L, 20 dB) with ALC	0.37	0.70	2.29	0.10	8.2	10.2
Mono diff mic input (AIN1L/AIN1R, 20 dB)	0.22	0.39	2.61	0.10	9.0	10.2
Mono diff mic input (AIN1L/AIN1R, 20 dB) with ALC	0.37	0.70	2.61	0.10	9.3	11.3
Condition: 256 f <sub>S</sub> , 16 bits, slave mode, zero data input, no load						

### Audio Serial Interface

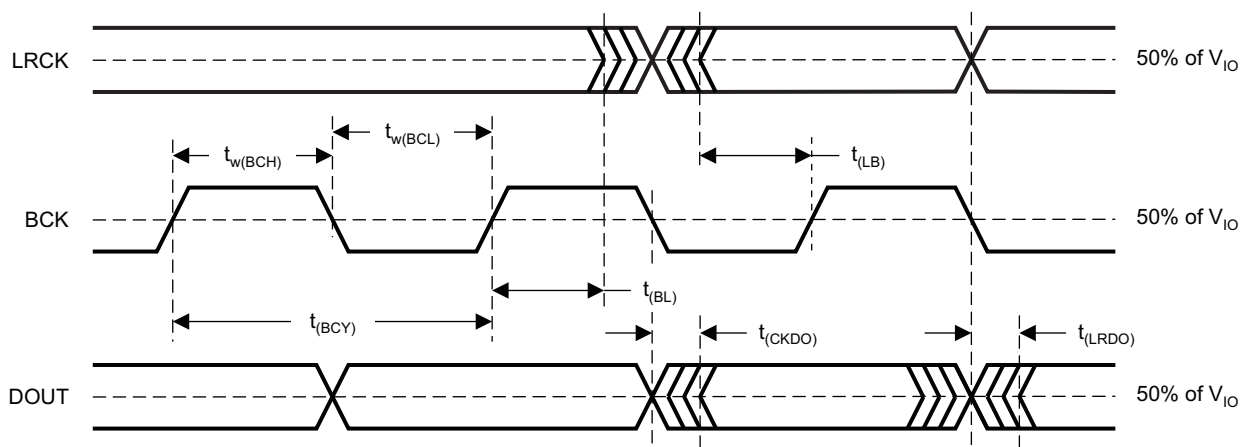
The audio serial interface of the PCM1870A consists of LRCK, BCK and DOUT. Sampling rate ( $f_s$ ), left and right channel are present on LRCK. DOUT transmits the serial data from the decimation filter for the ADC. BCK is used to transmit the serial audio data on DOUT at its high-to-low transition. BCK and LRCK should be synchronized with audio system clock. Ideally, it is recommended that they are derived from it.

The PCM1870A requires LRCK to be synchronized with the system clock. The PCM1870A do not need a specific phase relationship between LRCK and the system clock.

The PCM1870A has both master mode and slave mode interface formats, which can be selected by register 84 (MSTR). LRCK and BCK are generated from the system clock in master mode.

### Audio Data Formats and Timing

The PCM1870A supports I<sup>2</sup>S, right-justified, left-justified, and DSP formats. The data formats are shown in Figure 16, and they are selected using register 70 (RFM[1:0], PFM[1:0]). All formats require binary 2s-complement, MSB-first audio data. The default format is I<sup>2</sup>S. Figure 14 shows a detailed timing diagram.



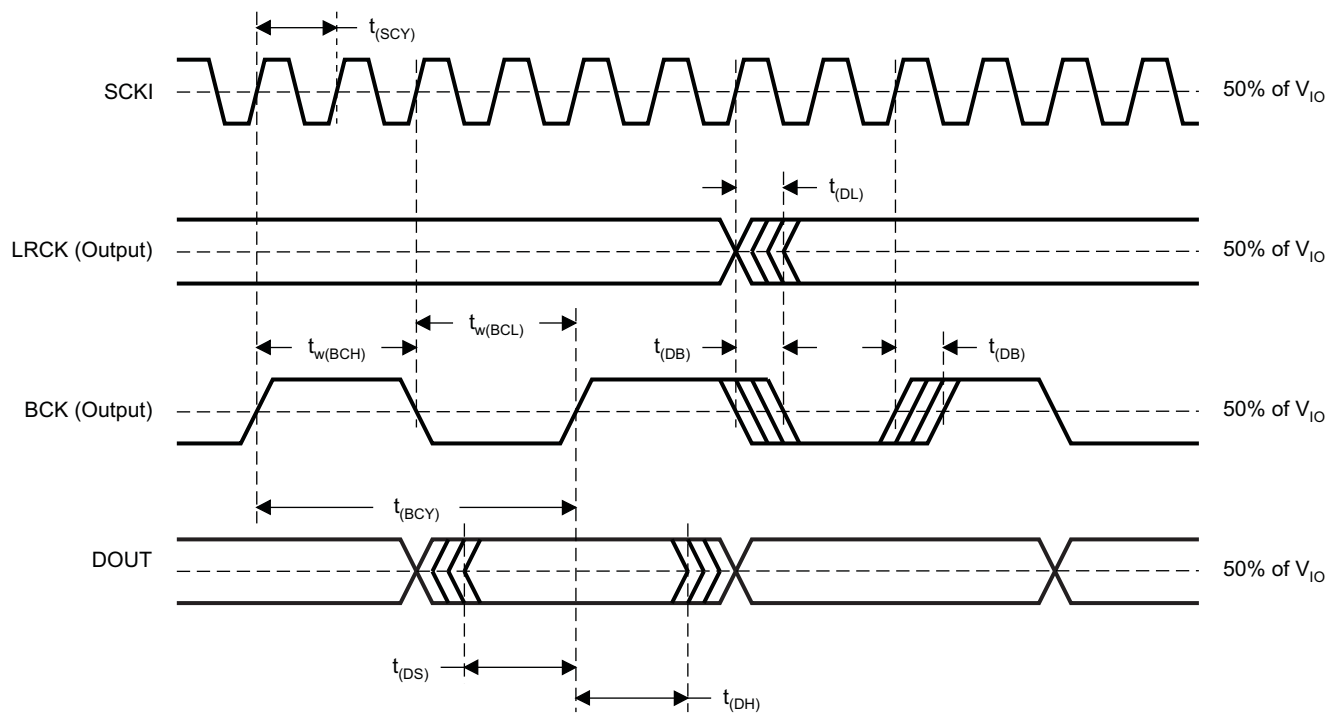
T0010-12

PARAMETERS	SYMBOL	MIN	MAX	UNIT
BCK pulse cycle time (I <sup>2</sup> S, left- and right-justified formats)	$t_{(BCY)}$	$1/(64f_s)^{(1)}$		
BCK pulse cycle time (DSP format)	$t_{(BCY)}$	$1/(256f_s)^{(1)}$		
BCK high-level time	$t_w(BCH)$	35		ns
BCK low-level time	$t_w(BCL)$	35		ns
BCK rising edge to LRCK edge	$t_{(BL)}$	10		ns
LRCK edge to BCK rising edge	$t_{(LB)}$	10		ns
DOUT delay time from BCK falling edge	$t_{(CKDO)}$		40	ns

Figure 14. Audio Interface Timing (Slave Mode)

(1)  $f_s$  is the sampling frequency.





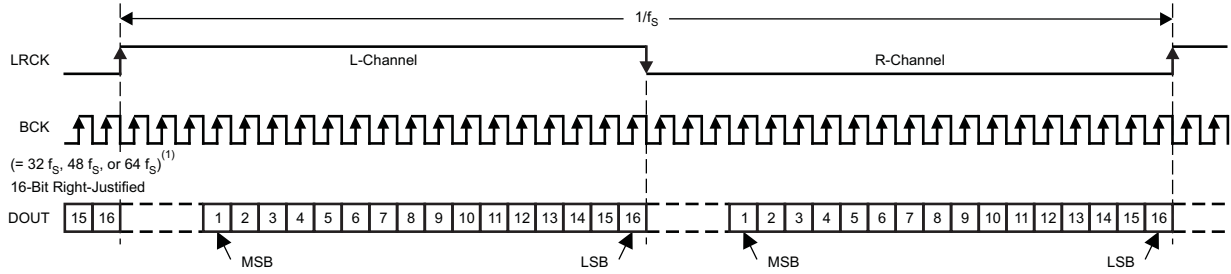
T0011-05

PARAMETERS	SYMBOL	MIN	MAX	UNIT
SCKI pulse cycle time	$t_{(SCY)}$	$1/(256f_S)^{(1)}$		
LRCK edge from SCKI rising edge	$t_{(DL)}$	5	140	ns
BCK edge from SCKI rising edge	$t_{(DB)}$	5	140	ns
BCK pulse cycle time	$t_{(BCY)}$	$1/(64f_S)^{(1)}$		
BCK high-level time	$t_{w(BCH)}$	146		ns
BCK low-level time	$t_{w(BCL)}$	146		ns
DOUT setup time	$t_{(DS)}$	10		ns
DOUT hold time	$t_{(DH)}$	10		ns

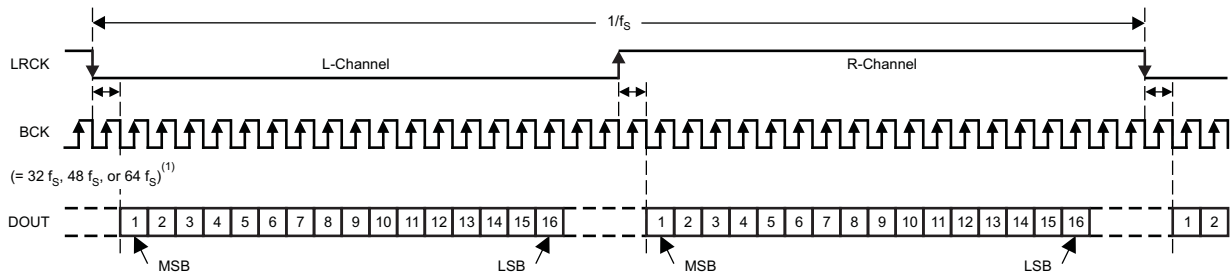
Figure 15. Audio Interface Timing (Master Mode)

(1)  $f_S$  is up to 48 kHz.  $f_S$  is the sampling frequency

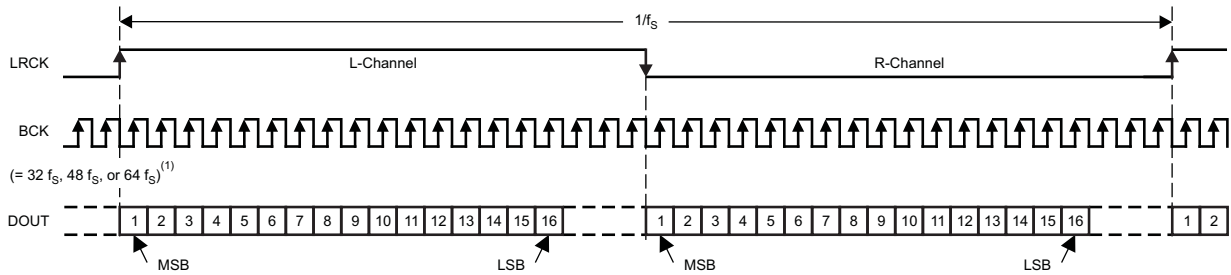
**(a) Right-Justified Data Format; L-Channel = HIGH, R-Channel = LOW**



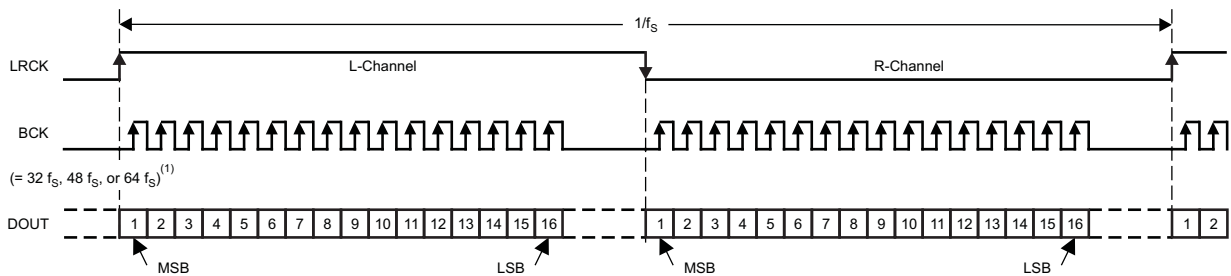
**(b) I<sup>2</sup>S Data Format; L-Channel = LOW, R-Channel = HIGH**



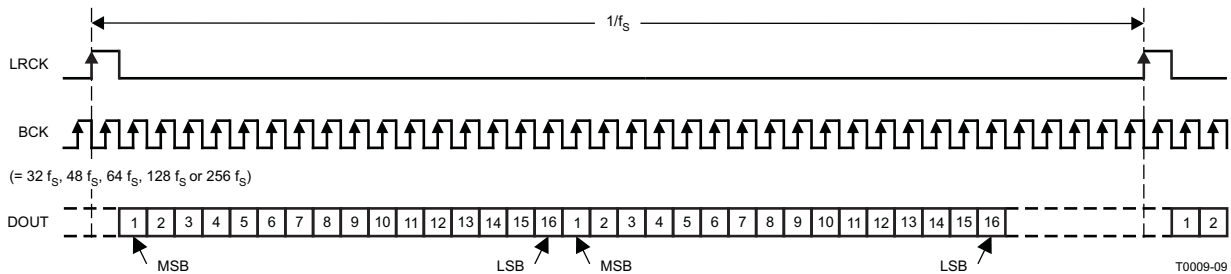
**(c) Left-Justified Data Format; L-Channel = HIGH, R-Channel = LOW**



**(d) Burst BCK Interface Format in Master Mode; L-Channel = HIGH, R-Channel = LOW**



**(e) DSP Format**



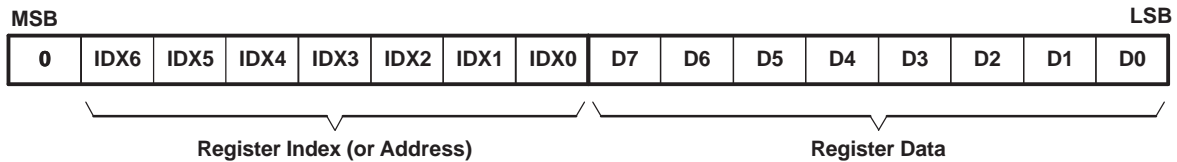
(1) All audio interface formats support BCK = 64 f<sub>S</sub> in master mode (register 69, MSTR = 1). When set to multisampling rate, f<sub>S</sub> of BCK is set to half the rate of the DSM operation frequency.

**Figure 16. Audio Data Output Formats**

### THREE-WIRE INTERFACE (SPI, MODE (PIN 28) = LOW)

All write operations for the serial control port use 16-bit data words. Figure 17 shows the control data word format. The most significant bit must be 0. There are seven bits, labeled  $IDX[6:0]$ , that set the register address for the write operation. The least-significant eight bits,  $D[7:0]$ , contain the data to be written to the register specified by  $IDX[6:0]$ .

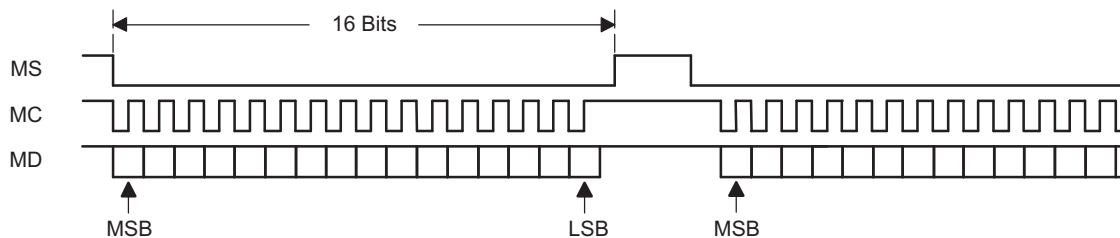
Figure 18 shows the functional timing diagram for writing to the serial control port. To write the data into the mode register, the data is clocked into an internal shift register on the rising edge of the MC clock. The serial data should change on the falling edge of the MC clock and should be LOW during write mode. The rising edge of MS should be aligned with the falling edge of the last MC clock pulse in the 16-bit frame. The MC can run continuously between transactions while MS is in the LOW state.



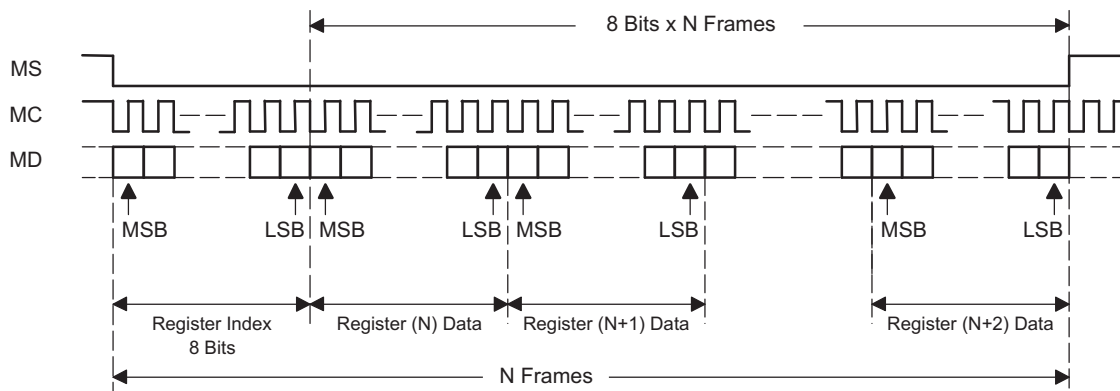
R0001-01

Figure 17. Control Data Word Format for MD

(1) Single Write Operation



(2) Continuous Write Operation

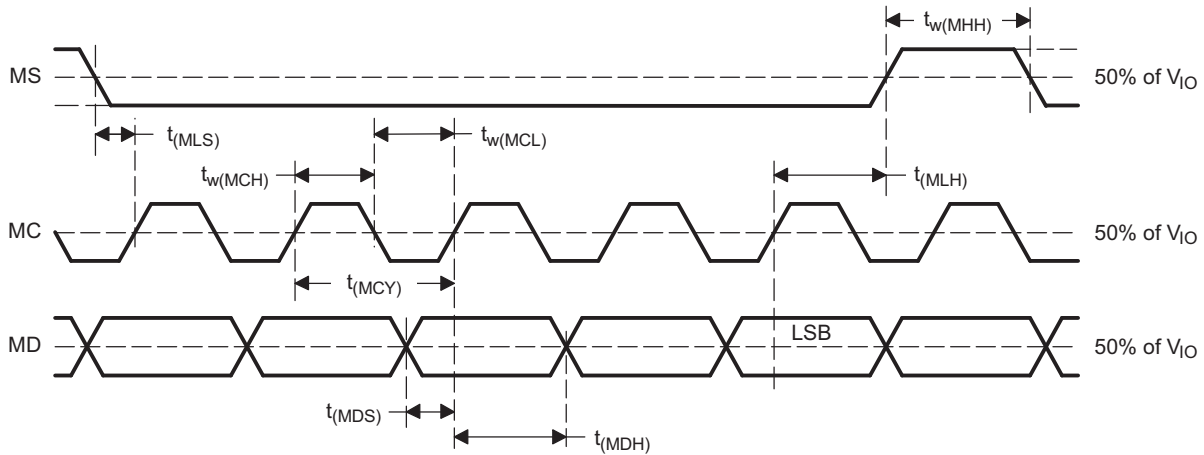


T0012-03

Figure 18. Register Write Operation

### Three-Wire Interface (SPI) Timing Requirements

Figure 19 shows a detailed timing diagram for the serial control interface. These timing parameters are critical for proper control port operation.



T0013-08

PARAMETERS	SYMBOL	MIN	TYP	MAX	UNIT
MC pulse cycle time	t <sub>(MCY)</sub>	500 <sup>(1)</sup>			ns
MC low-level time	t <sub>w(MCL)</sub>	50			ns
MC high-level time	t <sub>w(MCH)</sub>	50			ns
MS high-level time	t <sub>w(MMH)</sub>	See <sup>(1)</sup>			ns
MS falling edge to MC rising edge	t <sub>(MLS)</sub>	50			ns
MS hold time	t <sub>(MLH)</sub>	20			ns
MD hold time	t <sub>(MDH)</sub>	15			ns
MD setup time	t <sub>(MDS)</sub>	20			ns

Figure 19. SPI Interface Timing

(1) 3/(128 f<sub>S</sub>) s (min), where f<sub>S</sub> is the sampling frequency

## TWO-WIRE INTERFACE [I<sup>2</sup>C, MODE (PIN 28) = HIGH]

The PCM1870A supports the I<sup>2</sup>C serial bus and the data transmission protocol for the I<sup>2</sup>C standard as a slave device. This protocol is explained in I<sup>2</sup>C specification 2.0.

In I<sup>2</sup>C mode, the control terminals are changed as follows.

TERMINAL NAME	PROPERTY	DESCRIPTION
MS/ADR	Input	I <sup>2</sup> C address
MD/SDA	Input/output	I <sup>2</sup> C data
MC/SCL	Input	I <sup>2</sup> C clock

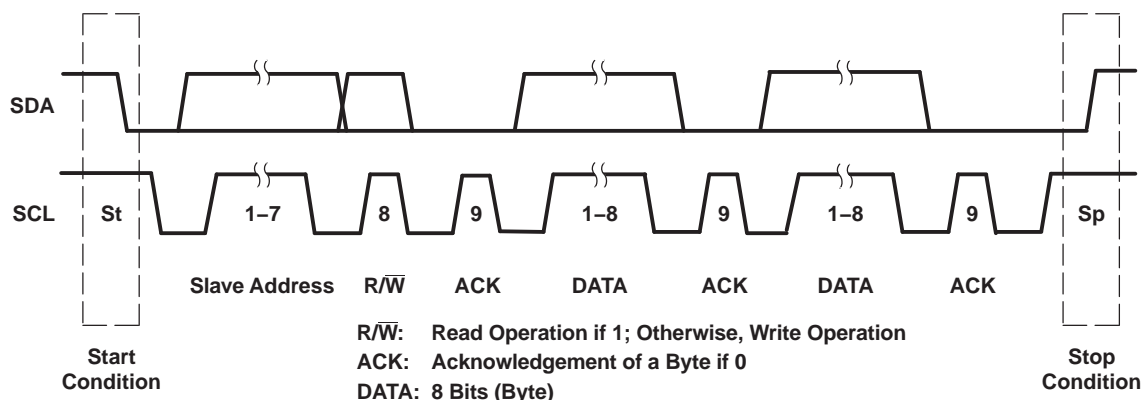
### Slave Address

MSB							LSB	
1	0	0	0	1	1	ADR	R/ $\bar{W}$	

The PCM1870A has its 7-bit slave address. The first six bits (MSBs) of the slave address are factory preset to 1000 11. The next bit of the address byte is the device select bit, which can be user-defined by ADR terminal. A maximum of two PCM1870As can be connected on the same bus at one time. Each PCM1870A responds when it receives its own slave address.

### Packet Protocol

A master device must control packet protocol, which is start condition, slave address with read/write bit, data if write or acknowledgement if read, and stop condition. The PCM1870A supports only slave-receiver and slave-transmitter.



#### Write Operation

Transmitter	M	M	M	S	M	S	M	S	M
Data Type	St	Slave Address	R/ $\bar{W}$	ACK	DATA	ACK	DATA	ACK	Sp

#### Read Operation

Transmitter	M	M	M	S	S	M	S	M	M
Data Type	St	Slave Address	R/ $\bar{W}$	ACK	DATA	ACK	DATA	NACK	Sp

*M:* Master Device      *S:* Slave Device  
*St:* Start Condition    *Sp:* Stop Condition

T0049-03

Figure 20. Basic I<sup>2</sup>C Framework

**Write Operation**

A master can write any PCM1870A registers using single access. The master sends a PCM1870A slave address with a write bit, a register address, and the data. When undefined registers are accessed, the PCM1870A does not send an acknowledgement. The [Figure 21](#) shows a diagram of the write operation.

Transmitter	M	M	M	S	M	S	M	S	M
Data Type	St	Slave Address	$\bar{W}$	ACK	Reg Address	ACK	Write Data	ACK	Sp

M: Master Device S: Slave Device  
 St: Start Condition  $\bar{W}$ : Write ACK: Acknowledge Sp: Stop Condition

R0002-01

**Figure 21. Framework for Write Operation**

**Read Operation**

A master can read the PCM1870A register. The value of the register address is stored in an indirect index register in advance. The master sends a PCM1870A slave address with a read bit after storing the register address. Then the PCM1870A transfers the data which the index register points to. [Figure 22](#) shows a diagram of the read operation.

Transmitter	M	M	M	S	M	S	M	M	S	S	M	M	
Data Type	St	Slave Address	$\bar{W}$	ACK	Reg Address	ACK	Sr	Slave Address	R	ACK	Read Data	NACK	Sp

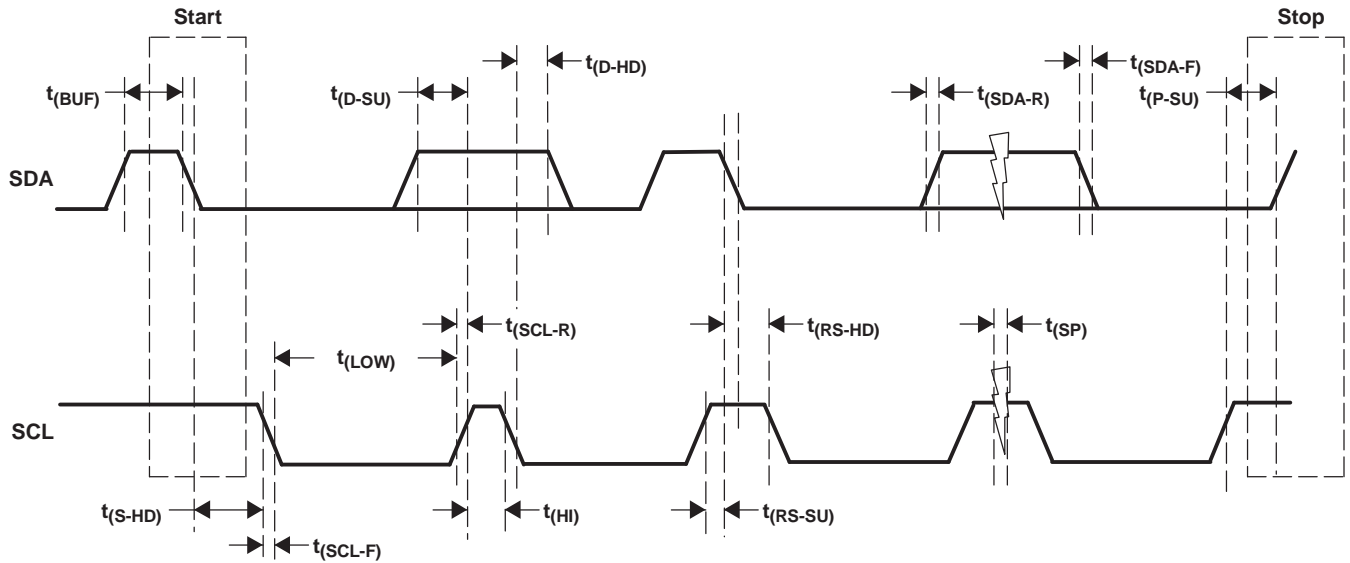
M: Master Device S: Slave Device St: Start Condition  
 Sr: Repeated Start Condition ACK: Acknowledge Sp: Stop Condition NACK: Not Acknowledge  
 $\bar{W}$ : Write R: Read

R0002-02

NOTE: The slave address after the repeated start condition must be the same as the previous slave address.

**Figure 22. Read Operation**

Timing Diagram



T0050-03

TIMING CHARACTERISTICS					
SYMBOL	PARAMETER	CONDITIONS	MIN	MAX	UNIT
$f_{SCL}$	SCL clock frequency	Standard		100	kHz
$t_{(BUF)}$	Bus free time between a STOP and START condition	Standard	4.7		$\mu$ s
$t_{(LOW)}$	Low period of the SCL clock	Standard	4.7		$\mu$ s
$t_{(HI)}$	High period of the SCL clock	Standard	4		$\mu$ s
$t_{(RS-SU)}$	Setup time for START condition	Standard	4.7		$\mu$ s
$t_{(S-HD)}$	Hold time for START condition	Standard	4		$\mu$ s
$t_{(D-SU)}$	Data setup time	Standard	250		ns
$t_{(D-HD)}$	Data hold time	Standard	0	900	ns
$t_{(SCL-R)}$	Rise time of SCL signal	Standard	$20 + 0.1 C_B$	1000	ns
$t_{(SCL-F)}$	Fall time of SCL signal	Standard	$20 + 0.1 C_B$	1000	ns
$t_{(SDA-R)}$	Rise time of SDA signal	Standard	$20 + 0.1 C_B$	1000	ns
$t_{(SDA-F)}$	Fall time of SDA signal	Standard	$20 + 0.1 C_B$	1000	ns
$t_{(P-SU)}$	Setup time for STOP condition	Standard	4		$\mu$ s
$C_B$	Capacitive load for SDA and SCL line			400	pF
$t_{(SP)}$	Pulse duration of spike suppressed			25	ns

Figure 23. I<sup>2</sup>C Interface Timing

## USER-PROGRAMMABLE MODE CONTROLS

### Register Map

The mode control register map is shown in [Table 6](#). Each register includes an index (or address) indicated by the  $IDX[6:0]$  bits.

**Table 6. Mode Control Register Map**

REGISTER	IDX [6:0] (B14–B8)	DESCRIPTION	B7	B6	B5	B4	B3	B2	B1	B0
Register 73	49h	Analog bias power up/down	PBIS	RSV	RSV	RSV	RSV	RSV	RSV	RSV
Register 74	4Ah	$V_{COM}$ power up/down	RSV	RSV	RSV	RSV	RSV	RSV	RSV	PCOM
Register 79	4Fh	Volume for ADC input (L-ch)	RSV	RSV	ALV5	ALV4	ALV3	ALV2	ALV1	ALV0
Register 80	50h	Volume for ADC input (R-ch)	RSV	RSV	ARV5	ARV4	ARV3	ARV2	ARV1	ARV0
Register 81	51h	ADC high pass-filter, soft mute, audio interface	HPF1	HPF0	RMUL	RMUR	RSV	DSMC	RFM1	RFM0
Register 82	52h	ADC, MCB, PG1, 2, 5, 6, D2S power up/down	RSV	RSV	PAIR	PAIL	PADS	PMCB	PADR	PADL
Register 83	53h	Auto level control for recording	RALC	RSV	RRTC	RATC	RCP1	RCP0	RLV1	RLV0
Register 84	54h	DOUT, BCK, and LRCK config, bit, master mode	RSV	RSV	RSV	BLHZ	DOHZ	MSTR	RSV	BIT0
Register 85	55h	System reset, sampling rate control	SRST	RSV	NPR5	NPR4	NPR3	NPR2	NPR1	NPR0
Register 86	56h	BCK config, master mode, zero cross	MBST	MSR2	MSR1	MSR0	RSV	RSV	RSV	ZCRS
Register 87	57h	Analog input select (MUX1, 2, 3, 4)	AD2S	RSV	AIR1	AIR0	RSV	RSV	AIL1	AIL0
Register 90	5Ah	Boost	RSV	RSV	RSV	RSV	RSV	RSV	G20R	G20L
Register 92	5Ch	Bass boost gain level (200 Hz)	LPAE	RSV	RSV	LGA4	LGA3	LGA2	LGA1	LGA0
Register 93	5Dh	Middle boost gain level (1 kHz)	RSV	RSV	RSV	MGA4	MGA3	MGA2	MGA1	MGA0
Register 94	5Eh	Treble boost gain level (5 kHz)	RSV	RSV	RSV	HGA4	HGA3	HGA2	HGA1	HGA0
Register 95	5Fh	Sound effect source select, 3D sound	RSV	3DEN	RSV	3FL0	3DP3	3DP2	3DP1	3DP0
Register 96	60h	2-stage notch filter, digital monaural mixing	NEN2	NEN1	NUP2	NUP1	RSV	RSV	RSV	MXEN
Register 97	61h	1st-stage notch filter lower coefficient (a1)	F107	F106	F105	F104	F103	F102	F101	F100
Register 98	62h	1st-stage notch filter upper coefficient (a1)	F115	F114	F113	F112	F111	F110	F109	F108
Register 99	63h	1st-stage notch filter lower coefficient (a2)	F207	F206	F205	F204	F203	F202	F201	F200
Register 100	64h	1st-stage notch filter upper coefficient (a2)	F215	F214	F213	F212	F211	F210	F209	F208
Register 101	65h	2nd-stage notch filter lower coefficient (a1)	S107	S106	S105	S104	S103	S102	S101	S100
Register 102	66h	2nd-stage notch filter upper coefficient (a1)	S115	S114	S113	S112	S111	S110	S109	S108
Register 103	67h	2nd-stage notch filter lower coefficient (a2)	S207	S206	S205	S204	S203	S202	S201	S200
Register 104	68h	2nd-stage notch filter upper coefficient (a2)	S215	S214	S213	S212	S211	S210	S209	S208
Register 124	7Ch	Mic boost	RSV	RSV	RSV	RSV	RSV	RSV	G12R	G12L

ADC: A/D converter    MCB: Microphone bias  
PGx: Analog input buffer    D2S: Differential to single-ended amplifier



**Register 73**

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 73	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	PBIS	RSV	RSV	RSV	RSV	RSV	RSV	RSV

**IDX[6:0]:** 100 1001b (49h) Register 73

**PBIS: Power Up/Down Control for Bias**

Default value: 0

This bit is used to control power up/down for the analog bias circuit.

PBIS = 0	Power down (default)
PBIS = 1	Power up

**Register 74**

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 74	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	RSV	RSV	RSV	RSV	RSV	PCOM

**IDX[6:0]:** 100 1010b (4Ah) Register 74

**PCOM: Power Up/Down Control for  $V_{COM}$** 

Default value: 0

This bit is used to control power up/down for  $V_{COM}$ .

PCOM = 0	Power down (default)
PCOM = 1	Power up

**Register 79 and 80**

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 79	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	ALV5	ALV4	ALV3	ALV2	ALV1	ALV0
Register 80	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	ARV5	ARV4	ARV3	AR2	ARV1	ARV0

**IDX[6:0]:** 100 1111b (4Fh) Register 79

**IDX[6:0]:** 101 0000b (50h) Register 80

**ALV[5:0]:** Gain Control for PG3 (ADC Analog Input R-Channel)

**ARV[5:0]:** Gain Control for PG4 (ADC Analog Input L-Channel)

Default value: 00

PG3 and PG4 can be independently controlled for ADC input from 30 dB to –12 dB in 1-dB steps. ADC output may have zipper noise when changing levels. In the PCM1870A, the noise can be reduced when making the change by using zero-cross detection (Register 85, ZCRS).

**Table 7. Gain Level Setting**

ALV[5:0], ARV[5:0]		GAIN LEVEL SETTING	ALV[5:0], ARV[5:0]		GAIN LEVEL SETTING
BINARY	HEX		BINARY	HEX	
10 1010	2A	30 dB	01 0100	14	8 dB
10 1001	29	29 dB	01 0011	13	7 dB
10 1000	28	28 dB	01 0010	12	6 dB
10 0111	27	27 dB	01 0001	11	5 dB
10 0110	26	26 dB	01 0000	10	4 dB
10 0101	25	25 dB	00 1111	0F	3 dB
10 0100	24	24 dB	00 1110	0E	2 dB
10 0011	23	23 dB	00 1101	0D	1 dB
10 0010	22	22 dB	00 1100	0C	0 dB
10 0001	21	21 dB	00 1011	0B	–1 dB
10 0000	20	20 dB	00 1010	0A	–2 dB
01 1111	1F	19 dB	00 1001	09	–3 dB
01 1110	1E	18 dB	00 1000	08	–4 dB
01 1101	1D	17 dB	00 0111	07	–5 dB
01 1100	1C	16 dB	00 0110	06	–6 dB
01 1011	1B	15 dB	00 0101	05	–7 dB
01 1010	1A	14 dB	00 0100	04	–8 dB
01 1001	19	13 dB	00 0011	03	–9 dB
01 1000	18	12 dB	00 0010	02	–10 dB
01 0111	17	11 dB	00 0001	01	–11 dB
01 0110	16	10 dB	00 0000	00	–12 dB (default)
01 0101	15	9 dB			

**Register 81**

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 81	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	HPF1	HPF0	RMUL	RMUR	RSV	DSMC	RFM1	RFM0

**IDX[6:0]:** 101 0001b (51h) Register 81

**HPF[1:0]: High-Pass Filter Selection**

Default value: 00

The PCM1870A has a digital high-pass filter to remove dc voltage at the input of the ADC. The cutoff frequency of the high-pass filter can be selected.

HPF[1:0]	High Pass Filter Selection
0 0	$f_C = 4$ Hz at 48 kHz (default)
0 1	$f_C = 240$ Hz at 48 kHz
1 0	$f_C = 120$ Hz at 48 kHz
1 1	High-pass filter disabled

**RMUL: Digital Soft Mute Control for ADC L-Channel**
**RMUR: Digital Soft Mute Control for ADC R-Channel**

Default value: 1

The digital output of the ADC can be independently muted or unmuted. The transition from the current volume level to mute, or the return to the previous volume setting from mute, occurs at the rate of one 1-dB step for each  $8/f_S$  time period. When RMUL and RMUR = 0, the digital data is increased from mute to the previous attenuation level, and when RMUL and RMUR = 1, the digital data is decreased from the current attenuation level to mute. In the PCM1870A, audible zipper noise can be reduced by using zero-cross detection (register 85, ZCRS).

RMUL, RMUR = 0	Mute disabled
RMUL, RMUR = 1	Mute enabled (default)

**DSMC: Waiting Time Turn ADC Mute Off at Power Up**

Default value: 0

ADC digital output has waiting time at power up when DSMC = 0. It is recommended to set DSMC = 0.

DSMC = 0	10 ms at 48 kHz (default)
DSMC = 1	No wait

**RFM[1:0]: Audio Interface Selection for ADC (Digital Output)**

Default value: 00

The audio interface format for ADC digital output has I<sup>2</sup>S, right-justified, left-justified, and DSP formats.

RFM[1:0]	Audio Interface Selection for ADC Digital Output
0 0	I <sup>2</sup> S (default)
0 1	Right-justified
1 0	Left-justified
1 1	DSP mode

**Register 82**

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 82	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	PAIR	PAIL	PADS	PMCB	PADR	PADL

**IDX[6:0]:** 101 0010b (52h) Register 82

**PAIR: Power Up/Down for PG2 and PG6 (Gain Amplifier for Analog Input R-Channel)**

**PAIR: Power Up/Down for PG1 and PG5 (Gain Amplifier for Analog Input L-Channel)**

Default value: 0

This bit is used to control power up/down for PG1, -2 and PG5, -6 (gain amplifier for analog input).

PAIR, PAIL = 0	Power down (default)
PAIR, PAIL = 1	Power up

**PADS: Power Up/Down for D2S (Differential Amplifier) of AIN1L and AIN1R**

Default value: 0

This bit is used to control power up/down for D2S (differential-to-single amplifier).

PADS = 0	Power down (default)
PADS = 1	Power up

**PMCB: Power Up/Down Control for Microphone Bias Source**

Default value: 0

This bit is used to control power up/down for the microphone bias source.

PMCB = 0	Power down (default)
PMCB = 1	Power up

**PADR: Power Up/Down Control for ADR (ADC and Digital Filter R-Channel)**

**PADL: Power Up/Down Control for ADL (ADC and Digital Filter L-Channel)**

Default value: 0

This bit is used to control power up/down for the ADC and decimation filter.

PADR, PADL = 0	Power down (default)
PADR, PADL = 1	Power up

**Register 83**

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 83	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RALC	RSV	RRTC	RATC	RCP1	RCP0	RLV1	RLV0

**IDX[6:0]:** 1010011b (53h) Register 83

**RALC: Automatic Level Control (ALC) Enable for Recording**

Default value: 0

Auto level control can be enabled with some parameters for microphone input or lower-level analog source.

RALC = 0	Disabled (default)
RALC = 1	Enabled

**RRTC: ALC Recovery Time Control for Recording**

Default value: 0

This bit selects the time during which a gain level change completes to compress the signal when the input to the ADC increases in amplitude.

RRTC = 0	3.4 s (default)
RRTC = 1	13.6 s

**RATC: ALC Attack Time Control for Recording**

Default value: 0

This bit selects the time during which a gain level change completes to expand the signal when the input to the ADC decreases in amplitude.

RATC = 0	1 ms (default)
RATC = 1	2 ms

**RCP[1:0]: ALC Compression Level Control for Recording**

Default value: 00

Auto level control can set the compression level to –2, –6, or –12 dB. Higher-level signals should be compressed to avoid saturation or degradation of sound quality.

RCP[1:0]	ALC Compression Level Control for Recording
0 0	–2 dB (default)
0 1	–6 dB
1 0	–12 dB
1 1	Reserved

**RLV[1:0]: ALC Expansion Level Control for Recording**

Default value: 00

Auto level control can set the expansion level to 0, 6, 14, or 24 dB. Lower-level signals should be expanded to make a small signal easy to hear. If set to 0 dB, the ALC can be operated only as a limiter.

RLV[1:0]	ALC Expansion Level Control for Recording
0 0	0 dB (default)
0 1	6 dB
1 0	14 dB
1 1	24 dB

**Register 84–86**

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 84	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	RSV	BLHZ	DOHZ	MSTR	RSV	BIT0
Register 85	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	SRST	RSV	NPR5	NPR4	NPR3	NPR2	NPR1	NPR0
Register 86	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	MBST	MSR2	MSR1	MSR0	RSV	RSV	RSV	ZCRS

**IDX[6:0]:** 101 0100b (54h) Register 84

**IDX[6:0]:** 101 0101b (55h) Register 85

**IDX[6:0]:** 101 0110b (56h) Register 86

**BLHZ: BCK and LRCK Output Configuration**

Default value: 0

This bit is used for control output configuration of LRCK and BCK pin. When BLHZ = 1, the application system can connect some devices to these signal lines.

BLHZ = 0	Output at master mode or Input at slave mode (default)
BLHZ = 1	High Impedance

**DOHZ: DOUT Output Configuration**

Default value: 0

This bit is used for control output configuration of DOUT pin. When DOHZ = 1, the application system can connect some devices to this signal line.

DOHZ = 0	Output (default)
DOHZ = 1	High Impedance

**MSTR: Master or Slave Selection for Audio Interface**

Default value: 0

This bit is used to select either master or slave mode for the audio interface. In master mode, PCM1870A generates LRCK and BCK from the system clock. In slave mode, it receives LRCK and BCK from another device.

MSTR = 0	Slave interface (default)
MSTR = 1	Master interface

**BIT0: Bit Length Selection for Audio Interface**

Default value: 1

This bit is used to select data bit length for the ADC output.

BIT0 = 0	Reserved
BIT0 = 1	16 bits (default)

**SRST: System Reset**

Default value: 0

This bit is used to enable system reset. All circuits are reset by setting SRST = 1. After completing the reset sequence, SRST resets to 0 automatically.

SRST = 0	Reset disabled (default)
SRST = 1	Reset enabled

**NPR[5:0]: System Clock Rate Selection**

Default value: 00 0000

These bits are used to select the system clock rate. See [Table 8](#) for details.

**MBST: BCK Output Configuration in Master Mode**

Default value: 0

This bit is used to control the BCK output configuration in master mode.  $V_{IO}$  (I/O cell power supply) power consumption can be reduced by adjusting BCK edge to bit number when setting MBST = 1. This is effective in master mode (register 69 MSTR = 1).

MBST = 0	Normal output (default)
MBST = 1	Burst output

**MSR[2:0]: System Clock Dividing Rate Selection in Master Mode (Register 70)**

Default value: 000

 These bits are used to set the dividing rate of the input system clock. See [Table 8](#) for details.

**Table 8. System Clock Frequency for Common Audio Clock**

SYSTEM CLOCK SCK (MHz)	ADC SAMPLING RATE ADC $f_s$ (kHz)	DAC SAMPLING RATE DAC $f_s$ (kHz)	REGISTER SETTING		BIT CLOCK BCK (fs)
			MSR[2:0]	NPR[5:0]	
6.144	24 (SCK/256)		010	00 0000	64
	16 (SCK/384)		011	00 0000	64
	12 (SCK/512)		100	00 0000	64
	8 (SCK/768)		101	00 0000	64
	6 (SCK/1024)		110	00 0000	64
	4 (SCK/1536)		111	00 0000	64
8.192	32 (SCK/256)		010	00 0000	64
	16 (SCK/512)		100	00 0000	64
	8 (SCK/1024)		110	00 0000	64
12.288	48 (SCK/256)		010	00 0000	64
	32 (SCK/384)		011	00 0000	64
	24 (SCK/512)		100	00 0000	64
	16 (SCK/768)		101	00 0000	64
	12 (SCK/1024)		110	00 0000	64
	8 (SCK/1536)		111	00 0000	64
18.432	48 (SCK/384)		011	00 0000	64
	24 (SCK/768)		101	00 0000	64
	12 (SCK/1536)		111	00 0000	64
5.6448	22.05 (SCK/256)		010	00 0000	64
	14.7 (SCK/384)		011	00 0000	64
	11.025 (SCK/512)		100	00 0000	64
	7.35 (SCK/768)		101	00 0000	64
	5.5125 (SCK/1024)		110	00 0000	64
	3.675 (SCK/1536)		111	00 0000	64
11.2896	44.1 (SCK/256)		010	00 0000	64
	29.4 (SCK/384)		011	00 0000	64
	22.05 (SCK/512)		100	00 0000	64
	14.7 (SCK/768)		101	00 0000	64
	11.025 (SCK/1024)		110	00 0000	64
	7.35 (SCK/1536)		111	00 0000	64

NOTE: Other settings are reserved.





**Table 9. System Clock Frequency for Application-Specific Audio Clock**

SYSTEM CLOCK SCK (MHz)	ADC SAMPLING RATE ADC $f_s$ (kHz)	DAC SAMPLING RATE DAC $f_s$ (kHz)	REGISTER SETTING		BIT CLOCK BCK ( $f_s$ )
			MSR[2:0]	NPR[5:0]	
13.5	48.214 (SCK/280)		010	00 0010	70
	44.407 (SCK/304)		010	00 0001	76
	32.142 (SCK/420)		010	10 0010	70
	24.107 (SCK/560)		100	00 0010	70
	22.203 (SCK/608)		100	00 0001	76
	16.071 (SCK/840)		100	10 0010	70
	12.053 (SCK/1120)		110	00 0010	70
	8.035 (SCK/1680)		110	10 0010	70
27	48.214 (SCK/560)		010	01 0010	70
	44.407 (SCK/608)		010	01 0001	76
	32.142 (SCK/840)		010	11 0010	70
	24.107 (SCK/1120)		100	01 0010	70
	22.203 (SCK/1216)		100	01 0001	76
	16.071 (SCK/1680)		100	11 0010	70
	12.053 (SCK/2240)		110	01 0010	70
	8.035 (SCK/3360)		110	11 0010	70
12	48.387 (SCK/248)		010	00 0100	62
	44.117 (SCK/272)		010	00 0011	68
	32.258 (SCK/372)		010	10 0100	62
	24.193 (SCK/496)		100	00 0100	62
	22.058 (SCK/544)		100	00 0011	68
	16.129 (SCK/744)		100	10 0100	62
	12.096 (SCK/992)		110	00 0100	62
	8.064 (SCK/1488)		110	10 0100	62
24	48.387 (SCK/496)		010	01 0100	62
	44.117 (SCK/544)		010	01 0011	68
	32.258 (SCK/744)		010	11 0100	62
	24.193 (SCK/992)		100	01 0100	62
	22.058 (SCK/1088)		100	01 0011	68
	16.129 (SCK/1488)		100	11 0100	62
	12.096 (SCK/1984)		110	01 0100	62
	8.064 (SCK/2796)		110	11 0100	62
19.2	48.484 (SCK/396)		011	00 0110	66
	44.444 (SCK/432)		011	00 0101	72
	32.323 (SCK/594)		011	10 0110	66
	24.242 (SCK/792)		101	00 0110	66
	22.222 (SCK/864)		101	00 0101	72
	16.161 (SCK/1188)		101	10 0110	66
	12.121 (SCK/1584)		111	00 0110	66
	8.080 (SCK/2376)		111	10 0110	66

**Table 9. System Clock Frequency for Application-Specific Audio Clock (continued)**

SYSTEM CLOCK SCK (MHz)	ADC SAMPLING RATE ADC $f_s$ (kHz)	DAC SAMPLING RATE DAC $f_s$ (kHz)	REGISTER SETTING		BIT CLOCK BCK ( $f_s$ )
			MSR[2:0]	NPR[5:0]	
38.4	48.484 (SCK/792)		011	01 0110	66
	44.444 (SCK/864)		011	01 0101	72
	32.323 (SCK/1188)		011	11 0110	66
	24.242 (SCK/1584)		101	01 0110	66
	22.222 (SCK/1728)		101	01 0101	72
	16.161 (SCK/2376)		101	11 0110	66
	12.121 (SCK/3168)		111	01 0110	66
	8.080 (SCK/4752)		111	11 0110	66
13	47.794 (SCK/272)		010	00 1000	68
	43.918 (SCK/296)		010	00 0111	74
	31.862 (SCK/408)		010	10 1000	68
	23.897 (SCK/544)		100	00 1000	68
	21.959 (SCK/592)		100	00 0111	74
	15.931 (SCK/816)		100	10 1000	68
	11.948 (SCK/1088)		110	00 1000	68
	7.965 (SCK/1632)		110	10 1000	68
26	47.794 (SCK/544)		010	01 1000	68
	43.918 (SCK/592)		010	01 0111	74
	31.862 (SCK/816)		010	11 1000	68
	23.897 (SCK/1088)		100	01 1000	68
	21.959 (SCK/1184)		100	01 0111	74
	15.931 (SCK/1632)		100	11 1000	68
	11.948 (SCK/2176)		110	01 1000	68
	7.965 (SCK/3264)		110	11 1000	68
19.68	48.235 (SCK/408)		011	00 1010	68
	44.324 (SCK/444)		011	00 1001	74
	32.156 (SCK/612)		011	10 1010	68
	24.117 (SCK/816)		101	00 1010	68
	22.162 (SCK/888)		101	00 1001	74
	16.078 (SCK/1224)		101	10 1010	68
	12.058 (SCK/1632)		111	00 1010	68
	8.039 (SCK/2448)		111	10 1010	68
39.36	48.235 (SCK/816)		011	01 1010	68
	44.324 (SCK/888)		011	01 1001	74
	32.156 (SCK/1224)		011	11 1010	68
	24.117 (SCK/1632)		101	01 1010	68
	22.162 (SCK/1776)		101	01 1001	74
	16.078 (SCK/2448)		101	11 1010	68
	12.058 (SCK/3264)		111	01 1010	68
	8.039 (SCK/4896)		111	11 1010	68

### ZCRS: Zero Cross for Digital Attenuation/Mute and Analog Gain Setting

Default value: 0

This bit is used to enable the zero-cross detector, which reduces zipper noise while the digital soft mute or analog gain setting is being changed. If no zero-cross data is input for a  $512/f_s$  period (10.6 ms at a 48-kHz sampling rate), then a time-out occurs and the PCM1870A starts changing the attenuation, gain, or volume level. The zero-cross detector cannot be used with continuous-zero and dc data.

ZCRS = 0	Zero cross disabled (default)
ZCRS = 1	Zero cross enabled

### Register 87

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 87	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	AD2S	RSV	AIR1	AIR0	RSV	RSV	AIL1	AIL0

**IDX[6:0]:** 101 0111b (57h) Register 87

### AD2S: Differential Amplifier Selector (MUX3 and MUX4)

Default value: 0

This bit is used to select whether a single-ended amplifier or differential amplifier (D2S) is used as the input for the ADC. MUX3 and MUX4 use the single-ended input when AD2S = 0. MUX3 and MUX4 use the monaural differential input when AD2S = 1.

AD2S = 0	Single-ended amplifier (default)
AD2S = 1	Differential amplifier

### AIL[1:0]: AIN1L and AIN2L Selector (MUX1)

Default value: 00

MUX1 selects the analog input, AIN1L or AIN2L.

AIL[1:0]	AIN L-channel Select
0 0	Disconnect (default)
0 1	AIN1L
1 0	AIN2L
1 1	Reserved

### AIR[1:0]: AIN1R and AIN2R Selector (MUX2)

Default value: 00

MUX2 selects the analog input, AIN1R or AIN2R.

AIR[1:0]	AIN R-channel Select
0 0	Disconnect (default)
0 1	AIN1R
1 0	AIN2R
1 1	Reserved

**Register 90**

	<b>B15</b>	<b>B14</b>	<b>B13</b>	<b>B12</b>	<b>B11</b>	<b>B10</b>	<b>B9</b>	<b>B8</b>	<b>B7</b>	<b>B6</b>	<b>B5</b>	<b>B4</b>	<b>B3</b>	<b>B2</b>	<b>B1</b>	<b>B0</b>
Register 90	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	RSV	RSV	RSV	RSV	G20R	G20L

**IDX[6:0]:** 101 1010b (5Ah) Register 90

**G20R: 20-dB Boost for PG2 (Gain Amplifier for AIN1R and AIN2R)**

Default value: 0

This bit is used to boost the microphone signal when the analog input is small.

<b>G12R (REGISTER 124)</b>	<b>G20R (REGISTER 90)</b>	<b>PG2 GAIN</b>
0	0	0 dB (default)
0	1	20 dB
1	0	12 dB
1	1	Reserved

**G20L: 20 dB Boost for PG1 (Gain Amplifier for AIN1L and AIN2L)**

Default value: 0

This bit is used to boost the microphone signal when the analog input is small.

<b>G12L (REGISTER 124)</b>	<b>G20L (REGISTER 90)</b>	<b>PG1 GAIN</b>
0	0	0 dB (default)
0	1	20 dB
1	0	12 dB
1	1	Reserved

**Register 92**

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 92	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	LPAE	RSV	RSV	LGA4	LGA3	LGA2	LGA1	LGA0

**IDX[6:0]:** 101 1100b (5Ch) Register 92

**LPAE: Automatic Attenuation Setting for Bass Boost Gain Control**

Default value: 0

A gain setting for bass boost may cause digital data saturation, depending on the input data level. Where this could occur, LPAE can be used to set the same attenuation level as the bass boost gain level for the digital input data.

LPAE = 0	Disble (default)
LPAE = 1	Enable

**LGA[4:0]: Bass Boost Gain Control**

Default value: 0 0000

These bits are used to set bass boost gain level for the digital data. The center frequency for boost is 200 Hz at 44.1 kHz.

LGA[4:0]	TONE CONTROL GAIN (BASS)	LGA[4:0]	TONE CONTROL GAIN (BASS)
0 0000	0 dB (default)	0 1111	0 dB
0 0011	12 dB	1 0000	-1 dB
0 0100	11 dB	1 0001	-2 dB
0 0101	10 dB	1 0010	-3 dB
0 0110	9 dB	1 0011	-4 dB
0 0111	8 dB	1 0100	-5 dB
0 1000	7 dB	1 0101	-6 dB
0 1001	6 dB	1 0110	-7 dB
0 1010	5 dB	1 0111	-8 dB
0 1011	4 dB	1 1000	-9 dB
0 1100	3 dB	1 1001	-10 dB
0 1101	2 dB	1 1010	-11 dB
0 1110	1 dB	1 1011	-12 dB

**Register 93**

	<b>B15</b>	<b>B14</b>	<b>B13</b>	<b>B12</b>	<b>B11</b>	<b>B10</b>	<b>B9</b>	<b>B8</b>	<b>B7</b>	<b>B6</b>	<b>B5</b>	<b>B4</b>	<b>B3</b>	<b>B2</b>	<b>B1</b>	<b>B0</b>
Register 93	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	RSV	MGA4	MGA3	MGA2	MGA1	MGA0

**IDX[6:0]:** 101 1101b (5Dh) Register 93

**MGA[4:0]: Middle Boost Gain Control**

Default value: 0 0000

These bits are used to set midrange boost gain level for the digital data. The center frequency for boost is 1 kHz.

<b>MGA[4:0]</b>	<b>TONE CONTROL GAIN (MID)</b>	<b>MGA[4:0]</b>	<b>TONE CONTROL GAIN (MID)</b>
0 0000	0 dB (default)	0 1111	0 dB
0 0011	12 dB	1 0000	-1 dB
0 0100	11 dB	1 0001	-2 dB
0 0101	10 dB	1 0010	-3 dB
0 0110	9 dB	1 0011	-4 dB
0 0111	8 dB	1 0100	-5 dB
0 1000	7 dB	1 0101	-6 dB
0 1001	6 dB	1 0110	-7 dB
0 1010	5 dB	1 0111	-8 dB
0 1011	4 dB	1 1000	-9 dB
0 1100	3 dB	1 1001	-10 dB
0 1101	2 dB	1 1010	-11 dB
0 1110	1 dB	1 1011	-12 dB

**Register 94**

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 94	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	RSV	HGA4	HGA3	HGA2	HGA1	HGA0

**IDX[6:0]:** 101 1110b (5Eh) Register 94

**HGA[4:0]:** Treble Boost Gain Control

Default value: 0 0000

These bits are used to set treble boost gain level for the digital data. The center frequency for boost is 5 kHz.

HGA[4:0]	TONE CONTROL GAIN (TREBLE)	HGA[4:0]	TONE CONTROL GAIN (TREBLE)
0 0000	0 dB (default)	0 1111	0 dB
0 0011	12 dB	1 0000	-1 dB
0 0100	11 dB	1 0001	-2 dB
0 0101	10 dB	1 0010	-3 dB
0 0110	9 dB	1 0011	-4 dB
0 0111	8 dB	1 0100	-5 dB
0 1000	7 dB	1 0101	-6 dB
0 1001	6 dB	1 0110	-7 dB
0 1010	5 dB	1 0111	-8 dB
0 1011	4 dB	1 1000	-9 dB
0 1100	3 dB	1 1001	-10 dB
0 1101	2 dB	1 1010	-11 dB
0 1110	1 dB	1 1011	-12 dB

**Register 95**

	<b>B15</b>	<b>B14</b>	<b>B13</b>	<b>B12</b>	<b>B11</b>	<b>B10</b>	<b>B9</b>	<b>B8</b>	<b>B7</b>	<b>B6</b>	<b>B5</b>	<b>B4</b>	<b>B3</b>	<b>B2</b>	<b>B1</b>	<b>B0</b>
Register 95	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	3DEN	RSV	3FL0	3DP3	3DP2	3DP1	3DP0

**IDX[6:0]:** 1011111b (5Fh) Register 95

**3DEN: 3D Sound Effect Enable**

Default value: 0

This bit is used for enabling the 3-D sound effect filter. This filter has two independently controlled parameters.

3DEN = 0	Disable (default)
3DEN = 1	Enable

**3FL0: Filter Selection for 3D Sound**

Default value: 0

This bit is used for selecting from two kinds of filter type, narrow and wide. These filters produce different 3-D effects.

3FL0 = 0	Narrow (default)
3FL0 = 1	Wide

**3DP[3:0]: Efficiency for 3D Sound Effect**

Default value: 0000

These bits are used for adjusting the 3-D sound efficiency. Higher percentages have greater efficiency.

<b>3DP[3:0]</b>	<b>3D Sound Effect Efficiency</b>
0000	0% (default)
0001	10%
0010	20%
0011	30%
0100	40%
0101	50%
0110	60%
0111	70%
1000	80%
1001	90%
1010	100%
1011	Reserved
:	:
1111	Reserved



**Register 96**

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 96	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	NEN2	NEN1	NUP2	NUP1	RSV	RSV	RSV	MXEN

**IDX[6:0]:** 110 0000b (60h) Register 96

**NEN2: Second-Stage Notch Filter Enable**

Default value: 0

The PCM1870A has a two-stage notch filter. The two stages can separately set filter characteristics. This bit is used to enable the second stage.

NEN2 = 0	Disabled (default)
NEN2 = 1	Enabled

**NEN1: First-Stage Notch Filter Enable**

Default value: 0

The PCM1870A has a two-stage notch filter. The two stages can separately set filter characteristics. This bit is used to enable the first stage.

NEN1 = 0	Disabled (default)
NEN1 = 1	Enabled

**NUP2: Second-Stage Notch Filter Coefficients Update**

Default value: 0

This bit is used to update the coefficients for the second-stage notch filter. The coefficients written to registers 101, 102, 103, 104 are updated when NUP2 = 1.

NUP2 = 0	No update (default)
NUP2 = 1	Update

**NUP1: First-Stage Notch Filter Coefficients Update**

Default value: 0

This bit is used to update the coefficients for the first-stage notch filter. The coefficients written to registers 97, 98, 99, 100 are updated when NUP1 = 1.

NUP1 = 0	No update (default)
NUP1 = 1	Update

**MXEN: Digital Monaural Mixing**

Default value: 0

This bit is used to enable or disable monaural mixing in the section that combines L-ch and R-ch digital data.

MXEN = 0	Disabled (stereo, default)
MXEN = 1	Enabled (monaural mixing)

**Registers 97–100**

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 97	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	F107	F106	F105	F104	F103	F102	F101	F100
Register 98	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	F115	F114	F113	F112	F111	F110	F109	F108
Register 99	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	F207	F206	F205	F204	F203	F202	F201	F200
Register 100	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	F215	F214	F213	F212	F211	F210	F209	F208

**IDX[6:0]:** 110 0001b (61h) Register 97

**IDX[6:0]:** 110 0010b (62h) Register 98

**IDX[6:0]:** 110 0011b (63h) Register 99

**IDX[6:0]:** 110 0100b (64h) Register 100

**F[107:100]:** Lower 8 Bits of Coefficient  $a_1$  for First-Stage Notch Filter

**F[115:108]:** Upper 8 Bits of Coefficient  $a_1$  for First-Stage Notch Filter

**F[207:200]:** Lower 8 Bits of Coefficient  $a_2$  for First-Stage Notch Filter

**F[215:208]:** Upper 8 Bits of Coefficient  $a_2$  for First-Stage Notch Filter

Default value: 0000 0000

These bits are used to change the characteristics of the first-stage notch filter. See [Calculating Filter Coefficients](#) for details.

**Registers 101–104**

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 101	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	S107	S106	S105	S104	S103	S102	S101	S100
Register 102	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	S115	S114	S113	S112	S111	S110	S109	S108
Register 103	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	S207	S206	S205	S204	S203	S202	S201	S200
Register 104	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	S215	S214	S213	S212	S211	S210	S209	S208

**IDX[6:0]:** 110 0101b (65h) Register 101

**IDX[6:0]:** 110 0110b (66h) Register 102

**IDX[6:0]:** 110 0111b (67h) Register 103

**IDX[6:0]:** 110 1000b (68h) Register 104

**S[107:100]:** Lower 8 Bits of Coefficient  $a_1$  for Second-Stage Notch Filter

**S[115:108]:** Upper 8 Bits of Coefficient  $a_1$  for Second-Stage Notch Filter

**S[207:200]:** Lower 8 Bits of Coefficient  $a_2$  for Second-Stage Notch Filter

**S[215:208]:** Upper 8 Bits of Coefficient  $a_2$  for Second-Stage Notch Filter

Default value: 0000 0000

These bits are used to change the characteristics of the second-stage notch filter. See [Calculating Filter Coefficients](#) for details.

### Calculating Filter Coefficients

The PCM1870A provides a dual-stage notch filter at the digital output of the ADC. The filter characteristics of each filter stage can be programmed. The characteristics are determined by calculating coefficients for three parameters, sampling frequency, center frequency and bandwidth, as shown in the following equations. All coefficients must be written as 2s-complement binary data into registers 97, 98, 99, 100, 101, 102, 103, and 104.

$f_S$  = Sampling frequency [Hz]

$f_C$  = Center frequency [Hz]

$f_b$  = Bandwidth [Hz]

$$a_1 = - (1 + a_2) \cos\left(\frac{2\pi f_C}{f_S}\right)$$

$$a_2 = \frac{1 - \tan\left(\frac{2\pi f_b / f_S}{2}\right)}{1 + \tan\left(\frac{2\pi f_b / f_S}{2}\right)}$$

### Register 124

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 124	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	RSV	RSV	RSV	RSV	G12R	G12L

**IDX[6:0]:** 111 1100b (7Ch) Register 124

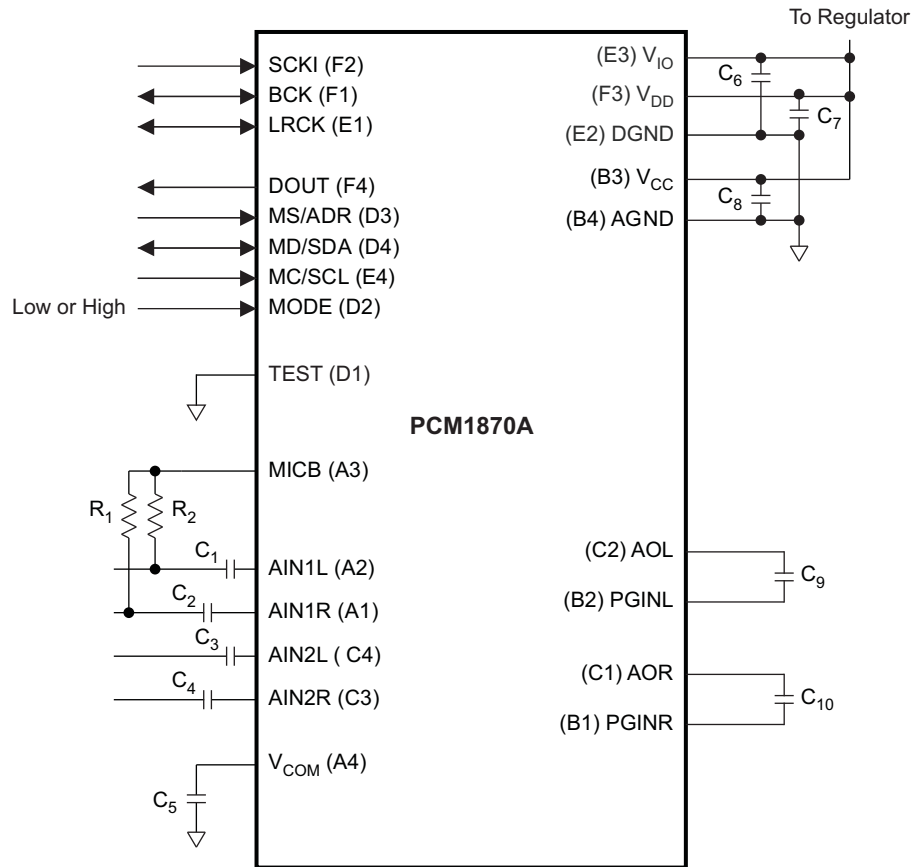
**G12R: 12-dB Boost for PG2 (Gain Amplifier for AIN1R and AIN2R)**

**G12L: 12-dB Boost for PG1 (Gain Amplifier for AIN1L and AIN2L)**

Default value: 0

These bits are used to boost the microphone signal when the analog input is small. See [Register 90](#) for the detailed settings.

CONNECTION DIAGRAM



S0262-01

Figure 24. Connection Diagram

Table 10. Recommended External Parts

C <sub>1</sub> –C <sub>4</sub>	1 μF–10 μF	C <sub>9</sub> , C <sub>10</sub>	1 μF–10 μF
C <sub>5</sub>	1 μF–4.7 μF	R <sub>1</sub> , R <sub>2</sub>	2.2 kΩ
C <sub>6</sub>	0.1 μF		
C <sub>7</sub>	1 μF		
C <sub>8</sub>	1 μF–4.7 μF		

## BOARD DESIGN AND LAYOUT CONSIDERATIONS

### $V_{CC}$ , $V_{DD}$ , and $V_{IO}$ Pins

The digital and analog power supply lines to the PCM1870A should be bypassed to the corresponding ground pins with 0.1- to 4.7- $\mu$ F ceramic capacitors or electrolytic capacitors, placed as close to the pins as possible to maximize the dynamic performance of ADC.

### AGND and DGND Pins

To maximize the dynamic performance of the PCM1870A, the analog and digital grounds are not connected internally. These grounds should have very low impedance to avoid digital noise feeding back into the analog ground. So, they should be connected directly to each other under the part to reduce the potential of noise problems.

### AIN1L, AIN1R, AIN2L, and AIN2R Pins

AIN1L, AIN1R, AIN2L, and AIN2R are single-ended inputs. AIN1L and AIN1R can also be used as a monaural differential input. The anti-aliasing low-pass filters are integrated on these inputs to remove the out-of-band noise from the audio. If the performance of these filters is not good enough for an application, appropriate external anti-aliasing filters are needed. The passive RC filter (100  $\Omega$  and 0.01  $\mu$ F to 1 k $\Omega$  and 1000 pF) is used in general. Any pins that are not used in an application should be left open. Do not select open pins through register settings.

### AOL, AOR, PGINL, and PGINR Pins

When AIN1L, AIN1R, AIN2L, and AIN2R pins are used as microphone inputs with high gain, AOL and AOR may have a large dc offset. It is recommended to locate a dc-blocking capacitor (1- to 10- $\mu$ F capacitor) between AOL/AOR and PGINL/PGINR. If an application is not affected by dc offset, the PCM1870A does not need the capacitors.

### $V_{COM}$ Pin

1- $\mu$ F to 4.7- $\mu$ F capacitor is recommended between  $V_{COM}$  and AGND to ensure low source impedance for the ADC common voltage. This capacitor should be located as close as possible to the  $V_{COM}$  pin to reduce dynamic errors on the ADC common voltage.

### BCK (Master Mode) and DOUT Pins

BCK in the master mode and DOUT have adequate load drive capability, but if the BCK and DOUT lines are long, locating a buffer near the PCM1870A and minimizing load capacitance is recommended in order to minimize crosstalk between digital and analog circuits, maximize the dynamic performance of the ADC, and reduce power consumption.

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)	Op Temp (°C)	Top-Side Markings (4)	Samples
PCM1870AYZFR	ACTIVE	DSBGA	YZF	24	2000	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM	-40 to 85	PCM1870A	<a href="#">Samples</a>
PCM1870AYZFT	ACTIVE	DSBGA	YZF	24	250	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM	-40 to 85	PCM1870A	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

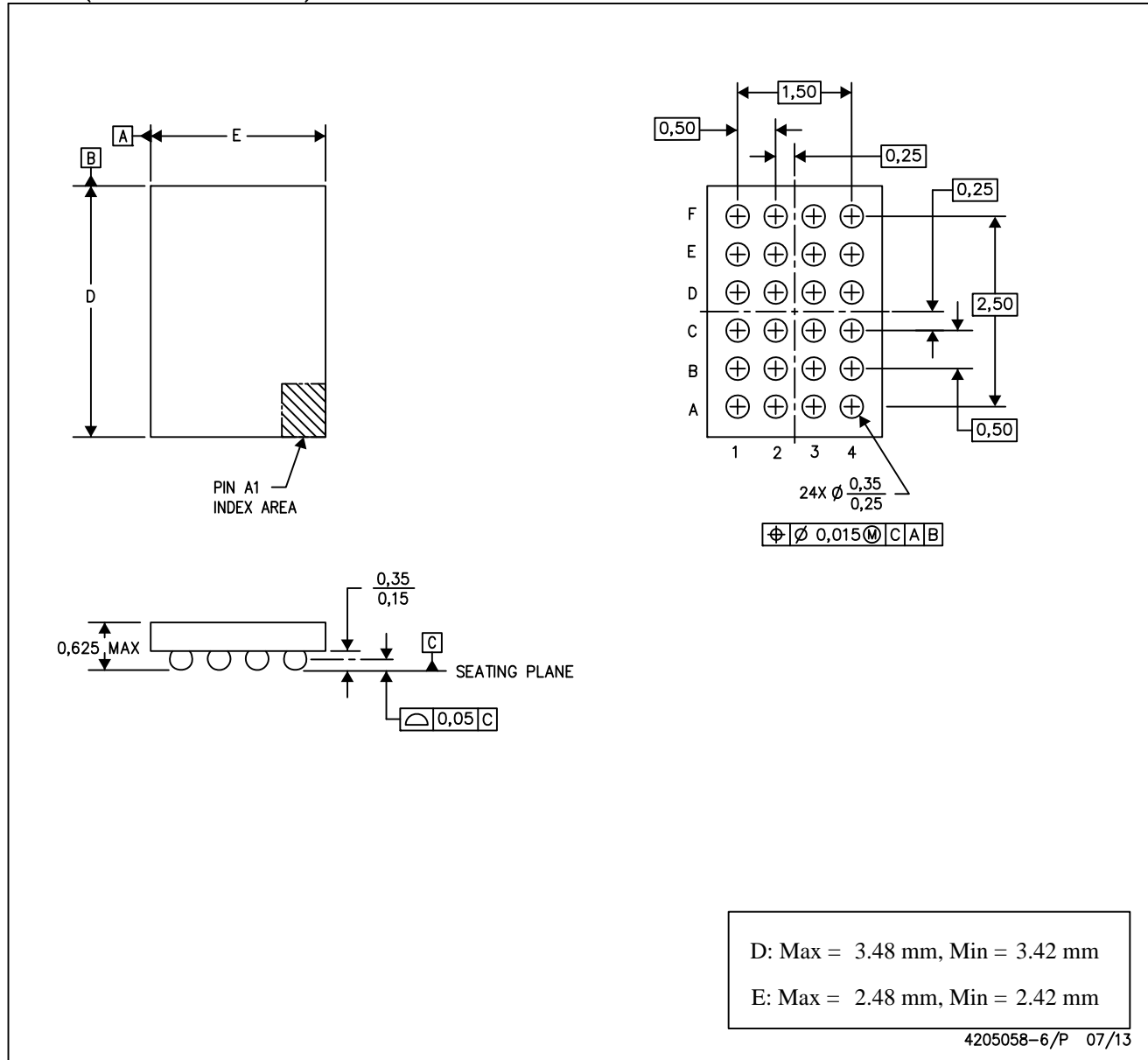
(4) Multiple Top-Side Markings will be inside parentheses. Only one Top-Side Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Top-Side Marking for that device.

**Important Information and Disclaimer:** The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

YZF (R-XBGA-N24)

DIE-SIZE BALL GRID ARRAY



- NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.  
 B. This drawing is subject to change without notice.  
 C. NanoFree™ package configuration.

NanoFree is a trademark of Texas Instruments.

## IMPORTANT NOTICE

Texas Instruments Incorporated (TI) reserves the right to make corrections, enhancements, improvements and other changes to its semiconductor products and services per JESD46, latest issue, and to discontinue any product or service per JESD48, latest issue. Buyers should obtain the latest relevant information before placing orders and should verify that such information is current and complete.

TI's published terms of sale for semiconductor products (<http://www.ti.com/sc/docs/stdterms.htm>) apply to the sale of packaged integrated circuit products that TI has qualified and released to market. Additional terms may apply to the use or sale of other types of TI products and services.

Reproduction of significant portions of TI information in TI data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations, and notices. TI is not responsible or liable for such reproduced documentation. Information of third parties may be subject to additional restrictions. Resale of TI products or services with statements different from or beyond the parameters stated by TI for that product or service voids all express and any implied warranties for the associated TI product or service and is an unfair and deceptive business practice. TI is not responsible or liable for any such statements.

Buyers and others who are developing systems that incorporate TI products (collectively, "Designers") understand and agree that Designers remain responsible for using their independent analysis, evaluation and judgment in designing their applications and that Designers have full and exclusive responsibility to assure the safety of Designers' applications and compliance of their applications (and of all TI products used in or for Designers' applications) with all applicable regulations, laws and other applicable requirements. Designer represents that, with respect to their applications, Designer has all the necessary expertise to create and implement safeguards that (1) anticipate dangerous consequences of failures, (2) monitor failures and their consequences, and (3) lessen the likelihood of failures that might cause harm and take appropriate actions. Designer agrees that prior to using or distributing any applications that include TI products, Designer will thoroughly test such applications and the functionality of such TI products as used in such applications.

TI's provision of technical, application or other design advice, quality characterization, reliability data or other services or information, including, but not limited to, reference designs and materials relating to evaluation modules, (collectively, "TI Resources") are intended to assist designers who are developing applications that incorporate TI products; by downloading, accessing or using TI Resources in any way, Designer (individually or, if Designer is acting on behalf of a company, Designer's company) agrees to use any particular TI Resource solely for this purpose and subject to the terms of this Notice.

TI's provision of TI Resources does not expand or otherwise alter TI's applicable published warranties or warranty disclaimers for TI products, and no additional obligations or liabilities arise from TI providing such TI Resources. TI reserves the right to make corrections, enhancements, improvements and other changes to its TI Resources. TI has not conducted any testing other than that specifically described in the published documentation for a particular TI Resource.

Designer is authorized to use, copy and modify any individual TI Resource only in connection with the development of applications that include the TI product(s) identified in such TI Resource. NO OTHER LICENSE, EXPRESS OR IMPLIED, BY ESTOPPEL OR OTHERWISE TO ANY OTHER TI INTELLECTUAL PROPERTY RIGHT, AND NO LICENSE TO ANY TECHNOLOGY OR INTELLECTUAL PROPERTY RIGHT OF TI OR ANY THIRD PARTY IS GRANTED HEREIN, including but not limited to any patent right, copyright, mask work right, or other intellectual property right relating to any combination, machine, or process in which TI products or services are used. Information regarding or referencing third-party products or services does not constitute a license to use such products or services, or a warranty or endorsement thereof. Use of TI Resources may require a license from a third party under the patents or other intellectual property of the third party, or a license from TI under the patents or other intellectual property of TI.

TI RESOURCES ARE PROVIDED "AS IS" AND WITH ALL FAULTS. TI DISCLAIMS ALL OTHER WARRANTIES OR REPRESENTATIONS, EXPRESS OR IMPLIED, REGARDING RESOURCES OR USE THEREOF, INCLUDING BUT NOT LIMITED TO ACCURACY OR COMPLETENESS, TITLE, ANY EPIDEMIC FAILURE WARRANTY AND ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE, AND NON-INFRINGEMENT OF ANY THIRD PARTY INTELLECTUAL PROPERTY RIGHTS. TI SHALL NOT BE LIABLE FOR AND SHALL NOT DEFEND OR INDEMNIFY DESIGNER AGAINST ANY CLAIM, INCLUDING BUT NOT LIMITED TO ANY INFRINGEMENT CLAIM THAT RELATES TO OR IS BASED ON ANY COMBINATION OF PRODUCTS EVEN IF DESCRIBED IN TI RESOURCES OR OTHERWISE. IN NO EVENT SHALL TI BE LIABLE FOR ANY ACTUAL, DIRECT, SPECIAL, COLLATERAL, INDIRECT, PUNITIVE, INCIDENTAL, CONSEQUENTIAL OR EXEMPLARY DAMAGES IN CONNECTION WITH OR ARISING OUT OF TI RESOURCES OR USE THEREOF, AND REGARDLESS OF WHETHER TI HAS BEEN ADVISED OF THE POSSIBILITY OF SUCH DAMAGES.

Unless TI has explicitly designated an individual product as meeting the requirements of a particular industry standard (e.g., ISO/TS 16949 and ISO 26262), TI is not responsible for any failure to meet such industry standard requirements.

Where TI specifically promotes products as facilitating functional safety or as compliant with industry functional safety standards, such products are intended to help enable customers to design and create their own applications that meet applicable functional safety standards and requirements. Using products in an application does not by itself establish any safety features in the application. Designers must ensure compliance with safety-related requirements and standards applicable to their applications. Designer may not use any TI products in life-critical medical equipment unless authorized officers of the parties have executed a special contract specifically governing such use. Life-critical medical equipment is medical equipment where failure of such equipment would cause serious bodily injury or death (e.g., life support, pacemakers, defibrillators, heart pumps, neurostimulators, and implantables). Such equipment includes, without limitation, all medical devices identified by the U.S. Food and Drug Administration as Class III devices and equivalent classifications outside the U.S.

TI may expressly designate certain products as completing a particular qualification (e.g., Q100, Military Grade, or Enhanced Product). Designers agree that it has the necessary expertise to select the product with the appropriate qualification designation for their applications and that proper product selection is at Designers' own risk. Designers are solely responsible for compliance with all legal and regulatory requirements in connection with such selection.

Designer will fully indemnify TI and its representatives against any damages, costs, losses, and/or liabilities arising out of Designer's non-compliance with the terms and provisions of this Notice.